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ELECTRICAL INSTRUMENTS AND TESTING

• N. H. SCHNEIDER •

• JESSE HARGRAVE •



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**ELECTRICAL INSTRUMENTS
AND
TESTING**

BY

NORMAN H. SCHNEIDER

AUTHOR OF "ELECTRIC LIGHT FOR THE FARM," "MANAGEMENT OF ELECTRIC
POWER PLANTS," ETC., ETC.

INCLUDING

TESTING TELEGRAPH WIRES AND CABLES

AND

LOCATING FAULTS

BY

JESSE HARGRAVE

ASSISTANT ELECTRICAL ENGINEER POSTAL TELEGRAPH CABLE COMPANY

FOURTH EDITION, REVISED AND ENLARGED

NEW YORK :

SPON & CHAMBERLAIN, 123-125 Liberty Street

LONDON :

E. & F. N. SPON, LIMITED, 57 Haymarket, S. W.,

1913

Eng. 4879.13.7

May 16, 1913
HARVARD UNIVERSITY
SCHOOL OF ENGINEERING.

203.

JUN 20 1917
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BURR PRINTING HOUSE
16-18 JACOB STREET, NEW YORK, U.S.A.

PREFACE

TO THE

FOURTH EDITION.

IN presenting this fourth edition, the author takes occasion to express his appreciation of the widespread and gratifying interest that has been shown in this book since its first appearance. In order to extend still further its usefulness the book has been carefully revised and much enlarged, always considering as paramount the needs of the practical worker.

Primarily, this book is intended as introductory to the larger and more mathematical works on the subject. It is expected, however, that the reader will also gain from it a thorough practical knowledge of the principles of electrical testing as applied to everyday conditions.

The importance of this knowledge is being emphasized more every day.

Broadly speaking, all testing is to discover defects, whether it be to locate a break in a line or to insure that a generator is up to voltage. Its ultimate object is to restore or maintain the maximum of efficiency.

The electrical worker who is familiar with the standard methods of testing becomes at once more valuable. He becomes an important factor in securing higher efficiency, and thereby leading to greater profit, or what is often as valuable, decreased expense.

In the process of thorough revision which this book has undergone the principal changes have been the addition of much important new text, illustrations and tables and the omission of extended descriptions of apparatus of limited application, or of instruments now in little use.

Prominent among the added text is that descriptive of instruments which, although recently introduced, are emphasizing their worth. This includes, amongst others, the Megger, the testing sets and galvanometers of Thompson Levering, the compact telephone testing cabinet of the Western Electric Company, and the new Roller Smith instruments.

In the testing work, much new text has been added, several chapters entirely rewritten, with new illustrations, and a number of pages added on testing in the telephone exchange.

The many uses of the voltmeter in testing resistances have been more fully discussed, with illustrations of new voltmeters suitable for this particular use.

Exceedingly useful tables on wires and cables have been added, particular attention being called to those treating on resistance wires and to those on copper wire.

One of the most valuable of the tables is that on the physical properties of copper wire, which has been prepared from the comprehensive booklet by Mr. S. W. Stratton, for the United States Government, Department of Commerce and Labor.

The author takes pleasure at this time in expressing his acknowledgments to those who assisted him in the preparation of the new edition—to Mr. O. T. Louis for hints on the selection of galvanometers; to Mr. W. J. Thompson for information concerning bridge-testing sets; to Mr. J. W. Roller for the interest taken in securing the cuts and information on the Roller-Smith and Whitney apparatus, and to Mr. W. D. Lindsey of the Western Electric Company for assistance in preparing the text on telephone line testing and on lead-covered cable.

Acknowledgment is also due for the loan of cuts and for data to the General Electric Company, Westinghouse Electric Manufacturing Company, the Weston Electrical Instrument Company, the O. T. Louis Company, and others noted elsewhere in the book.

NORMAN H. SCHNEIDER,

FEBRUARY, 1913.

Cleveland, Ohio.

PREFACE TO THIRD EDITION.

Since this work was first published it has found a ready sale among telegraph and telephone workers, and it was pointed out that the testing of telegraph and telephone wires was not as fully covered as it might have been. With a view to making the book still more attractive and useful to workers in those important branches of the electrical field, Mr. Jesse Hargrave, Assistant Electrical Engineer of the Postal Telegraph Cable Company, was asked to contribute two chapters devoted to the testing of telegraph and telephone wires and cables, and he agreed to undertake the work.

In compiling the matter contained in Chapters XII and XIII Mr. Hargrave has endeavored to carry out Mr. Schneider's ideas of clearness and simplicity. The testing of the wires is taken up in its first stage—the first test of the day, the early-morning test—and the student is taken step by step through the different methods of testing for faults, insulation and conductivity, cable faults, etc.

The different tests and measurements, aside from being described in the simplest possible language, are treated from a strictly practical standpoint, Mr. Hargrave having made use of them in actual practice.

Thanks are due Messrs. Leeds & Northrup for sketches, etc.

PREFACE.

This book is intended for practical use and also as an introduction to the larger works on Electrical Testing.

The apparatus described is modern and universally adopted.

The tests are such as occur daily in the work of the engine room, power house or the technical school.

The illustrations except those of instruments by specified makers are drawn for this book. Detail unnecessary to the subject is omitted, such can be well studied in the apparatus itself.

The formulas given are explained in the text preceding them, and examples are worked out in simple arithmetic. Formulas are often looked upon as intricate mysteries by those who do not understand them. In reality they are often very simple and very necessary. The use of formulas is but a method of arithmetical shorthand. A few minutes devoted to a formula will render it clear and its use easy to those of even limited mathematical education.

The thanks of the author are due to Mr. O. T. Louis for cuts and suggestions, to Mr. F. W. Roller for information on his improved type of hot wire instruments; to Messrs. Willyoung & Gibson, The Weston Electrical Instrument Co., The Keystone Electrical Instrument Co., The General Electric Co., and others.

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INTRODUCTION.

Classification. Electrical instruments used in the detection or measurement of electrical currents are divisible into two classes.

The first class includes such as show the presence of a current without directly indicating its value in units of measurement.

To this class belong the various kinds of galvanometers.

Instruments of the second class indicate by the movement of pointers over dials or some such means the value of the electrical current passing.

The dial under the pointer or index is graduated into divisions representing units of measurement. Examples of this class are the voltmeter, ammeter and wattmeter.

Instruments which leave a record of the fluctuations of the current strength for a given period of time are Recording Voltmeters, Recording Watt-

meters according to the unit by which their records are read.

The two general purposes in electrical testing are to determine the condition of a circuit or apparatus, and to measure in electrical units the value of a resistance, a current strength, or an electromotive force.

The first determinations may be made with a galvanometer, a Wheatstone bridge, or some special form of apparatus.

The second case most conveniently requires the use of direct reading apparatus such as an ammeter or a voltmeter, the measurement of resistance units belonging by right to the first case. Much of the apparatus is interchangeable, as will be seen in the succeeding pages.

Electrical tests may again be grouped as to the degree of refinement required in the results. Where the object is to obtain the closest possible reading, it becomes a matter for the laboratory, and refined methods which require much expensive apparatus. But where readings are required of such accuracy as would be needed in every-day commercial electrical work, the operations may be classed as commercial testing.

In the following pages, the subject is treated more from the commercial aspect than from that of the laboratory, bringing the subject into the hands of the many rather than the few.

CHAPTER I.

GALVANOMETERS.

The Simple Galvanometer. Perhaps the simplest form of galvanometer is shown in Fig. 1, and although home made, is capable of many uses.

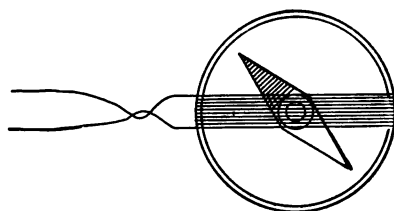


FIG. 1.

An ordinary mariner's compass has a few turns of fine insulated copper wire wound around its case, a suitable gauge being No. 26 B. & S. or No. 28 S.W.G.

If current from a battery or other source be sent through the coil of wire, the magnetic needle will tend to turn at an angle to the wire turns, providing the galvanometer has first been set so that both needle and wire turns are parallel.

The stronger the current through the coil, the farther the needle swings out from it.

And reversing the connections will cause the needle to swing in a reversed direction as the current is reversed in its flow around the coil.

But it will be noticed that the amount of the needle swing is not in proportion to the strength of the current.

Let the circle or card forming the bottom of the compass case be divided into 360 degrees having the zero where one end of the needle normally rests.

On connecting the coil ends to a dry cell suppose the needle swings out to 30 degrees.

If a second cell is put in series with the first and the two cells connected to the coil as before the swing may now only be 35 degrees, a gain of but five degrees by doubling the current strength instead of a gain of 30 degrees.

It is to be remarked before proceeding that these readings as they are termed are only for sake of example, not being necessarily exactly the number of degrees obtained in any experiment.

If a galvanometer gives a large reading with one cell as compared with a reading taken under precisely similar conditions on another galvanometer the one giving the largest reading will be said to have the greater sensibility.

The movement of a galvanometer needle from one position to another is termed its deflection and is measured in degrees.

Now the primary laws of electricity teach that the current strength or rate of flow depend upon the electromotive force (e.m.f.) or voltage impelling that current and upon the resistance of the circuit through which it travels.

Increasing the e.m.f. or decreasing the resistance increases the current flow, and vice versa, decreasing the e.m.f. or increasing the resistance decreases the current flow.*

In order to decrease the deflection of the needle resistance must be added to the circuit, this may be done by means of a rheostat.

One form of rheostat would consist of coils of wire so connected that one or any desired number of coils could be inserted in series with the galvanometer. The electrical energy expended in these coils being therefore not available in the galvanometer coils would have no effect upon its needle deflection which would be thereby reduced.

The sensibility of this galvanometer will depend upon the ampere turns of its coil (neglecting friction, etc.), and the coil winding should be suitable for the current in the circuit in which it is used.

If the galvanometer is to measure currents in a circuit of low resistance, its coil should be also of low resistance and vice versa.

The meaning of ampere turns must be thoroughly understood. It will suffice here to remark that one ampere traveling round the needle for

*See "Electricity and Its Laws, for Beginners," for the elementary laws governing electric current.

one turn would equal one hundredth of an ampere for one hundred turns and the needle deflection would be the same in each case.

In a circuit of low resistance, a large current would flow (presumably) and this current would not need to travel around the needle as many times as the lesser current in a circuit of higher resistance to produce the same deflection. In this case using a coil of many fine wire turns in the low resistance circuit would only reduce the current around the needle. And using a coil of large wire and few turns would not give ampere turns enough in a circuit of high resistance.

An apparent contradiction may be met by the experimenter who in adding more turns of wire to a simple galvanometer gets a greater deflection of the needle. This, however, only shows that the ampere turns were not sufficient in the first case, and therefore the contradiction does not exist.

To sum up, the reason for using a large number of fine wire turns as is done in some galvanometers is that the current strength is often so small as to require its circulation many times around the needle to produce the requisite ampere turns.

It is very common practice to speak of a coil as being wound to a given resistance when ampere turns are really desired. If resistance were the object, a resistance metal could be used to advantage.

The relation of resistance and diameter of copper wire is given in Table I., from which may be

TABLE I
(Bureau of Standards. S. W. Stratton, Director)
STANDARD ANNEALED COPPER WIRE
American Wire Gage (B. & S.)

Gage No.	Diameter in Mils	Circular Mils	Ohms per 1000 Feet			
			0° C (=32° F)	15° C (=59° F)	20° C (=68° F)	25° C (=77° F)
0000	460.0	211 600.	0.045 14	0.048 04	0.049 01	0.049 98
000	409.6	167 800.	.056 93	.060 58	.061 80	.063 03
00	364.8	133 100.	.071 78	.076 39	.077 93	.079 47
0	324.9	105 500.	.090 52	.096 33	.098 27	.1002
1	289.3	83 690.	.1141	.1215	.1239	.1264
2	257.6	66 370.	.1439	.1532	.1563	.1594
3	229.4	52 640.	.1815	.1931	.1970	.2009
4	204.3	41 740.	.2288	.2436	.2485	.2534
6	162.0	26 250.	.3639	.3872	.3951	.4029
8	128.5	16 510.	.5786	.6158	.6282	.6406
9	114.4	13 090.	.7296	.7765	.7921	.8078
10	101.9	10 380.	.9200	.9792	.9989	1.019
12	80.81	6530.	1.463	1.557	1.588	1.620
13	71.96	5178.	1.845	1.963	2.003	2.042
14	64.08	4107.	2.326	2.475	2.525	2.576
15	57.07	3257.	2.983	3.121	3.184	3.248
16	50.82	2588.	3.699	3.936	4.015	4.095
17	45.26	2048.	4.664	4.963	5.061	5.161
18	40.30	1624.	5.881	6.259	6.365	6.512
19	35.89	1288.	7.416	7.892	8.051	8.210
20	31.96	1022.	9.352	9.953	10.15	10.35
21	28.46	810.1	11.79	12.55	12.80	13.06
22	25.35	642.4	14.87	15.82	16.14	16.46
23	22.57	509.5	18.75	19.95	20.36	20.76
24	20.10	404.0	23.64	25.16	25.67	26.18
25	17.90	320.4	29.81	31.73	32.37	33.01
26	15.94	254.1	37.59	40.01	40.82	41.62
27	14.20	201.5	47.40	50.45	51.46	52.48
28	12.64	159.8	59.77	63.61	64.90	66.18
29	11.26	126.7	75.37	80.22	81.84	83.46
30	10.03	100.5	95.05	101.2	103.2	105.2
31	8.928	79.70	119.8	127.6	130.1	132.7
32	7.950	63.21	151.1	160.8	164.1	167.3
33	7.080	50.13	190.6	202.8	206.9	211.0
34	6.305	39.75	240.3	255.7	260.9	266.1
35	5.615	31.52	303.0	322.5	329.0	335.5
36	5.000	25.00	382.1	406.6	414.8	423.0
37	4.453	19.83	481.8	512.8	523.1	533.5
38	3.965	15.72	607.5	646.6	659.6	672.7
39	3.531	12.47	766.1	815.4	831.8	848.2
40	3.145	9.888	956.1	1028.	1049.	1070.

For British Standards see Table VIII.

computed the size of wire to use for a galvanometer coil.

Deflections not Uniform. The needle does not move twice as far for twice the current because as soon as it starts to move it commences to get out of the strongest influence of the coil, or out of the intense part of the field.

If the coil is very large and the needle very small, then the latter can deflect more uniformly as the field is more uniform.

Tangent Galvanometer. The tangent galvanometer has a coil wound around a ring perhaps twelve inches in diameter, and the needle is a fraction of an inch in length.

In this case the deflections of the needle are compared by referring to a table of tangents, Table II.

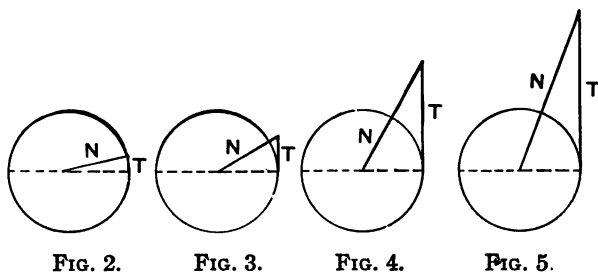
A tangent is a straight line perpendicular to the diameter of a circle, limited in length by the sides of the angle one of which is the diameter of the circle.

In Fig. 2 the dotted line is the diameter of the circle and passes clear through the zero mark. *N* is a straight line drawn through the centre of the pointer or needle, in other words shows the number of degrees of deflection. The degrees between this line and the zero will be the number of degrees forming the angle of deflection, angles being measured in degrees. If the pointer were at

45° the angle would be one of 45° or half a right angle (90°).

T is a straight line perpendicular to the diameter but cut by N . The portion between the diameter line and where it is cut by N is the tangent. Fig. 2 represents an angle of about 15° , that is, the needle is deflected 15° . Fig. 3 shows the tangent of an angle of about 30° , Fig. 4 the tangent of 60° . and Fig. 5 the tangent of about 75° .

It will be seen that the tangent increases in



length very rapidly, far more so than the deflection angle.

To make a comparison, let the needle be deflected 5° , its tangent will be .0875, double the deflection, the tangent of 10° is .1763, more than double the one of 5° .

Deflect the needle to 30° or six times its first deflection and the tangent is .5774, about seven times.

A still farther deflection to 60° , or twelve times

the first one, the tangent is now 1.7321, or about twenty times that of 5°.

In a tangent galvanometer when it is desired to compare the strength of two currents, readings

TABLE II.
Natural Tangents of Angles.

Angle Degrees	Tangent	Angle Degrees	Tangent	Angle Degrees	Tangent
1	.0174	31	.6008	61	1.8040
2	.0349	32	.6248	62	1.8807
3	.0524	33	.6494	63	1.9626
4	.0699	34	.6745	64	2.0503
5	.0874	35	.7002	65	2.1445
6	.1051	36	.7265	66	2.2460
7	.1227	37	.7535	67	2.3559
8	.1405	38	.7812	68	2.4751
9	.1583	39	.8097	69	2.6051
10	.1768	40	.8391	70	2.7475
11	.1943	41	.8692	71	2.9042
12	.2125	42	.9004	72	3.0777
13	.2308	43	.9325	73	3.2709
14	.2493	44	.9656	74	3.4879
15	.2679	45	1.0000	75	3.7321
16	.2867	46	1.0355	76	4.0108
17	.3057	47	1.0724	77	4.3315
18	.3249	48	1.1106	78	4.7046
19	.3443	49	1.1504	79	5.1446
20	.3639	50	1.1918	80	5.6713
21	.3838	51	1.2349	81	6.3138
22	.4040	52	1.2799	82	7.1154
23	.4244	53	1.3270	83	8.1443
24	.4452	54	1.3764	84	9.5144
25	.4663	55	1.4281	85	11.4301
26	.4887	56	1.4826	86	14.3007
27	.5095	57	1.5399	87	19.0811
28	.5317	58	1.6003	88	28.6863
29	.5543	59	1.6643	89	57.2900
30	.5773	60	1.7321	90	infinite

are made of the two deflections obtained. The tangents of these angles or deflections are then compared.

If one deflection were 11° and the second one

63°, the second current would be nearly ten times the greater, or as .1944 is to 1.9626.

Influence of the Earth's Magnetic Field. It was necessary to move the galvanometer around so as to bring the needle point to zero.

As the galvanometer was made out of a compass this was done because the needle tended to point to the magnetic *N* and *S* of the earth.

In order to avoid having to turn the instrument about before using it, the influence of the earth's

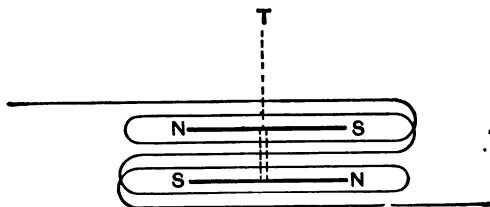


FIG. 6.

magnetic field is compensated for in the astatic galvanometer.

The Astatic Galvanometer. In Fig. 6 two magnetic needles are rigidly attached to a stem in such manner that the *N* pole of one is over the *S* pole of the other. If both needles are equally strong, such a combination on being suspended by a fine thread *T* or balanced on a point, would stay in whatever direction it was placed. It would be entirely unaffected by the earth's mag-

netic field, one needle neutralizing the other. But this is not obtained in practice, the pair of needles take a set position in which, however, their movement is little restrained by the earth's magnetism.

The coil is wound as shown in order that the

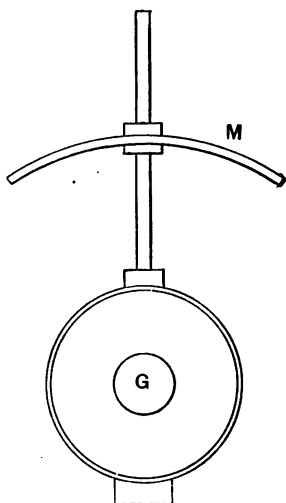


FIG. 7.

current may tend to turn each needle in the same direction.

Compensating Magnet. A compensating magnet is used with certain forms of galvanometers to neutralize the effect of the earth's magnetism on the needle.

It is generally of a curved form as in Fig. 7 at *M*, being mounted on a rod upon which it can be slid up or down over the galvanometer *G*. It can also be rotated on its axis.

An ordinary bar magnet is used for the same purpose by placing it in different positions with relation to the galvanometer until the desired result is obtained.

Where the latter method is pursued it is a good plan to attach the magnet to a block of lead so that a slight jar will not displace it.

Laws of Resistances. Having briefly surveyed the principles of the galvanometer and a few of the electrical laws relating to testing, it is of importance to have a clear idea of resistance and conductivity.

Resistance is the opposite of conductivity. A substance freely conducting an electric current would have a low resistance but a high conductivity. Copper therefore, having a high conductivity, offers a low resistance. Resistance depends upon several factors. First, upon the specific resistance of the substance, that is, upon the particular resistance of the substance as compared with that of another substance. Taking the specific resistance of gold as 1, iron would be approximately 7. Gold would have seven times the conductivity of iron, but iron would have seven times the specific resistance of gold. Resistance also depends upon the area and length of the piece of metal being tested. For example, a metal



FIG. 7A. The Gurley Tangent Galvanometer.

rod one inch square and one foot long would have four times the resistance of a rod two inches square but of the same length. The area in the latter case is four times that of the one inch rod.

EXAMPLE.—To prove this connect a piece of thin iron or german silver wire in circuit between the tangent galvanometer or any other galvanometer that is at hand, and a cell of battery. Upon completing the circuit a deflection will result which is to be noted. Then remove the wire and substitute a piece of wire of the same length but larger in diameter. The resulting deflection will be greater. The larger wire has the greater conductivity and has permitted the flow of a greater current as shown by the increased deflection. The smaller wire has the greater resistance as shown by the lesser deflection.

NOTE. Both wires must be of sufficiently high resistance to cause a readable deflection on the galvanometer. This will of course depend upon the galvanometer and battery used.

Likewise a one inch rod one foot in length would be one-half the resistance of one having the same area but which is two feet long.

EXAMPLE.—Connect up the thin wire, the galvanometer and the battery and note the deflection of the needle. Then cut off half the wire and connect it up again and the deflection will be greater. The length having been lessened, the resistance is lessened and the conductivity is increased, permitting a greater current to flow.

The total resistance of a number of resistances

connected together in series will be merely the sum of them all, but where they are connected in multiple or parallel the result will be different.

Where the resistances are all of equal value, the multiple resistance will be that of one divided by the number of them.

EXAMPLE.—Find the resistance of 6 wires connected in multiple, each having 6 ohms resistance. Then 6 ohms, the resistance of one, divided by 6, the number of wires, will equal 1 ohm, which is their combined resistance.

EXAMPLE.—Find the resistance of 9 wires in multiple, each of 3 ohms resistance. Then 3 divided by 9 equals .33 and a fraction.

Where the wires are of unequal resistance other methods must be pursued.

For two unequal resistances in multiple, divide their product by their sum, that is, multiply them together and divide the result by the figure obtained by adding them together.

EXAMPLE.—One wire measures 12 ohms and the other 4 ohms. Find their combined resistance. 12 times 4 equals 48. 12 plus 4 equals 16. 48 divided by 16 equals 3, the combined resistance in ohms.

Where there are more than two unequal resistances in multiple two methods are possible.

First Method. Find the resistance of any two, then pair this result with another; find the answer

and pair this again with another resistance; and so on.

EXAMPLE.—Given four wires of 12, 4, 8, and 16 ohms respectively. The first pair is 12 and 4, which was found in the last example to equal 3 ohms. This 3 paired with 8 equals 2.2 approximately. Pairing 2.2 with 16 gives 1.9 approximately. But this is a roundabout way and the better method is as follows:

Second Method. Find the reciprocal of the sum of the reciprocals of all the resistances. The reciprocal of 12 is $\frac{1}{12}$, the reciprocal of 4 is $\frac{1}{4}$. The reciprocal of 8 is $\frac{1}{8}$ and the reciprocal of 16 is $\frac{1}{16}$. The sum of all these is

$$\frac{1}{12} + \frac{1}{4} + \frac{1}{8} + \frac{1}{16} \text{ or } \frac{4 + 12 + 6 + 3}{48} = \frac{25}{48}$$

The reciprocal of $\frac{25}{48}$ is $\frac{48}{25}$ or 1.9 ohm as before.

There are a number of different methods by which the actual measurement of resistances may be made. The examples described with the galvanometer may be worked out with an ammeter, in which case by ohms law the actual resistance may be computed. The most commonly adopted methods in practical use are those employing the Wheatstone bridge, the direct deflection method by means of a galvanometer, and those using some one of the forms of instruments which read directly in ohms.

CHAPTER II.

GALVANOMETERS.

In describing galvanometers, the terms constant, figure of merit and sensibility are used.

A full description of these terms rightly belongs to the pages on practical testing.

But it may be stated here, that these terms are those by which galvanometers are compared when considering their sensitiveness.

The constant or sensibility is the number of ohms resistance through which an e.m.f. of one volt will cause the galvanometer to give a deflection of one degree on a standard scale.

A more complete explanation of the terms will be found in a later chapter.

The Thompson Reflecting Galvanometer. In the Thompson reflecting galvanometer, a flat mirror of less than one-half inch in diameter has a number of fine steel magnets fastened on its back. This mirror-magnet combination is suspended by a fibre of cocoon silk inside a coil of many turns of fine insulated wire.

When the magnets are deflected by current flowing in the coil, the mirror also turns, and a beam of light directed on it is reflected on a curved scale.



FIG. 8.

This scale has the zero mark in the centre, readings being thus possible in either direction.

A compensating magnet is attached on a rod fixed to the top of the case and leveling screws are provided to level the instrument.

As the mirror is small and light it has little momentum, its moving in a small chamber gives it an air cushion, and it quickly comes to rest or is "dead beat."

In Fig. 8 is the tripod form of Thompson reflecting galvanometer.

This instrument is made with interchangeable coils from 150 ohms to 5000 ohms in resistance. The sensibility for minute currents with the 5000 ohm coil is very great.

A four coil reflecting galvanometer of the Thompson pattern in Fig. 9 has a hinged door which opens, breaking the circuit through the coils.

This gives easy access to the suspension system. Spring contacts *C* seen at the lower part of the base make connection at *P* when the door is shut.

The magnetic system is suspended from a tube *B* equipped with an adjusting screw *K*.

A compensating or control magnet *M* is fitted in a tube and adjusted by the milled head *S*.

Moving Coil Galvanometers. One of the greatest drawbacks to the use of galvanometers built on the Thompson pattern is their sensitiveness to external influences. In large cities where trolley cars and buildings of iron frame construction abound, this type of instrument needs constant adjustment. Allowances and checks have to

enter into tests made with them so largely as to almost prohibit their use.

A class of galvanometers based upon the rotation of a coil carrying current in a magnetic field

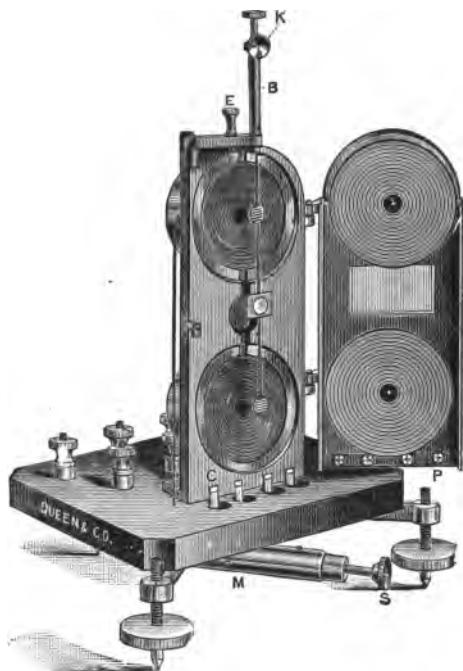


FIG. 9.

has been so perfected that they are being almost universally adopted.

The pioneer moving coil galvanometer is that invented by D'Arsonval.

The D'Arsonval Galvanometer. In Fig. 10 *H* is a horseshoe shaped magnet or series of magnets bolted together and to the base *B* which is supplied with leveling screws *LL*.

A core of soft iron *I* is held stationary, around

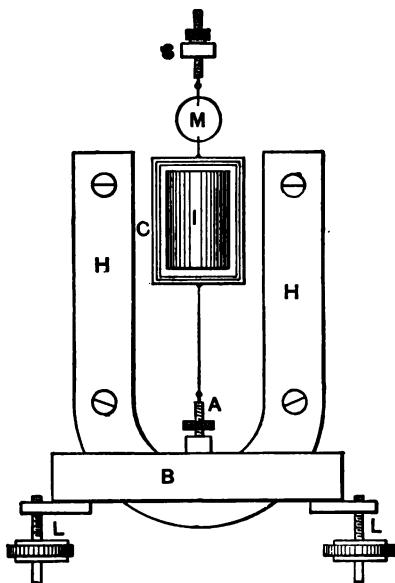


FIG. 10.

which turns a coil of fine insulated wire *C*. This coil is suspended between adjustable screws *SA* by fine wires in such manner that current entering at the screw *S* passes through the coil and out at *A*, or vice versa.

A tiny mirror M of circular form is also carried by the suspension wires and turns with the coil.

When current is applied to S and A , the coil tends to turn in the magnetic field formed by the magnet poles and strengthened by the soft iron core I .

The instrument is so constructed that the magnetic field is as nearly uniform as possible, and the deflections of M are proportional to the current flowing through C .

Horizontal Magnet. This type of D'Arsonval galvanometer is shown in elevation in Fig. 11.

In this galvanometer the coil and mirror are suspended in a tube by a phosphor bronze strip.

This form of construction permits the mirror and coil to swing but to return to zero without any distortion from twisting of the suspension or "set."

Interchangeable tubes are furnished with coils of varying resistances and sensibilities. The 20 ohm tube has a sensibility of 75 megohms (75 million ohms). The 4000 ohm tube has the extreme sensibility of 1750 megohms.

The deflections in these instruments are proportional to the current.

The moving system is either dead beat or ballistic.

The ballistic type is adapted for the measurement of currents of short duration. The current has ceased before the deflection is complete. But

the impetus received by the coil carries it to a distance and gives the possibility of measuring the current strength from the distance of swing.

A form of D'Arsonval galvanometer, Fig. 12, is adapted for attachment to a wall. The galvano-



FIG. 11.

meter is entirely enclosed in a cylindrical case, the mirror deflections being observed through a window in front.

A removable bracket carries a reading telescope and scale so proportioned that it gives the same

effect as a standard millimeter scale at a distance of one meter.

Gimbal suspension of the galvanometer itself causes it to be self-leveling.

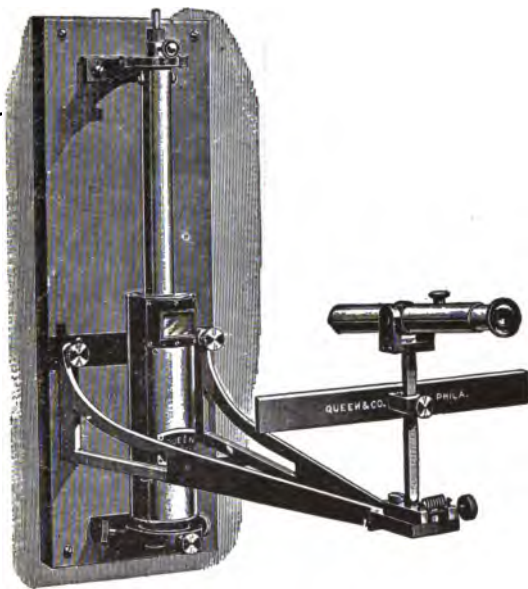


FIG. 12.

The scale deflections are directly proportional to the current causing them.

The sensibility is about 500 megohms per volt of battery used, a sensibility sufficient for all ordinary measurements.

The Thompson-Levering Type of D'Arsonval galvanometer is shown in Fig. 13. It is of high



FIG. 13.

sensibility, having generally a value of 1000 megohms, or higher if so desired.

A portable type of D'Arsonval galvanometer also made by Thompson-Levering, is shown in Fig. 14. The sensibility is sufficiently high for Wheatstone bridge measurements and also for many commercial direct deflection tests. With an e.m.f. of one volt and through a resistance of 750,000 ohms, the pointer will move one division on the scale, which



FIG. 14.

has 15 divisions on either side of the zero mark. A set of shunts having values of 1, .1 and .01 can also be furnished with this portable instrument, increasing its range of usefulness.

As has been before noted, instruments constructed on the D'Arsonval system read in direct proportion to the current passing, and therefore require no tables of conversions.

And they are not affected by external magnetic fields. An instrument similar to the above with 100 cells of battery would be capable of measuring insulations up to 100 megohms, and higher if fractions of a deflection be read.



FIG. 15.

Scale and Lamp. The oscillations of the galvanometer mirror are read in two ways. By watching a spot of light on a scale or observing the image of the scale reflected in the mirror by means of a telescope.

The scale is made in various forms but in principle resembles that in Fig. 15.

Choice of a Galvanometer. The galvanometer should be suited to the work for which it will be used. A galvanometer of high sensibility would be unnecessary and undesirable for use with a portable Wheatstone bridge, while a galvanometer with a sensibility sufficient for the bridge would be unsuited to the testing of high insulations.

For field work and general commercial testing the instrument must be strong and simple, yet of sufficient sensibility. For laboratory work the element of strength and simplicity may be subservient to the requirements for the utmost sensibility.

For marine work where there is the condition of unrest owing to the motion of the vessel, the marine galvanometer has straight and tightly stretched suspensions carefully balanced so that the inclinations due to these motions will not materially affect the readings. But this advantage is gained with a loss of sensibility owing to the greater resistance offered to torsion by the suspension.

Then again the matter of portability may or may not be important. The condition of portability may be met, as in the case of the galvanometer used with the portable testing sets, where the required sensibility is not high, or as in the portable cable testing outfits where the sensibility is necessarily of a high order.

The relative rapidity with which the system moves may also be an important point. The requisite may be either to measure currents which are but momentary, or those which continue for a long period and

fluctuations of which are to be watched. The swing of the mirror or needle may be periodic or aperiodic. It is periodic where the deflection swings from its initial direction back and through the zero or point of normal rest. It is aperiodic where it returns to the zero without passing it. The better class of D'Arsonval type voltmeters are examples of aperiodicity.

The galvanometer should be suited to the conditions under which it must work.

A galvanometer for use aboard ship should have a short, well-balanced suspension and as high a sensibility as is consistent with the proper suspension to render it unaffected by the motion of the vessel.

A portable galvanometer for cable tests ashore would merely have to be built to stand transportation. It would work on a motionless surface and therefore the suspension might be as fine as required.

The galvanometer for use in the laboratory for fine tests of insulation or high resistances might have the maximum of sensibility consistent with the conditions peculiar to the location. For example, a Thompson reflecting galvanometer is often made with a sensibility of 18,000 megohms and can be made with one as high as 100,000 megohms. But it would be difficult to insulate from the effect of trolley cars even at a distance, from magnetic disturbances of the atmosphere and from the presence of moving masses of iron, such as elevators.

The speed with which the moving system reaches its maximum or returns to zero is also of impor-

tance and must be considered with reference to the choice of the instrument. A ballistic galvanometer which has a slow period for testing the charge and discharge of condensers would be too slow for ordinary insulation tests.

A galvanometer intended to use with a Wheatstone bridge need not have a greater sensibility than one-half megohm, but should be quick. The latter being essentially a zero instrument, the actual deflections are of less value than the point of balance when the indicator remains at zero. It should have a light suspension capable of a rapid movement, but responding quickly to the dampening effect of the short-circuiting key.

CHAPTER III.

RHEOSTATS, KEYS AND SHUNTS.

Rheostats. The resistances used in testing-rheostats are mostly made of insulated wire having a high specific resistance, and but little liable to change under varying temperature conditions.

They are also wound non-inductively so that no currents due to inductance shall affect the galvanometer. This end is obtained by winding the wire so that the current shall flow half around the spool in one direction and half around it in the reverse direction.

The piece of wire may be either doubled upon itself and then wound on, or may be wound on in two equal lengths, the inside ends being soldered together. The outside ends are for attachment to the circuit.

It is always desirable that the binding posts or terminals of a resistance coil be made large enough that they do not offer any perceptible resistance to the circuit.

In the practical construction of rheostats for testing, they are adjustable in several ways. The

cutting in or out of the coils is effected by plugs or sliding switches. In the plug form of switch, Fig. 16, a brass plug tapering a trifle so as to fit tight, is inserted between two brass strips or discs *S*, to which the resistance coils are connected. Normally, when all the coils are cut out, these plugs remain in their places, short circuiting the coils. When it is desired

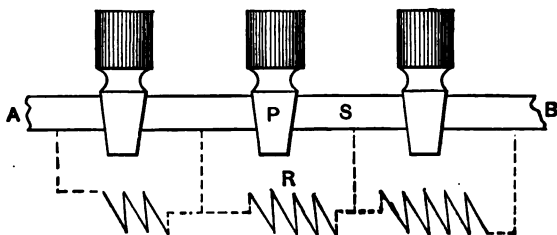


FIG. 16.

to add the resistance of one of the coils, the adjacent plug is removed, making the path of the circuit through the brass strips or discs and through the coil controlled by the plug which has been removed.

The chief advantage of this system is that it is simple, but it has a disadvantage. If the plugs are not properly tightened, the coils are not entirely cut out.

Resistance Coil Combinations. The resistance coils in a rheostat are arranged so that any value

within their range may be obtained by suitable combination. The values of the coils are chosen with the view to limit their number, saving space, decreasing the number of contacts, reducing expense and facilitating adjustments.

There are several plans upon which the coils may be arranged. In one plan the coils have the ratio of 1. 2. 3. and 4 (ohms). All values up to and including ten are obtained by combining the coils in series and adding the values.

EXAMPLE.— $1 + 2 + 3 + 4 = 10$. or $4 + 2 = 6$. and so on. Values of tens up to and including 100 are obtained from coils of 10. 20. 30. and 40 ohms, and other combinations as will suggest themselves.

Another plan employs coils having the resistance ratios of 1. 2. 2. and 5.

In some designs of rheostats and Wheatstone bridges, coils are cut in to increase resistance by removing plugs which normally short circuit the coils which they control. This method necessitates a plug and two fixed contacts for each resistance coil.

A system employing the minimum number of plugs and contacts is that termed the "decade" plan on account of the general arrangement of the coils in rows of tens or decades.

EXAMPLE.—There would be 10 coils each of 1 ohm, 10 coils each of 10 ohms, 10 coils each of 100

ohms and so on. Some systems use only 9 coils in the decade, but the 10 coil plan permits a greater facility in checking the rheostat. By using 10 coils of 1 ohm each in a decade, this total of 10 ohms could be checked against the 10 ohm coil in the next higher decade.

The usual layout of two decades is illustrated in

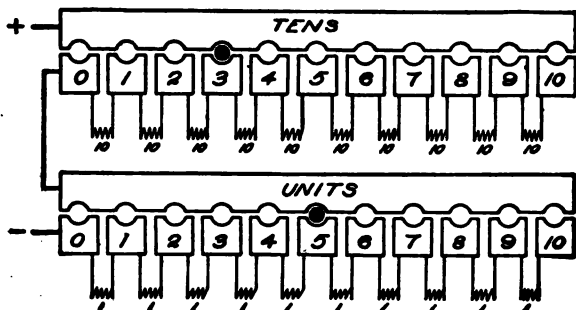


FIG. 16A.

Fig. 16A. The circuit runs through the bar to a plug, thence through the coil or coils connected in by the plug to the next bar and so on to the terminal post. In the diagram shown, two decades on the 10 coil plan have their plugs arranged to cut in 8 coils with a total resistance in series of 35 ohms.

If the plug is removed entirely from a decade, the series is broken and the rheostat is on open circuit.

If the plug in any decade is placed in the 0 hole, that decade is cut out entirely but the circuit is not broken.

Leeds and Northrup Arrangement. There are but four coils to the decade arranged as in Fig. 16b. Four coils having the values of 1, 3, 3, and 2 ohms respectively, are arranged in series. The terminals of the 1 ohm coil, the 2 ohm coil and the three points of junction between the coils are numbered as in the illustration (1) (2) (3) (4) (5). The path of the current will be from (1) through the coils and to (5). When (1) and (5) are short circuited, all the coils are also short circuited. By connecting points (2) and (5) the interposed resistance will be 1 ohm as the coils having the values of 3, 3, and 2 ohms are short circuited. On connecting points (1) and (3) the two coils in between will be short circuited and the ohmic value will be 5 ohms. A study of the diagram will show all the combinations.

Fig. 16c shows the actual connections to the discs, the ohmic value of the combination being shown by the figures at the right, while the connections to the various junction points are shown by the figures in the discs. The highest value of this particular decade, that of 9 ohms, is obtained by removing the plug entirely, or placing it between the last pair of discs which have no coil connections.

The majority of the testing sets are now made with sliding switches instead of plugs, which method presents many advantages, but is more expensive.

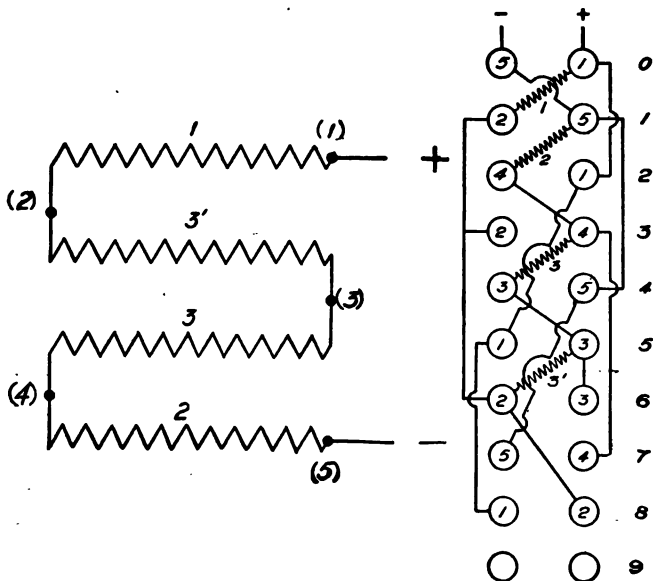


FIG. 16B.

FIG. 16C.

With the sliding switch system there are no plugs to be lost, get loose or introduce false values through dirt or through being improperly inserted. The correct way to insert a plug is to push it in the hole with a slight screwing motion, but the amount of pressure with which it is inserted varies with the person using the set. It is possible by careless handling to

spread the discs or loosen them, and it is rare for the average operator to clean the plugs or their seats. With sliding switches the proper pressure has been determined and set by the manufacturer, and no dirt errors or those due to improper insertion of plugs can arise. The sliding motion keeps the contacts clean and insures perfect contact, while the switch is far more convenient to handle than the plug style.

TABLE IIa

WEIGHTS, RESISTANCES AND BREAKING STRAINS OF
IRON AND STEEL TELEPHONE WIRE

American Steel & Wire Co.

Size B.W. G.	Approximate Weight in Pounds		Approximate Breaking Strain in Pounds			Resistance per Mile (International Ohms) at 68° F.		
	Per 1,000 Feet	Per Mile	Ex. B. B.	B. B.	Steel	Ex. B. B.	B. B.	Steel
4	153	811	2,028	2,271	2,433	5.80	6.91	8.01
6	112	590	1,475	1,652	1,770	7.97	9.49	11.02
8	74	390	975	1,092	1,170	12.05	14.36	16.71
9	60	314	785	879	942	14.97	17.84	20.70
10	49	258	645	722	774	18.22	21.71	25.29
11	39	206	515	577	618	22.82	27.19	31.55
12	32	170	425	476	510	27.65	32.94	38.23
14	19	99	247	277	297	47.48	56.56	65.66

Resistance Wires. The choice of a metal for resistances to be used for testing purposes must be

determined by several points. The metal must change in resistance as little as possible under varying temperatures.

It is also desirable that the metal used shall have a high specific resistance, that is, its resistance per unit of length shall be as great as possible in comparison with other metals. This is in order to keep the coil as small as possible, resistance coils with figures as high as 100,000 ohms being in common use.

German silver, although formerly much used, has been replaced by several compound metals.

The composition of some commonly used resistance wires is as follows: German silver, copper 60 parts, zinc 26 parts, nickel 14 parts. This alloy is also made with 30 per cent. nickel.

Platinum silver contains platinum 67 parts, silver 33 parts.

Platinoid is German silver 98 parts, tungsten 2 parts.

Manganin contains copper 84 parts, manganese 4 parts, nickel 12 parts.

The relative resistances of the above are in comparison with copper as 1, as follows: German silver 12, platinum silver 15, platinoid 20 and manganin 30.

As the proportions of these alloys is varied by different manufacturers, the figures given are but approximate. For example, if the platinum silver alloy were platinum 50 and silver 50 parts, its relative resistance would be about 20

as compared with copper, or more nearly that of platinoid.

Of the above alloys, manganin seems to be the most favored by manufacturers of high-class instruments, although platinoid is preferred by some.

Glass Slab Resistances. It often becomes necessary to use a high resistance for testing purposes when none is at hand. A ground glass slab covered with lead pencil lines will serve for temporary work.

To make one proceed as follows:

Take a slab of ground glass about 1 inch by 4 inches and drill a hole at each end large enough to receive the machine screws at the bases of two binding posts.

Rub a soft lead pencil over the slab around these holes covering the area of a circle of one inch or thereabouts. Blow off the dust and lay a few strips of tin-foil over the holes nearly covering the lead pencil circles. Slip a washer over one machine screw and also a few washers of tin-foil and insert the screw in the hole from the plain side of the glass. The screws should penetrate the tin-foil on top of the glass also. Then screw down the binding post.

The tin-foil below the slab acts as a cushion between the washer and the glass. That on top of the slab also acts as a cushion, but it ensures a better contact between the binding post and pencil circle.

A similar binding post is to be fitted at each end of the slab.

Now draw a few lines joining the pencil circles. Measure the resistance of this penciled path. To lower its resistance rub on more pencil lead between the circles. To raise its resistance, rub out some of the lead. Of course all dust should be blown off.

The object of the circles was to give a point of junction between the lines and binding posts.

Resistance slabs as above may be varnished or covered with insulating compound if desired. But pencil lead or plumbago being carbon is extremely erratic in the matter of maintaining a given resistance.

Galvanometer Shunts. As in most cases it is not desirable to permit the entire current used in a test to flow through the galvanometer, part of it is shunted or caused to pass around the latter.

A shunt bears a definite ratio to the resistance of the galvanometer, being generally adjustable to $\frac{1}{9}$, $\frac{1}{99}$, or $\frac{1}{999}$, of its resistance so that $\frac{1}{10}$, $\frac{1}{100}$, or $\frac{1}{1000}$ part of the current only passes into the galvanometer.

Fig. 17 shows the connections of the shunts *S* and galvanometer *G*.

The degree in which the shunt increases the

range of deflection of a galvanometer is termed its multiplying power.

If one-tenth of the current flowing went through the galvanometer and nine-tenths through the shunt, the current in the circuit would be actually ten times that through the galvanometer.

The current therefore in the galvanometer must be multiplied by the multiplying power of the shunt to show its true value in the circuit.

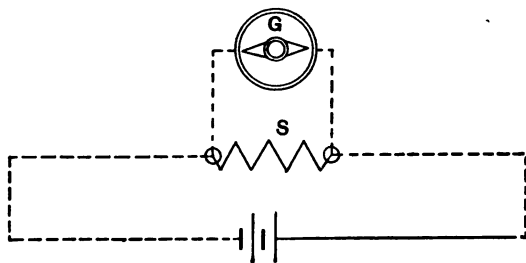


FIG. 17.

In order to find the resistance necessary in a shunt to be used with a certain galvanometer, the resistance of the latter is to be divided by the multiplying power desired less one.

As an example, let a shunt be needed for a galvanometer of 2000 ohms resistance where only one-fifth the total current is to pass through the galvanometer. This would equal a multiplying power of 5; then $\frac{2000}{5-1} = 500$ ohms.

Formula. Let G be resistance of galvanometer; n be multiplying power of shunt; S resistance of shunt; then $S = \frac{G}{n-1}$.

To find the multiplying power of a shunt of given resistance add its resistance to that of the galvanometer and divide the answer by the resistance of the shunt.

For example, galvanometer is 10,000 ohms, shunt 1000 ohms, $10,000 + 1000 = 11,000$, divided by 1000 = 11, the multiplying power of the shunt.

Formula. Let resistance of galvanometer be G and shunt S ; then $\frac{G+S}{S} =$ multiplying power.

Shunts should be connected to the galvanometer by wires of ample size, no undue resistance should be introduced by the connecting wires.

It is a good plan always to use the shunt of greatest multiplying power at first and reduce as occasion requires. Otherwise a heavy current in the galvanometer might injure suspension.

Condensers. The modern condenser is but a handy form of the Leyden jar; it consists of leaves of tin-foil alternating with insulated paper, mica, glass, etc., insulation between the tin-foil layers being of prime importance.

The layers are built up as follows, Fig. 18: First a sheet of insulation A (or dielectric), then a sheet of tin-foil B projecting at one end; another sheet

of insulation *C* and the next sheet of tin-foil *D* projecting at the other end.

Sheet after sheet is built up, Fig. 19, until the desired number is obtained, the whole mass then being subjected to immersion in paraffin or other insulator and kept under heavy pressure until set.

The portions of tin-foil projecting from one end being pressed together form one connection, the ends at the other being the second connection.

This results in every other sheet of tin-foil being connected, in fact the condenser can be considered

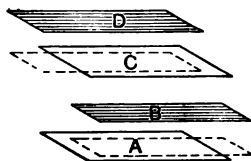


FIG. 18.



FIG. 19.

as being made up of two large sheets of tin-foil highly insulated from each other.

Condensers are made up with many different dielectrics, mica being one of the best for testing purposes. Tin-foil is most always used for the metal plates.

Adjustable condensers permit of their capacities being varied by means of plugs which cut in or out of multiple, portions of the foil sheets. In Fig. 20 is the diagram of an adjustable condenser.

CC are the tin-foil sheets, *BB* brass strips which carry the terminals or binding posts.

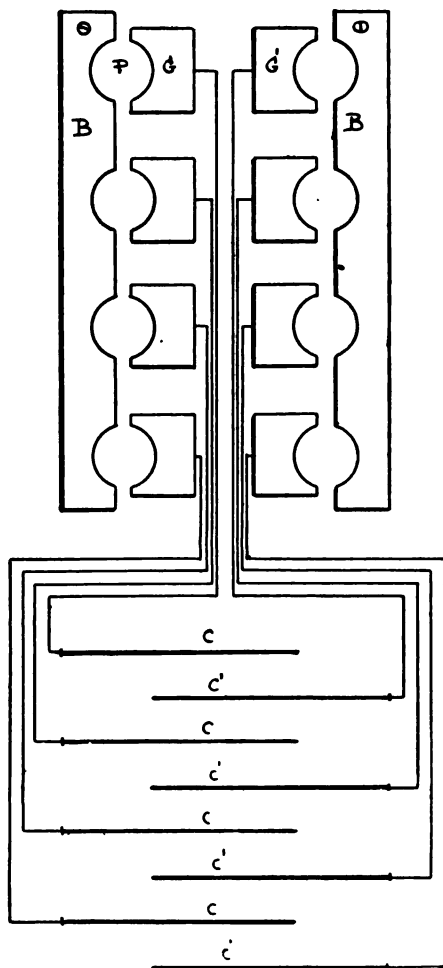


FIG. 20.

The tin-foil sheets are connected to brass blocks GG in such proportion of the total number in the condenser as desired.

By inserting taper brass plugs in holes between the brass strips and blocks as at P various combinations may be made.

A standard adjustable condenser is arranged so that the pairs of foil sheets may be connected in multiple as above, or in series.

Keys. In electrical testing various kinds of switches or keys are used; for opening or closing a circuit, for short circuiting, for ground connection and for reversing the direction of current flow.

The insulated parts are made of hard rubber, contacts being equipped with platinum.

In order to ensure the greatest possible insulation, the hard rubber pillars are often encircled by grooves.

This decreases the leakage from dampness as it increases the distance between the conductor supported and the table, measuring up the surface of the pillar.

Reversing Key. A reversing key, Fig. 21, is used to reverse the direction of current flowing into the galvanometer. Two spring brass levers AB provided with hard rubber knobs are connected by means of binding posts to the galvanometer G . A brass strip O carrying platinum con-

tacts bridges the levers and a similar brass strip *U* passes under them.

The battery is connected as shown, one terminal to the lower strip and the other terminal to the upper strip.

When both *A B* are up they make contact with *O*, and form a short circuit to the galvanometer *G*, and likewise when both levers are held down on strip *U*.

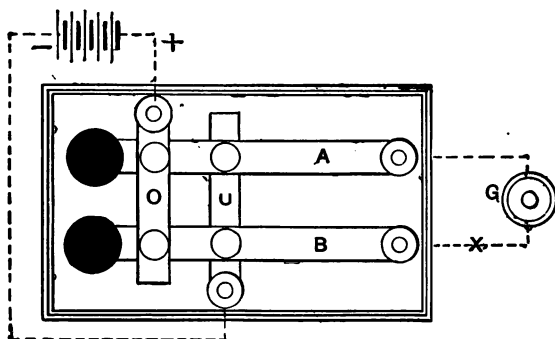


FIG. 21.

But if one lever *A*, for example, be pressed down the negative terminal of the battery is connected to *A* and the positive terminal to *B* and thence to the galvanometer.

On the other hand, if *B* only is pressed down the current to *G* is reversed because *B* is now connected to the negative terminal of the battery and *A* to the positive.

This key should not be so connected that the

platinum contacts form part of a circuit, the resistance of which is being measured. The contact resistance will vary as it is not possible to hold a spring key down with an unvarying pressure.

It will be seen that under no circumstances should the battery be connected to the terminals



FIG. 22.

at the end of *A* and *B*, a short circuit of the battery would result.

The Rymer Jones reversing key shown in Fig. 22 makes a rubbing or wiping contact. The levers moved by the hard rubber handles are equipped with large contacts of platinum.

The contacts on which these levers work are also faced with platinum.

A rubbing contact of this nature ensures but a nominal resistance. In fact, it can be neglected in ordinary operations.

Keys of this style of construction are always preferable to those with striking contacts.

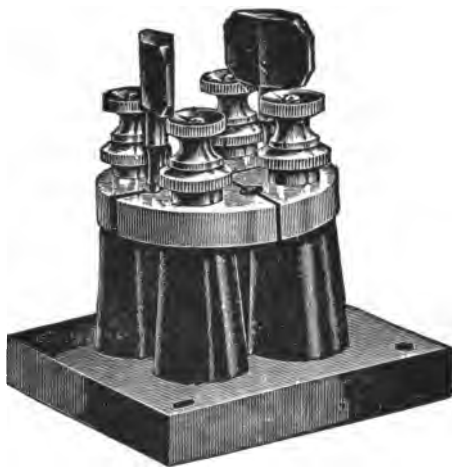


FIG. 23.

Commutator. A form of reversing switch or commutator is shown in Fig. 23. By changing the plugs into different holes, circuits connected to the commutator may be reversed, put to earth, or short circuited.

The brass pieces are mounted on hard rubber pillars to increase the insulating distance from each other and from the base.

Discharge Key. A discharge key of the Kempe pattern, Fig. 24, has two triggers controlled by buttons. One button is marked "insulate," the other "discharge."

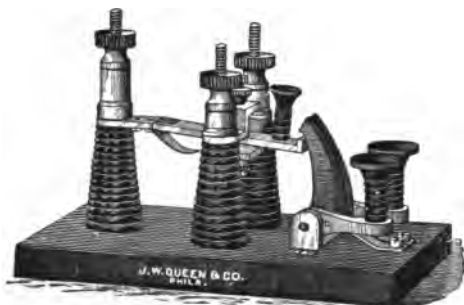


FIG. 24.

There are three binding posts, one connected to a movable lever operated by a button, one to a contact underneath this lever, and one to a contact over this lever.

When the lever is up, it presses against the upper contact completing a circuit connected to the lever binding post and top contact binding post.

When the lever is down, it presses against the lower contact; the circuit is now through the lever and lower contacts and their respective binding posts.

But in certain tests it becomes necessary to leave open the circuit connected to the lever.

The key provides for this in its second operation as follows:

Depress lever all the way down and it locks under a trigger controlled by the "insulate" button. Lower contact and lever are together. Press "insulate" button and lever flies up. But not all the way, it is caught and held in mid-air by the trigger attached to the "discharge" button. The circuit attached to the lever is thus broken.

Third, press "discharge" button and lever being released rises and rests against top contact.

The Standard Cell. A standard cell is a battery cell used in testing which maintains a steady e.m.f. of known value. It is used principally for comparison of other cells and to check voltmeters.

A standard cell is not required to give a large current, in actual work the current flow is kept as low as possible.

A standard cell has a thermometer included in the case as a small correction must be made for changes of temperature.

For tests not requiring extreme accuracy, the Daniel or copper sulphate cell may be used. Its e.m.f. is so near one volt as to be usable as such. A number of Daniel cells carefully set up and connected in series may be measured as to e.m.f. and the reading used for tests.

But for accurate work the Clark or the Weston cell is necessary.

The Clark cell has a positive element of mercury and a negative element of zinc sulphate and mercurous sulphate. Platinum electrodes form the connections in these elements. The e.m.f. at 15° C. is 1.434 volts.



FIG. 25.

The Weston cell, Fig. 25, has a positive element of mercury also, but the negative element is cadmium amalgam in a saturated solution of cadmium sulphate. Its e.m.f. is 1.019 volts.

The actual construction of standard cells may be found in any of the large works on testing.

CHAPTER IV.

VOLTMETERS AND AMMETERS

The Voltmeter. A voltmeter is merely a galvanometer of high resistance connected across two conductors of opposite polarity.

As resistance of the voltmeter is made extremely high in comparison with that of the conductors, but a minute current flows through it.

Moreover, as this resistance is fixed, the only way to vary the amount of current flow is to vary the e.m.f. An increased e.m.f. will increase the current flow and likewise a decrease of one will produce a decrease of the other.

These current variations producing corresponding deflections of the needle, the deflections will show actually the changes of e.m.f.

In the laboratory the scale divisions are calibrated, that is, given a value in volts, either by using e.m.fs. of known value, or by means of a standard voltmeter.

Voltmeters are connected across the circuit as in Fig. 26.

The Series Ammeter. These are divided into two classes, the series ammeter and the shunt ammeter.

In the former the *entire current* to be measured passes through the coils of the instrument, Fig. 27, and its changes of value directly affect the needle as in the simple galvanometer. Changes of e.m.f. will therefore not affect an ammeter directly but only in that they vary the current flowing in the circuit in which the ammeter is connected.

Series ammeters are unwieldy for large currents as the coils must be large enough to carry all the current of the circuit. They are therefore almost entirely displaced by the shunt ammeters.

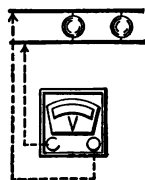


FIG. 28.

The Shunt Ammeter. The shunt ammeter is really a voltmeter calibrated to read in amperes and controlled by the changes of e.m.f. in a portion of the circuit.

The current in a circuit of given resistance is controlled by the e.m.f. at its terminals.

Ohms law teaches that the current or I equals the e.m.f. or E divided by the resistance or R , as a formula $I = \frac{E}{R}$.

If then there is an increased current flow in a circuit of constant resistance it must be due to an increased e.m.f.

By connecting the voltmeter across a portion of the circuit, the variations of e.m.f. causing the flow of current will be indicated on it.

In the shunt ammeter, Fig. 28, a German silver shunt S or one made of a special alloy is connected in series with the main circuit $A B$.

A low reading voltmeter A , actually a millivoltmeter, is connected to the opposite ends of the shunt.

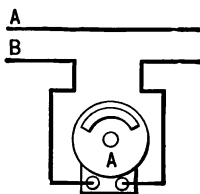


FIG. 27

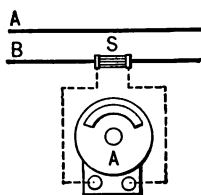


FIG. 28.

Variations of e.m.f. across the terminals of S will affect A , which has been already calibrated in amperes by comparison with a standard, or by calculation.

The Weston Voltmeter. The principle of the Weston voltmeter or ammeter for direct current is that of the D'Arsonval galvanometer. The moving coil is mounted on pivots between jeweled centres instead of being suspended from wires or threads. This mounting permits of portability and compactness and adds strength.

The construction of the moving system and magnet pole pieces is shown in Fig. 29.

The portable type of Weston voltmeter and ammeter is shown in Fig. 30.



FIG. 29.

In the portable instruments the data of resistance, etc., will be found printed on the inside of the lid. Care should be taken not to interchange

the case lids, as the data is of value in various applications of the instruments.

The laboratory type, Fig. 31, has a very large scale of peculiar design, Fig. 32, which admits of readings being made to a fraction of one degree of deflection.



FIG. 30.

The resistance of the laboratory type is above 1000 ohms per volt and readings of e.m.f. may be made with great accuracy.

Sensibility. In the Weston direct current instruments of the portable or the Type *B* form, a resistance of about 100 ohms per volt is added.

In the switchboard voltmeters the resistance is usually about 65 ohms per volt.

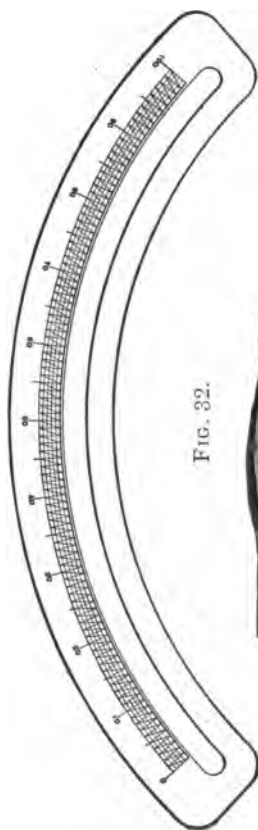


FIG. 32.



FIG. 31.

In case the exact resistance is not marked, it can be obtained from the makers by quoting the serial number of the particular voltmeter.

As the figures are needed in testing work, they should always be recorded.

The sensibility of a Weston instrument type B or of the portable form is about 10,000 ohms. One scale division of deflection represents one volt through 10,000 ohms if 100 divisions equal one volt through 100 ohms.

The current used will equal the voltage indicated divided by the resistance of the instrument.

EXAMPLE.—In a 150-volt voltmeter of 15,000 ohms, $\frac{150}{15000} = \frac{1}{100}$ ampere for full scale deflection.

One scale division here representing one volt would equal $\frac{1}{15000}$ of an ampere.

The moving element of the Roller Smith D'Arsonval type direct current voltmeters and ammeters is shown in Fig. 33. The reference of the letters is as follows:

A two-piece cylindrical shaft *D D* is equipped with sapphire bearings, which are spun into place at the two extremities. By means of a collar *C*, the distance between the jewels may be varied by lengthening or shortening the shaft. The spiral springs *F F* are so attached as to act as cushions to the shaft, nullifying the effect of a shock or blow to which

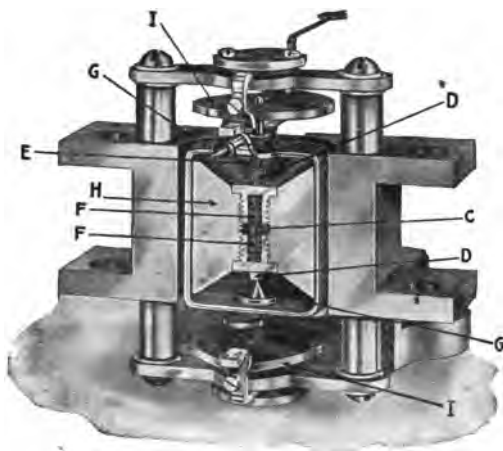


FIG. 33.



FIG. 34.

most portable instruments are subjected during ordinary handling. The coil *C* bears the pointer, *E* being one of the pivots moving in the sapphire bearings. The moving system is accurately balanced by means of counter weights.

A double scale portable instrument is shown in Fig. 34. The voltmeters have an average resistance of 100 ohms per volt, but can be made higher if desired. The ammeters up to 300 ampere capacity have self-contained shunts, all shunts being constructed of a metal with a zero temperature coefficient and with a uniform drop of 50 millivolts at full load. The latter quality enables instruments having separate shunts to be used interchangeably with other shunts.

A special combined voltmeter and ammeter made by the Roller Smith Company for testing in railway signal work is shown in Fig. 35. This instrument is of course applicable to all testing on a direct current of the same range. The mechanism is of the D'Arsonval type and is similar to that shown on page 46. The instrument has the following ranges of measurement:

One scale reads from 0 to 3 volts in divisions having a value of .02 volt. A second scale reads from 0 to 30 volts in divisions of .2 volt each. The current readings are three, from 0 to 15 amperes in .1 ampere divisions. From 0 to 1.5 amperes in .01 ampere divisions, and from 0 to 150 milliamperes in divisions of one milliampere.

As will be seen in the figure, the binding posts are generously large, avoiding contact errors and furnishing means to use the several ranges without possessing the unreliability of a switch. The bind-



FIG. 35.

ing posts are, moreover, so constructed as to avoid any chance of twisting or becoming loose, which is a valuable feature.

By means of a conveniently placed screw the

pointer may be brought to zero should it become bent. This device works in such a manner as to eliminate all chance of turning the pointer too far and injuring it. Other instruments of this class are furnished with ranges suitable to any desired employment.

The mechanism of a type D P voltmeter made by



FIG. 36.

the General Electric Company is shown in Fig. 36. These type D P instruments are portable, being designed for laboratory or general testing purposes. The line comprises ammeters, voltmeters, milliammeters and millivoltmeters. The D'Arsonval principle is employed, the mechanism being placed in a drawn steel case. This steel case is set in the centre

of a mahogany ring fastened in the centre of a square mahogany carrying case.

The ammeters are built with self-contained shunts up to 30 amperes capacity. For high ranges up to 2000 amperes, a millivoltmeter is furnished with its scale directly calibrated in amperes and is used with a portable shunt.



FIG. 37.

The milli-voltmeters are adjusted for full scale deflection on 200 millivolts, all shunts and instruments being interchangeable.

A combination portable volt-ammeter for general testing purposes is that made by the Keystone Electrical Instrument Company and shown in Fig. 37. The instrument comprises two current ranges and

two potential ranges. The current ranges are up to and include 50 amperes, while the potential ranges include 150 volts. All resistances and shunts are self-contained.

Connections to the instrument are provided for by leads equipped with tapered plugs which fit into receptacles in the upper end of the case. A contact key on the front of the instrument cuts it in or out.

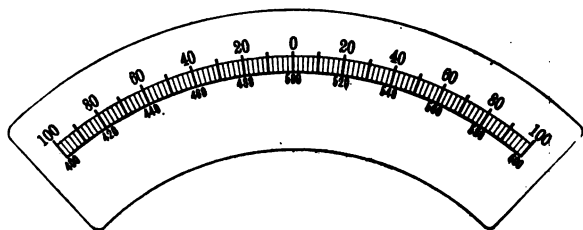


FIG. 38.

The D'Arsonval principle is employed in the movement which is dead beat and with scale divisions of even value.

A very excellent series of D. C. testing instruments is that of the Westinghouse type PL meters, which comprises voltmeters, ammeters, milli-voltmeters and combined volt-ammeters. They are direct-reading and dead-beat, also are calibrated to within 0.2 per cent. of absolute accuracy and readable up to 0.1 per cent. of the full scale.

The complete movement is attached to a brass

casting which makes a dust-proof joint with the case. The single air gap construction allows the moving coil to be removed if desired without also removing the pole pieces. There is thus no disturbance of the magnetic circuit, a great advantage, as

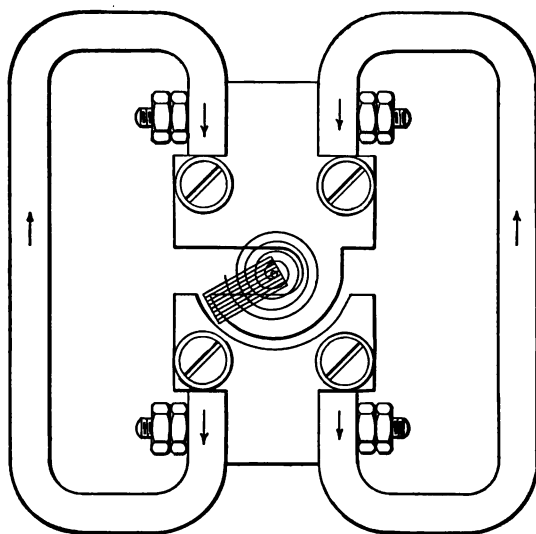


FIG. 38A.

the strength of the magnetic field is not altered and the instrument maintains its highest accuracy.

The design is of the permanent magnet and moving coil D'Arsonval type. The design of the magnetic field is peculiar consisting of two magnets, but

only a single air gap in which the coil swings. This and other details of design are clearly brought out in Fig. 38a.

The eccentrically pivoted coil so counterbalances the pointer that the counter weight is reduced to the minimum, this reducing the wear on the pivot jewels.

It is often impossible to set a testing instrument sufficiently far from an external field as to be out of its influence. In the present instrument the movement and pole pieces are so completely surrounded by the permanent magnets as to be practically independent of external influences.

The standard voltmeters have a resistance of 100 ohms per volt, with manganin resistances. Full scale deflections are obtained on the milli-voltmeters with 100 milli-volts applied at the terminals. For work requiring high sensibility, values of 250 and 500 milli-volts for full scale deflections can also be supplied.

Shunts for the ammeter class are made of manganin and have 100 milli-volts drop at full load.

Potential Indicators. These instruments are so constructed as to give large indications for slight changes of current or e.m.f.

In order to keep the scale within limits, the pointer does not move until the e.m.f. (for example) is near the average. In Fig. 38 is an illustrative scale for an e.m.f. of 500 volts.

The index or pointer commences to move as soon as the e.m.f. increases above 400, reading e.m.f. on lower scale.

The upper scale with central zero shows the number of volts above or below 500 as it will be seen that adding the upper left scale reading and that on the lower left scale directly underneath it, the total will be 500. And subtracting the upper reading from the lower reading on the right scale also gives 500.

Differential Voltmeters. Differential voltmeters have a central zero and are used in connection with two independent sources of e.m.f.

In the case of two generators which are to be run in parallel, the generators are adjusted until the needle stands at zero, when the e.m.fs. of both are equal. At other times the needle points towards the side of higher e.m.f.

When the two generators are being operated the readings on either side must not be taken to represent the e.m.f. of that particular generator. The readings indicate the difference between the generators.

Multipliers. A multiplier is an extra resistance used in series with a voltmeter to increase its range of measurement. The multiplier must be suitable in resistance for the voltmeter with which it is used, this resistance being calculated as follows:

Rule. Divide the desired full scale deflection value of the voltmeter by the present value, in volts, then multiply the resistance of the voltmeter by this number and subtract from the result the present resistance of the voltmeter.



FIG. 89.

EXAMPLE. Given a voltmeter with a full scale deflection value of 15 volts and a resistance of 1500 ohms. The desired new range is from 0 to 150 volts.

This maximum of 150 is divided by 15, giving 10 as its dividend. The present resistance of the voltmeter being 1500 ohms, this latter resistance is multiplied by 10, which equals 15,000, or the total resistance of the whole combination. As, however, there

is already a resistance of 1500 ohms in the instrument, this is to be subtracted. Therefore, 15,000 minus 1500 equals 13,500 ohms, the value of the multiplier.

A secondary resistance of the foregoing value would be called a multiplier by 10, the manufacturers being able to supply multipliers in almost any range desired.

The value of having a good instrument in the first case is enhanced by this multiplier system, which immensely increases the value to its owner. Multipliers are portable and readily attached to or detached from the voltmeter with which they are used.

Millivolt and Milliampere Instruments. The prefix milli means one thousandth, a millivolt is an e.m.f. of one thousandth of a volt, a milampere or milliampere, one thousandth of an ampere.

Instruments of which the total scale reading is about one volt or one ampere, and those in which each scale division represents one thousandth of a unit, are millivoltmeters or milammeters respectively.*

Such instruments are of the utmost service in measuring low resistances, temperature changes, etc., and their use will be described in the chapter on testing.

*Sometimes written Milli-Voltmeter and Mil-Ammeter.

Ammeter Shunts. The ammeter shunts used with Weston switchboard instruments are shown in Fig. 40.

The shunt itself is made of one or more sheets of alloy resembling german silver in appearance. The ends of these sheets are fastened into brass or copper blocks.



FIG. 40.

Large clamping screws are provided for attachment of the shunt to the bus-bar, which latter are inserted in the large slots at each end of the shunt. The small screws are for the instrument cords.

The Whitney shunt, Fig. 41, is made of a solid block of metal slotted as shown so as to give the resistance of a long bar.

In all cases the connections to shunts must be

as perfect as possible. Poor contact causes heating if at the bus-bar connections, and thereby incorrect indicating at the instrument terminals.

The shunt cords must also be properly clamped or incorrect readings will result.

For the same reason the cords must on no account be tampered with. If too long loop them up and bind the looped part if desired with tape. A



FIG. 41.

shortened cord would increase reading of instruments as it lowers the resistance in series with the ammeter, while a poor connection would decrease readings on account of its resistance.

The importance of these precautions will be thoroughly appreciated when the principle of the shunt ammeter is considered.

Too little attention is often given to these points by those in charge of a plant, an instrument cannot give correct readings if improperly connected.

Shunts for portable instruments of low range are included in the base of the instrument.

Calculation of the Shunt. The actual calculation of a shunt for an ammeter is similar to that before given in the calculation of a galvanometer shunt. The formula is $\frac{A}{n-1}$, where A is the resistance of the ammeter and n is the constant of the shunt required.

The constant of the shunt is found as follows:

Rule. Divide the desired range of the ammeter by the present range.

EXAMPLE to find the constant. Let the present range be 15 amperes and the desired range be 150 amperes. Then the constant will be $\frac{150}{15} = 10$. If the ammeter resistance is .09 ohm, in this case the value of the new shunt will be $\frac{.09}{10-1}$, or .01 ohm.

Practical Method. It is often desired to measure a heavier current than the ammeter at hand is adapted for. In this case a shunt is to be used. Connect the ammeter so as to get a large deflection. Then with a shunt made of resistance wire or whatever is available reduce the deflection until it reads in proportion to the desired new value of the ammeter.

EXAMPLE. Suppose the ammeter available reads to 15 amperes total scale and it is desired to give it a value of 150 amperes. Keep a steady current

passing through the ammeter for instance of 10 amperes. Connect the new shunt across the terminals and adjust it until the reading on the ammeter is 1 ampere. The ammeter with this new shunt will now give readings which are to be multiplied by 10 to obtain their true value.

Hot Wire Instruments. Voltmeters and ammeters indicating from the expansion and contraction of wires carrying current were first represented by the Cardew types.

A fine wire was run over a pulley carrying an index, one end of the wire being rigidly attached and the other end held taut by a spring. When current passed along this wire, the latter expanded, the slack was taken up by the spring and the pulley rotated. Upon cessation of the current, the wire cooled and contracted, the pulley reversed and its pointer returned to zero.

Instruments made on this principle could be used on either direct or alternating currents.

But they possessed the disadvantages of burning out on accidental overloads, of the wire becoming stretched or otherwise distorted, of consuming a larger current than the other types which would render them less accurate, of requiring a long case and an undesirable position of that case and many minor defects.

The Whitney Hot Wire Instruments. The Whit-

ney type of hot wire instruments represents the practical solution of the before-mentioned defects.

The mechanism is shown in diagram in Fig. 42.

A wire *AB* of high resistance, low temperature coefficient and non-oxidizable metal is secured at

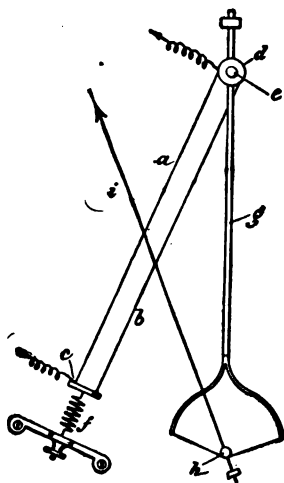


FIG. 42.

one end to a plate *C*. It runs over a pulley *D* mounted on a shaft *E*, and its free end is attached to the other end of *C*, but is insulated therefrom. A spring *F* keeps the wire taut and takes up any slack imparted to the wire by the passage of current.

Current only flows in the portion of wire *A* be-

tween *C* and the pulley *D* where the coiled wire connection is shown.

When *A* is heated by current flow it expands, *F* taking up the slack, pulls *A* round *D*, and rotates *D* and *E*.



FIG. 43.

If a pointer were directly attached to shaft *E* it would indicate on a scale the movement of *D* and *E*.

But a magnifying device is employed. *G* is a forked rod rigidly attached to *E*. At its lower end a silk fiber fastened between the forks passes around a pulley *H* which carries the pointer.

The rotating of E therefore moves G and the silk fiber rotates H and the pointer, thus magnifying the movement of G .

Any temperature variations will affect AB equally, it is therefore self-compensating.

The tension of F being slight, wire AB is not under a heavy strain, it more nearly returns to its original length after current ceases.



FIG. 44.

The instrument, however, can be adjusted to zero very readily if it should become necessary.

Two complete types of these instruments are shown in Figs. 43 and 44.

Other Types of Instruments. Voltmeters and ammeters are constructed on principles other than

that of the D'Arsonval galvanometer, or the stretching of a hot wire.

It is unquestioned that the D'Arsonval system is the only correct one for instruments to be used *on direct current circuits only*. Owing to patent rights it has been necessary for manufacturers to seek other methods of construction, with the result that while no new principles are used, those less suitable than the D'Arsonval have been perfected with such ingenuity as to make most excellent substitutes.

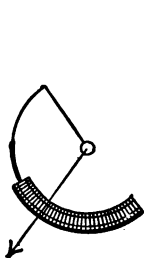


FIG. 45.

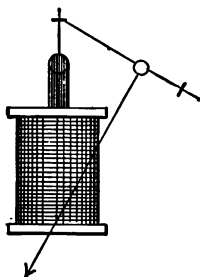


FIG. 46.

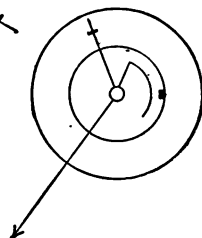


FIG. 47.

It is also unquestioned that the D'Arsonval system will be the only one that will survive. That is, unless the electrostatic instruments are much improved and made practicable for measurements of low potentials. This refers more to laboratory and portable instruments requiring accuracy than to those designed for switchboard work.

The solenoid type of Fig. 45 has a pivoted soft iron armature curved to enter a solenoid. Current

being passed through the wire of the solenoid causes the armature to be attracted more or less against a restraining force. The latter may be gravity or springs.

A pointer attached to the armature indicates on a dial.

A simpler solenoidal instrument is shown in Fig. 46, but acts in the same manner.

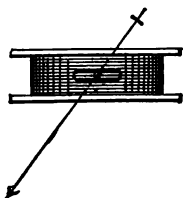


FIG. 48.

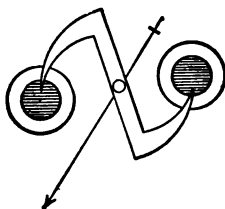


FIG. 49.

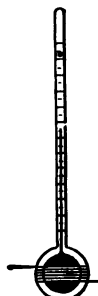


FIG. 50.

A type of magnetic vane instrument, Fig. 47, depends upon the repulsion between two pieces of iron in the same field. One is stationary inside a coil of wire, the other is movable on an axis.

As both lie in the coil the same way they both have *N* poles or *S* poles at the same end. As similar magnetic poles repel each other the stronger these iron pieces become magnetized the farther apart they move.

The galvanometer principle, Fig. 48, is used

where a short magnet is pivoted inside a coil large enough to give a fairly uniform field.

In Fig. 49 the attraction is between two curved armatures and the two poles of an electro magnet.



FIG. 51.

This is purely magnetic attraction and is counter-balanced by the force of gravity, a spring, or an adjustable weight.

The heating effects of current are used in Fig. 50, where a coil of wire surrounds a thermometer bulb. By the rise of the mercury the heat and thereby the current flow is measured.

Potential Indicator of the General Electric Company. An inexpensive form of ammeter and voltmeter is that made by the General Electric Company and illustrated in Fig. 51. Advantage is taken of the characteristic movements of an armature in a solenoid as before described. The armature as shown in the illustration is controlled by gravity, adjustment being obtained by weights on the small arm at the right hand side of the instrument. The ammeter is intended for series connection in the circuit in which it is used.

Roller Smith Alternating Current Voltmeter. The mechanism of the Roller Smith alternating current voltmeter as shown in Fig. 52 is of the electro dynamometer type, giving direct reading deflections. The mechanism comprises two fixed coils within which is mounted the rotating coil which carries the pointer.

The moving coil is set at an angle, so that lugs on the coil frame can be secured to the central shaft, which is claimed to be a more secure method than that of using pivots on plates shellacked to the coil. The moving system is made dead beat by the use of aluminum vanes which are attached to the shaft and damp the movements of the mechanism by their resistance to the air.

There being no iron in these instruments, they are free from hysteresis errors and may therefore be used on circuits of both direct or alternating current.

On direct current circuits two readings should be taken with the current reversed and the mean of the two readings accepted.

NOTE. With regard to the use of all instruments made primarily for alternating current, it is far bet-

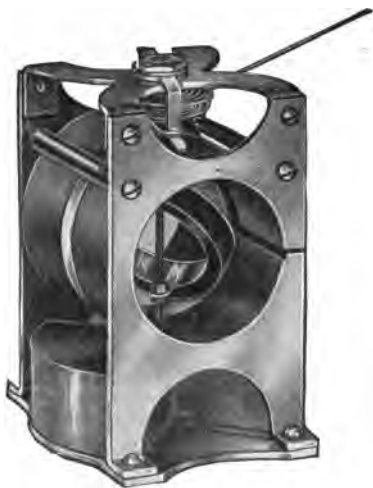


FIG. 52.

ter not to use them on direct current work unless no direct current instrument is at hand. There is no instrument on the market that will give good results on both circuits for close testing.

General Electric Portable Instrument. A port-instrument, either voltmeter, ammeter or watt-

meter, is the Type P3, shown in Fig. 53. The voltmeters are constructed on the direct reading dynamometer principle, and the ammeters have the Thomson inclined coil movement. These instruments may be used in circuits of alternating current or direct current. In the case of the ammeters two readings should be taken, reversing the leads, and the mean deflection calculated between the two readings.

Thomson Inclined Coil Meter. The Thomson inclined coil ammeter is adapted for use on circuits of alternating current. (Figs. 54, 55 and 56.)

The instrument is constructed on the magnetic vane principle, in which an iron vane or wing strives to turn itself parallel with the axis of a coil carrying the current to be measured.

The name comes from the peculiar position of the coil. Instead of lying flat on the base its axis is inclined.

A soft iron vane is mounted obliquely on a shaft which latter is held in a vertical position and controlled by springs.

The shaft also carries a pointer moving over a graduated dial.

When current is sent through the coil the vane turns against the springs so as to set itself parallel with the axis of the coil. The inclined position of the latter and the peculiar shape of the vane give a more evenly divided scale than is obtained in most other instruments of this class.



VOLTMETER, TYPE P—INTERIOR

FIG. 54.



AMMETER, TYPE P—INTERIOR

FIG. 55.

The voltmeter has a similarly placed stationary coil, but in place of the iron vane is equipped with a moving coil in series with the other coil.

The pointer is carried by the moving coil system which is controlled by springs. This type of instru-

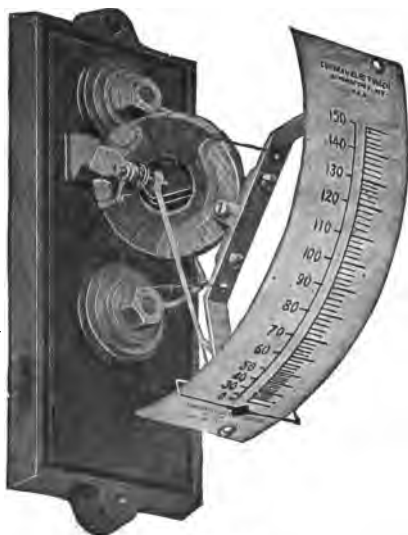


FIG. 56.

ment can also be used on direct current circuits where extreme accuracy is not imperative.

The Thomson Astatic Voltmeter. The mechanism is illustrated in Fig. 57, the same principle being used in the astatic ammeter which, however, is a shunt instrument.

The pointer shaft carries a coil through which passes the current to be measured, and also two pieces of magnetic metal.

The field magnets are electromagnetic, not permanent, as in the D'Arsonval instruments.

An astatically arranged magnetic field is mounted



FIG. 57.

perpendicularly to the shaft and restrains the pieces mounted on the shaft.

Deflections of the coil due to current flowing through it are opposed by the magnetic field acting on the pieces of metal which are carried by the same shaft. There are thus no restraining springs, current to the moving coil being conveyed by torsionless

spirals of silver wire. The illustration shows the illuminated dial voltmeter one of the miniature lamps for lighting the dial being visible. The resistance spools are also shown, two on each side of the electromagnet coils.

The dead beat or damping effect is produced by an aluminum disc which moves in the field of an electromagnet.

Thomson astatic instruments can be provided with polarity indicators, a red disc showing in the scale card when polarity is reversed.

The Wattmeter. The most efficient forms of wattmeters are constructed on the electro-dynamo-meter principle.

If two coils of wire are constructed, one stationary and the other free to turn inside it, the movable one will tend to turn until its coils are parallel with the stationary one when current is applied to both.

In the wattmeter, the stationary coil is of heavy wire and the main current flows through it. The movable coil is of fine wire in shunt across the circuit.

Changes of current strength in the main circuit thus affect one coil, and changes of e.m.f. the other.

As the watt is the product of the e.m.f. multiplied by the current, the index attached to the movable coil is influenced by the watts in the circuit.

This type of instrument can be used on direct or alternating current lines.

Queen Instruments. The Queen Wirt voltmeters and ammeters can be used on circuits of either direct or alternating current.

They are electromagnetic, the principle being that in which an iron armature tends to move into the strongest part of a magnetic field.

A tube of soft iron lies parallel with the axis of a coil of wire through which current passes. Being mounted away from the true centre of the coil it moves towards that centre where the field is strongest.

The electromagnetic principle does not allow of absolutely even scale divisions, but those in the middle of the scale are fairly uniform.

Although for use on both circuits, the scales must be graduated for either one or the other.

For alternating current the divisions are most uniform.

In the Queen portable form, an upper scale can be graduated for direct current and a lower scale for alternating current.

Electro-Static Instruments. A type of instrument much used on circuits of extremely high e.m.f. is made on the principle that two oppositely charged plates will attract each other.

A movable plate of aluminum is suspended between fixed plates of similar metal. The movable plate carries an index and is controlled either by gravity or springs, generally the former.

The fixed plates are connected to **one side** of

the circuit and the movable plate to the other side.

The instrument in fact is a form of condenser.

When a source of high e.m.f. is connected as above to the instrument the mutual attraction between the two series of plates causes the movable plate to swing and its index to indicate on a scale.

As the attractive force depends upon the *square* of the e.m.f., the scale divisions are not uniform.

Such instruments consume no current.

The chief advantage of the electrostatic voltmeter is that it can be directly connected to a circuit of extremely high potential.

In the case of alternating current circuits with voltages up in the tens of thousands it becomes necessary with other types of voltmeters to reduce the voltage before applying it to the instrument.

This is done by means of a "potential transformer," which is a small transformer especially constructed to reduce the high voltage to a lower one suitable to the instrument in use.

The voltmeter is thus not directly connected to the circuit but is inductively connected.

Reading Instruments—Parallax. One source of error in reading voltmeters and ammeters lies in the fact that the pointer or index does not touch the scale.

If the eye is not directly in front of the index a reading to one side or the other is liable to be made.

Where an index stands as much as a quarter of an inch away from the scale, a large error in reading might result.

This error is called parallax and is prevented in portable instruments for accurate testing by the following device:

The index is made thin and flat, lying edgewise towards the eye. A strip of mirror is placed be-



FIG. 57A.

neath the index. The reading is made when by looking down on index and mirror, the index hides its own reflection.

Some switchboard instruments are made for edgewise reading, that is, the scale is at right angles to the base, Fig. 57a. The scale is curved and the index is bent so as to follow it.

In this type, the error of parallax is likely to be greater than with the ordinary type.

Care of Instruments. Electrical measuring instruments must be handled with at least ordinary care. Although made as air tight as possible, dampness and excessive heat should not be permitted to attack them.

Violent swinging of a pointer through excessive current is liable to do more than bend the pointer. Even if it does not permanently injure the instrument it does it no good. This is more likely to happen in using a double scale instrument, or by inadvertently reversing the connections. The glass front should never be rubbed before taking a reading, or the pointer will be influenced by so-called static currents. They may often be dissipated by touching the finger to the middle of the cover, but are best prevented.

A voltmeter made exclusively for direct current should not be tried on an alternating current circuit under any circumstances.

Ammeters for storage battery circuits are made with a central zero as the deflections of charge or discharge are in different directions. This is of course only true of that class of instrument in which the direction of deflection depends upon the direction of current flow, that is, polarized instruments. If a polarized instrument without a central zero be used for storage battery work, a reversing switch becomes necessary between the shunt and the am-

meter. This is however but a makeshift and is by no means to be recommended.

Any instrument built to be used only in a vertical position must not be used when laid horizontally and vice versa. Readings may be made, but they are liable to be inaccurate, owing to friction or gravity.

Although most standard voltmeters and ammeters may be left in circuit continuously without heating, some may not be. It is however a good plan to leave the instrument on open circuit except at the exact time the test or reading is being made, then there is the certainty that the coils have not heated and no possible error introduced.

When taking measurements the instrument should not be set down on a generator or its bed plate.

Most instruments of the D'Arsonval system are unaffected by the proximity of external magnetic fields, but there again precaution is better than uncertainty. Where the instrument cannot be entirely removed from the influence, the presence of the latter may be detected and allowed for by taking several readings with the instrument turned around on an axis running through its centre. If there is a difference, the mean of the several readings should be accepted. Instruments having powerful magnetic fields should be kept apart. Instruments made primarily for alternating current but adapted for occasional use on direct current circuits generally require that the leads be reversed and the mean of the two readings taken.

Multipliers should be tested from time to time to insure that they have not been injured by overheating or careless handling.

Ammeters having internal shunts should be normally left cut out by means of a switch connecting together the external circuit leads. Where the shunt is external it may be left in circuit but one lead from the ammeter should be left open. This is in case there is an unforeseen rise in the current strength while the tester is absent or otherwise occupied.

When measuring large coils such as field coils or those encircling iron, precaution must be taken to avoid a sudden break in the testing circuit or the sudden rush of current due to self-induction in the coil may burn out the instrument.

When two or more compound wound generators are connected together to run in multiple, a so-called equalizer is used. This is a bus-bar or cable connecting together a point on each dynamo between the series field winding and its brush connection. The ammeter or its shunt is not to be connected in series with the lead from the series field or equalizer side, but in that lead which runs directly from the brush to the main switch or bus-bar. Otherwise current flowing in the equalizer bus would not be indicated.

Note. Above all, refrain from tampering with the adjustments of a sealed instrument.

CHAPTER V.

THE WHEATSTONE BRIDGE.

The Wheatstone Bridge. In Fig. 58 is shown the lozenge or kite diagram of the Wheatstone bridge.

A B are two adjustable resistances. These form the proportional arms of the bridge.

R is an adjustable resistance or rheostat, G a galvanometer, B the battery and K K keys closing the battery or galvanometer circuits respectively. x is a resistance which it is desired to measure.

If the battery key be closed, current will divide between A and B and flow through R and x back to battery B .

If A B , R and x are equal, the current will divide equally down each side of the lozenge. Even if the galvanometer key be closed, no current will flow into G .

If x be greater in resistance than R , part of the current through B will travel through G and thence through R in accordance with the law of shunts. The galvanometer needle is thus deflected. But by adjusting R to agree with x the two paths R and x being equal no current flows through G .

Should R be greater than x , current flows through

A and through x by way of the galvanometer, and the needle is deflected, but in the opposite direction to the former case.

The Wheatstone bridge is based on the fact that no current flows between points of equal potential.

Consider the two sides of the bridge AR and Bx as two wires of equal resistance per unit length in multiple between the battery terminals.

The drop of e.m.f. will be the same in each wire for the same length.

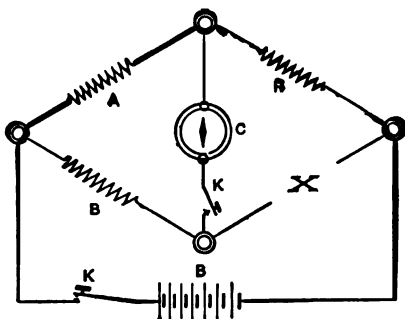


FIG. 58.

If the galvanometer be connected to a point in each wire equally distant from either end, say the middle of each, no current will flow in the galvanometer.

The effect is the same as if both terminals of the galvanometer were placed on the same spot in a single wire carrying current.

Shift one galvanometer connection, this has the

same effect as separating the two contacts on a single wire.

The galvanometer in the latter case will be in shunt with the piece of wire included between its contacts. As the points are at a different potential, the galvanometer will show the potential in that piece of the circuit.

The same applies to the double wires. When the bridge galvanometer is connected as in the figure but assuming ARB and x as equal, the galvanometer is connected between two points of the same potential.

If x be increased in resistance over that of R it would be the same as increasing the length of the wire x .

In the latter case the galvanometer contact could be shifted along the x wire until it again stood at the middle, when the galvanometer would remain undeflected.

In the bridge the resistance R would be adjusted until it was equal to x , A being in the same proportion to R as B was to x , the galvanometer would be at the points of equal potential. Of course B could be changed but R is the adjustable resistance.

If A is to B as R is to x , then x will be the same as B multiplied by R and divided by A .

This is just simple proportion in arithmetic as the following example shows. In simple proportion, to get the fourth term or answer, the second and third are multiplied together and divided by the first.

For example, if $1 = 2$ what will 4 equal?

$$1 : 2 :: 4 : x \text{ or } \frac{2 \times 4}{1} = 8.$$

And in the foregoing formula, if $A = B$ what will R equal?

$$A : B :: R : x, \text{ or } \frac{B \times R}{A} = x.$$

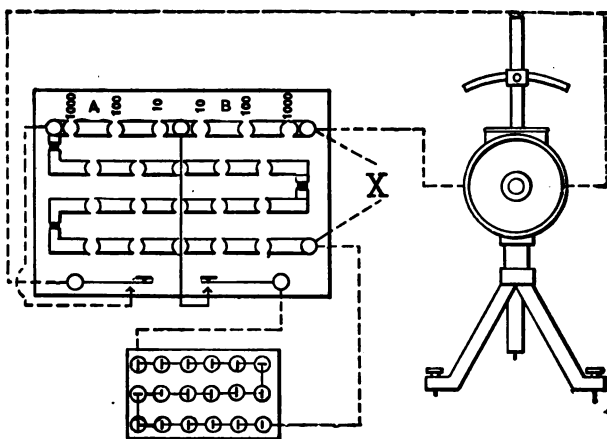


FIG. 59.

The Post-Office Bridge. The connection for resistance tests of the form of bridge known as the (English) Post-Office Bridge is in Fig. 59.

A tripod form of Thompson galvanometer is illustrated and an 18 cell battery but it is evident that a larger or smaller battery could be used.

And any form of galvanometer would be connected up in the same way.

The shunt is not shown, it would be connected across the galvanometer terminals. The resistance to be measured is connected at x .

The key on the left of the bridge is the battery key and should always be depressed first. The right hand key is the galvanometer key and is to be only tapped until a close balance is obtained. Otherwise were the bridge rheostat not adjusted to the measurement being made, the deflection would be violent.

Much damage to a galvanometer is done by neglecting this rule.

In a reflecting galvanometer where the mirror is suspended by a silk fibre, the latter may be broken. Replacing it is a work requiring expert knowledge and vast patience.

Testing with the Bridge. For illustration of the use of the bridge the one shown in the last figure will be selected.

Suppose it is desired to measure the resistance of a coil of wire. The two ends are bared of insulation, scraped clean and inserted in the binding posts as shown at x .

The other connections of the figure are already made as described. For the first test, however, it will be well to use only one cell of battery.

All plugs are in the holes in the rows of resistance coil connections.

Remove from both *A* and *B* the plugs from the 10 ohm coils.

Remove a 10 ohm coil plug from the rheostat part.

Depress the battery key and tap the galvanometer key.

The deflection is noted both as to amount and direction. Suppose it is 60° to the left.

Remove another plug say from 100 ohm coil and tap key again.

Deflection in same direction only 10° .

Remove 20 ohm plug, deflection 5° to right. This indicates that too much resistance has been unplugged.

Replace 20 ohm plug and remove 10 ohm plug, deflection is now 1° but to left.

Indicates that still too little resistance is out.

Remove 1 ohm plug; no deflection perceptible.

Then read resistance unplugged. $10 + 100 + 10 + 1$ or 121 ohms is resistance of wire coil.

When a resistance equal to that being measured is unplugged, no deflection takes place.

It may be stated here that in these examples no actual statement of deflection due to resistance is meant. The values given are only illustrative.

It will be seen that the deflection is to one side when too much resistance is unplugged and to the other side when too little resistance is unplugged.

Most galvanometers fitted in the cases of portable testing sets are marked + and - on opposite ends of the scale.

As the battery is connected so as to send current always in the proper direction, a deflection to the + side means too much resistance and one to the - side too little resistance unplugged. Of course reversing the battery connections will reverse the value of these signs.

In the bridge of the instruments being described in this section, there are three coils in each arm, reading from left to right 1000-100-10 in *A* and 10-100-1000 in *B*.

At present only equal ones are used in each arm. For measuring low resistances the 10 ohm coils in each arm are unplugged. For high resistances, 100 ohm coils in each arm are used.

For very high resistances 1000 ohms in each arm.

The rule is to use the nearest bridge coils to the resistance being measured.

As the total resistance of the rheostat is in this case only 11,110 ohms, it is evident that by the method before pursued, no higher resistance can be measured and none lower than 1 ohm as that is the lowest coil in the rheostat.

Proportional Arms of Bridge. This is where the proportional arms of the bridge come in to use.

By unplugging a higher coil in the *B* side, the value of each coil in the rheostat is multiplied by the number of times the unplugged coil in *A* divides into that of *B*.

And by reversing this and unplugging higher in *A*, the rheostat coils are divided by the number of

times A is greater than B . As a memory aid— A divides, B multiplies.

This gives new values to the rheostat coils and vastly extends the range of measurement.

For example, let A be 10 and B 100, then B is ten times A ; *multiply* the rheostat figures by 10.

The 1 ohm coil then becomes equivalent to a 10 ohm coil and the 1000 ohm a 10,000 ohm coil, and so on.

And let A be 10 and B 1000 and the rheostat is multiplied by 100, the 1 ohm coil is thus equal to 100 ohms, the 1000 ohm to 100,000 ohms.

This last setting of the arms will give a value to the rheostat of $11,110 \times 100$ or 1,111,000 ohms.

If then the bridge arms were so set and a balance obtained at 162 ohms, the actual resistance would be 162×100 or 16,200.

In some bridges there are 1 ohm coils, then the range will be 1000 times and $\frac{1}{1000\text{th}}$.

The dividing method or higher coil in A is on the same plan.

Unplug 10 ohms in B and 100 in A , then the readings will be *divided* by 10, which is the number of times 10 goes into 100. And the 1 ohm coil becomes $\frac{1}{10}$ ohm, the 10 ohm coil now being equal to a one ohm coil.

This proportional property of the arms not only enables larger or smaller measurements to be made

than are in the rheostat, but it gives closer readings.

Take the first example, the coil of wire. Set the bridge arms A 1000, B 10, making a division by 100 necessary of the rheostat readings. As the first test showed 121 ohms, unplug one hundred times 121 ohms or 12,100 in the rheostat.

As the bridge now stands this only equals 121 ohms. It may be necessary to increase the battery.

Having depressed the battery key, depress the galvanometer key and there will probably be a deflection one way or the other. Suppose it is to the - side.

Unplug 10 ohms now $\left(\frac{1}{10}\right)$ and try deflection.

Let it take 18 ohms more to get a balance or zero on the scale.

Then the rheostat will read 12,118 which divided by 100 gives 121.18 ohms, a far closer reading than with equal arms.

It is a good plan to make a number of tests of coils of known resistance, or to check up tests by trying various combinations of the arms.

It must be remembered that as the resistance of copper increases from heat, readings will vary from time to time owing to the current flowing in the wire being tested. The relation between heat and resistance will be found elsewhere in these pages.

In testing coils, electro magnets, etc., they must

not lie near the galvanometer. Where leads are necessary from the bridge to the apparatus being tested, the leads also should be tested as their added resistance would give a false value to the test.

Formula. A stating of the foregoing rules as formulas will be as follows: .

Let R be the resistance unplugged in rheostat, A that in arm A of bridge, B that in arm B of bridge, x the unknown resistance.

Then $A : B :: R : x$, and $A x = B R$; therefore

$$x = \frac{B R}{A}.$$

CHAPTER VI.

PORTABLE TESTING SETS.

Queen Dial Decade Sets. A diagram of the internal connections of the Queen Dial Decade testing set are shown in Fig. 59A. This set has switches instead of plugs, the dials being capable of continuous rotation to gain speed in adjustment.

The connections are plainly lettered on the diagram by the use of contractions, such as *Gr* for ground, and *Ga* for galvanometer. The extra connection and binding post *C* is to enable two unknown resistances or two lines to be conveniently tested without disconnecting either. The proportional arms *A* and *B* of the bridge contain resistance values of 10. 100. 1000 and 10,000 ohms. In addition the *B* arm has a zero contact stud and the *A* arm an infinity stud. The bridge arms provide a ratio of 1 to 1000 or 1000 to 1, while the rheostat has four decades having values of units, tens, hundreds, and thousands. An Ayrton shunt is provided with values of 1. .1 .01, permitting these fractions of the testing current to pass through the galvanometer.

The range of this bridge is from .001 ohm to 10 megohms, although the extreme ranges are rarely suitable to bridge work.

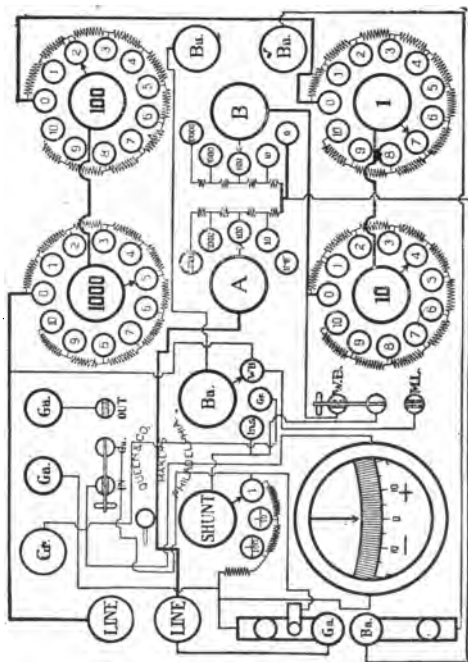


FIG. 59A.

The Peerless Set. This is a small portable Wheatstone bridge, galvanometer and battery, all contained

in a convenient carrying case, and made by the Thompson Levering Company.

The resistance coils are non-inductively wound of manganin wire the degree of accuracy of the



FIG. 60.

rheostat coils being within 1-10 of 1 per cent. and the bridge arm coils 1-20 of 1 per cent.

The brushes of the switches are set at an angle in order to distribute the wear evenly and prevent tracking on the blocks. Permanent connec-

tions between the switches and the coils are used instead of flexible cords, thus permitting the switch to be entirely revolved in either direction and assuring permanent and unvarying contact.

An Ayrton Shunt is included whereby the current passing through the galvanometer may be either full value, one tenth or one hundredth.

The galvanometer is of the D'Arsonval type, hav-

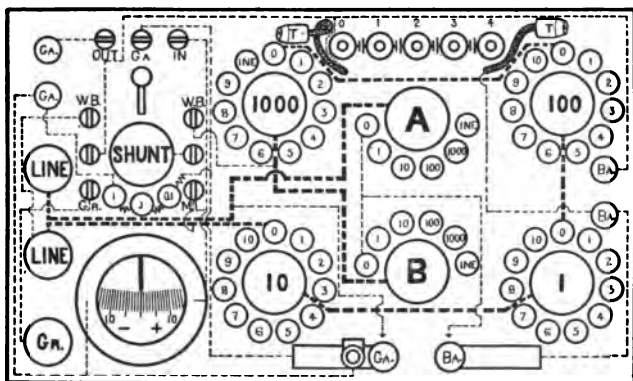


FIG. 61

ing a scale of 30 millimeter divisions with center zero and a short circuit device for arresting the movement of the needle, thus securing quick readings.

A commutator is provided to reverse the arms, providing a testing range from .001 part to 1,000 times the value of the rheostat.

As will be seen in Figs. 60 and 61 the switch

system is used for the manipulation of the resistances instead of the plug system, although either may be furnished if desired. To measure a resistance with the Wheatstone bridge of the Peerless set, the unknown resistance X is connected to the two line posts and both the knife switches set on $W.B.$

Then the proportional arms of the bridge A and B are set as described in the last chapter for bridge measurements and a balance obtained on the galvanometer. While making this balance both the keys Ga and Ba are to be used and sufficient battery cut in with the cords provided for the purpose.

The value of X will then equal $R \frac{A}{B}$.

Provision is made for using an additional battery or another galvanometer if either is required for special tests.

Rule. To find the constant of the galvanometer. The switch B should be set on "inf." The A arm of the bridge is set to 0, and the B arm set to 10,000, the line posts being left open. Then using one cell and the most suitable value of the shunt provided in the case to obtain a good deflection, press down both keys and note the deflection of the galvanometer. The constant will be found by multiplying the resistance and the X deflection together, and dividing the result by the value obtained in multiplying together the number of cells used and the value of the shunt.

All the tests usually required in practical work may be performed with the Peerless set, including the checking of ammeters or voltmeters, the Varley or Murray tests for faults in cables, and others. A study of the diagram Fig. 61 will show the internal connections, and the application of the set to any other test may be then readily traced out.

Caution as to Battery Used. The battery used with a Wheatstone bridge should never be increased over that recommended by the manufacturers. Generally five cells is the limit.

The Queen Portable Testing Set. The Queen Acme portable testing set, Fig. 62, is adapted for all resistance measurements. There are three rows of brass blocks and plugs controlling resistance coils or the sliding switch system may be furnished if preferred.

The middle row is the Bridge, the top and bottom rows, the rheostat.

In the centre of the Bridge is a split block commutator $R X$ which can be connected to the Bridge arms by plugs.

If the plugs are inserted in this commutator in the direction of arrow L as shown, the resistance in the rheostat is divided by the quotient obtained in dividing the higher Bridge arm by the lower.

If plugs are in direction of arrow H , multiply rheostat by quotient. For example, let rheostat

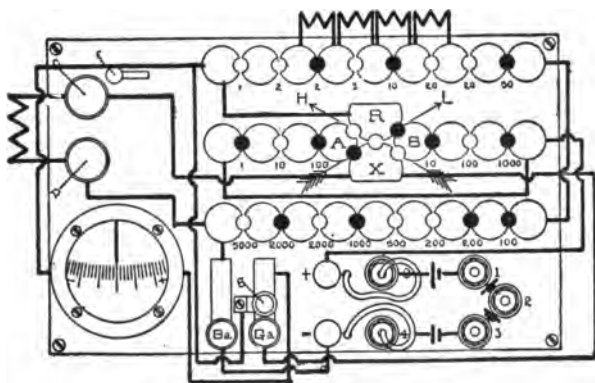


FIG. 62.

equal 100 ohms and commutator be set as in figure. 100 ohms in arm B divided by 10 ohms in arm $B = 10$. Divide rheostat by 10, $\frac{100}{10} = 10$ ohms.

But let commutator be set in direction of arrow H . Then same example, $\frac{100}{10} = 10$, 100 ohms in rheostat $\times 10 = 1000$ ohms.

The galvanometer scale is marked + and -. If the needle swings towards +, reduce the rheostat if to -, increase it.

The range of the Bridge arms is 1 to 1000 ohms, and the rheostat 11,100 ohms. This gives a range of testing from .001 to 11,100,000 ohms. For resistances above one megohm, however, more battery is required than will be found in the case.

Formula. For this Bridge, the formula for the commutator setting in direction of arrow L is

$$\frac{A}{B} = \frac{x}{R}, \text{ and for direction } H, \frac{A}{B} = \frac{R}{x}.$$

The general directions for this set are similar to the regular Wheatstone Bridge tests. It is very simple and easily handled. Other adoptions of the set will be found in later pages of this book.

Table III. gives the Bridge setting for various resistance measurements. It is used in conjunction with the directions regarding the commutator before given.

TABLE III.

Showing setting of Bridge arms to measure resistances as in first column.

"x" Ohms.		Ohms A	Ohms. B
Below	1.5.....	1	1000
From	1.5 to 11.....	1	100
"	11 to 78.....	10	100
"	78 to 1,100.....	100	1000
"	1,100 to 6,100.....	100	100
"	6,100 to 110,000.....	1000	100
"	110,000 to 1,110,000.....	1000	10
"	1,110,000 to 11,110,000.....	1000	1

Whitney Testing Set. This portable set, Fig. 63, has 11,110 ohms in the rheostat and 1, 10, 100 and 1000 ohm coils in each arm of the Bridge.

One plug only is used in each row of the rheostat inserting the plug cuts in a resistance equal to the number marked on the block.

The battery and galvanometer keys can be controlled by one button, the battery contacts being made first. Or they can be depressed separately, as desired.

The inside connections are shown in Fig. 64.

To Use Bridge. Connect the terminals of the unknown resistance to the lower right hand binding posts. The terminal of the flexible cord projecting from the hard rubber top between the bridge arms should then be adjusted to one of the



FIG. 63.

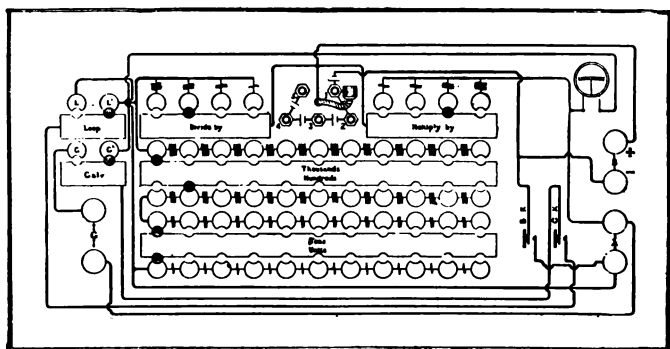


FIG. 64.

five posts, likewise so located; thus throwing in circuit from one to five cells of battery, as will be seen by the diagrammatic illustration herewith.

If more battery power is needed remove the flexible cord terminal from the post and attach the terminals of an external source of direct current to the binding posts *B B*, the positive terminal being attached to the post marked " + ."

This external e.m.f. should not exceed 15 volts.

Assuming that nothing is known of the magnitude of the resistance to be measured, insert the plug in the bridge arm lettered " multiplied by " in the gap numbered " 100 " and the plug in the bridge arm " divide by " in the gap numbered " 100." Place the plug in the " hundreds " row in the gap numbered " 1." The three remaining plugs numbered " 0 " in their respective rows. This act has inserted 100 ohms in the rheostat arm of the bridge.

Depress the combination key momentarily.

If the galvanometer needle deflects towards " - " the rheostat resistance is too small and more should be added by moving the plugs along.

If the deflection is towards " + " the rheostat resistance is too much and less is to be substituted by putting the " hundreds " plug into its " 0 " gap and moving the " tens " plug along.

Proceed in this manner until some combination of coil values is found where the galvanometer needle will no longer deflect when the key is depressed.

The value of the unknown resistance is then indicated by the position of the rheostat plugs.

The above applies only where the unknown resistance has a value between 1 ohm and 11,110 ohms.

If the resistance is lower than 1 ohm insert the plug in the left hand ratio arm into "10" and the right hand ratio arm plug into "100" or "1000". Balance may now be obtained as before, then

$$R \frac{\text{Multiply by}}{\text{Divide by}} = X \text{ or } \frac{B}{A} R = x.$$

For resistances higher than 11,110 ohms the plug in the left hand bridge arm must be inserted in a higher number than the plug in the right hand bridge arm.

When using the set for the Varley loop as a Wheatstone bridge see that the plug in the "loop" bar is in block marked "V. & B." When using set for Murray loop test shift the plug into block marked "M."

The lower right hand binding posts are used for all measurements with this set.

The upper right hand binding posts are for extra battery.

If an external galvanometer is to be used, it is to be connected to the left hand binding posts. The plug is shifted from the right hand hole in block marked galv. to the left hand hole in the same

block. This cuts out the galvanometer in the set and cuts in the external galvanometer.

The Slide Wire Bridge. A simple form of Wheatstone bridge is in Fig. 65.

A piece of resistance wire about No. 24 B. & S. is stretched between two binding posts *D E*. Their distance apart is a little over one metre.

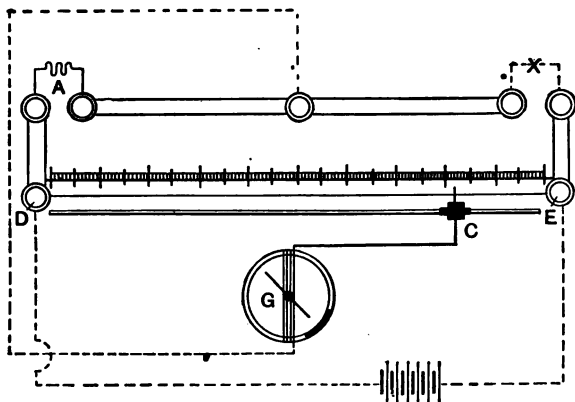


FIG. 65.

Under this wire lies a scale graduated into 1000 equal divisions with a zero at each end.

Brass or copper strips of too small a resistance to be considered lie on the base connecting the binding posts as shown.

The resistance to be measured is connected at *x*. An adjustable resistance is connected at *A*,

partly taking the place of the rheostat in the P.O. and other Bridges.

A battery and galvanometer being connected as shown, the slider C is moved along the wire DE until the galvanometer needle stands at zero.

When the balance is obtained, the resistance of x will equal the result obtained by multiplying A by the length of wire between CE and dividing it by the length between DE .

The length of the wire can be read either in millimetres or divisions of one thousand.

Formula. Let x be the unknown resistance, R the resistance at A , A the length between DC and B the length between CE . Then $R = \frac{R \times B}{A}$.

The Stearns Bridge. An ingenious application of the Wheatstone bridge to the measurement of the resistance of bare wire in continuous lengths is found in the Stearns bridge.

One each of the x terminals of a bridge is connected to a contact device consisting of a metal roller and a form of knife edge.

The bare wire is wound from one drum on to a second drum passing in its passage through both of the contact devices.

The bridge is adjusted to zero by making the rheostat equal the resistance of a length of wire between the contacts.

The drums being then started, each successive

length passes the contacts and a continuous measurement takes place.

The distance between these contacts being unchangeable, the resistance of the wire included between them should also remain unchanged.

The Sage Ohmmeter. The Sage ohmmeter, Fig. 66, is a portable slide wire Wheatstone bridge of



FIG. 66.

peculiar pattern with a telephone receiver as well as a galvanometer.

The adjustable rheostat takes the form of a fine wire, under which resistances are marked in various colors.

A stylus *S* is connected to the telephone *T* as shown in Figs. 67 and 68 and can be touched at

any point on the wire rheostat. Plug *P* cuts in various resistance coils in the Bridge arm, the plug holes being colored to correspond with the markings on the scale.

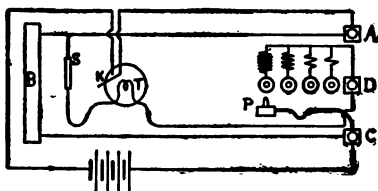


FIG. 67.

In operation the resistance to be measured is connected to posts *A* and *D*. The telephone is held to the ear and the battery key on the telephone is closed. The stylus is tapped at various points on the rheostat wire until no sound is heard

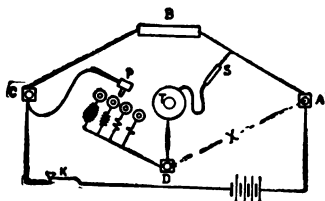


FIG. 68.

in the telephone, the resistance being measured will be found in the figures at the point of stylus contact, where no sound was obtained.

The numbers are to be read in that color corre-

sponding to the plug socket in use. If the plug is in a red hole, read the red figures, and so on. The plug must always be in one of the sockets when testing.

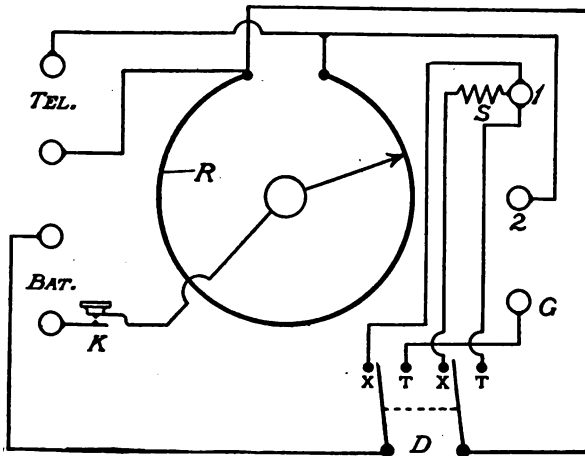


FIG. 69.

The Roller-Smith Fault Spotter. A special form of slide wire bridge (Figs. 69 and 69a) is known as the Roller-Smith Fault Spotter. The wire in this instrument is not stretched out straight but

is wound spirally around a mandrel. This increases the available length of the wire and its resistance,



FIG. 69A.

which are material advantages, increasing the accuracy of the test and the adaptability of the bridge,

These fault spotters are made in two styles, one with a galvanometer and self-contained battery, the other without either battery or galvanometer but with connection posts for the same.

The range of measurements for both models is from 0 to 5000 ohms, there being one scale for resistance values and a second scale for percentages used when locating faults. The standard coil has a resistance of 100 ohms and is adjusted to an accuracy of 1/10 of 1%. The operations when measuring a resistance are as follows: The unknown resistance is attached to binding posts 1 and 2, the switch *D* having been first turned to position *XX*. The hard rubber central button is then rotated, which varies the resistance *R*. Push button *K* is tapped as in all bridge operations and a balance is attained when there is no deflection of the galvanometer or no click if a telephone receiver is employed instead of the galvanometer.

The needle will now indicate on the outer scale the correct resistance.

These operations are extremely simple, in fact the whole test can be made in far less time than it takes to read these descriptions. There is moreover no chance of error as the readings are direct in plain figures.

In locating a ground the line to be tested is looped at its distant end to a good line as in similar tests described elsewhere. The resistance of this loop is then measured and its length in feet calculated from

a table of wire resistances. The switch D is then thrown to TT and the binding post G connected to a good ground. A balance is again attained by rotating the hard rubber button and tapping the key. The position of the brass needle on the inner scale will show the percentage of length of line from the test station. The actual number of feet to the fault is easily calculated from the figures obtained.

Similar instruments are made by several other manufacturers, but as the general principles are the same, descriptions are omitted.

The Evershed Megger. A cursory examination of Ohms law will show that as a galvanometer may be constructed with a fixed integral resistance to show the value of an e.m.f. in volts, so may it be constructed to operate on a fixed e.m.f. and show the value of an interpolated resistance in ohms. Provided the e.m.f. be kept at a fixed value, an ordinary voltmeter may be provided with a scale reading in ohms, as will be described in a later chapter.

It is often inconvenient to keep the e.m.f. at a constant yet practical value in a portable instrument. To obviate this the principle of the differential galvanometer is taken into account in the Evershed Megger. As will be seen in Fig. 70 the moving system of the Megger consists of a coil of insulated wire A connected in series with the armature of the dynamo D and by means of external terminals, in series also with the resistance to be measured. A

pair of coils *BB* connected in series with each other and with a resistance but connected in multiple across the circuit of the dynamo are rigidly attached to the coil *A* and rotate with it. These coils are entirely free to move, being unrestrained by springs or gravity. The electrical connections are made by means of copper strips so as not to offer any restraining force to free movements. The needle or pointer which is actuated by the movements of these coils is therefore liable to rest at any point of the scale while the instrument is not in use.

Current being sent through the coils *BB* by means of the hand dynamo, causes the coils and the pointer attached to take up a position in the weakest part of the magnetic field. The pointer then indicates zero on the scale or infinity.

Upon connecting an unknown resistance at the external terminals, the circuit through coil *A*, which was heretofore open, is now closed, having the unknown resistance in series with it. The current flowing through the coils now induces the latter to take up a definite position. This position indicated by the pointer on the scale is the resistance in ohms of the unknown resistance under measurement.

An increase in e.m.f. will increase the current through both coils to a proportionate extent, the accuracy of the reading being thus practically independent of variations in the speed of the dynamo.

The Megger is dead beat, constant in calibration and does not deteriorate through any weakening of

the magnets. A friction clutch interposed between the armature and the driving gear of the dynamo keeps the e.m.f. constant, invaluable when measuring insulations having considerable capacity.

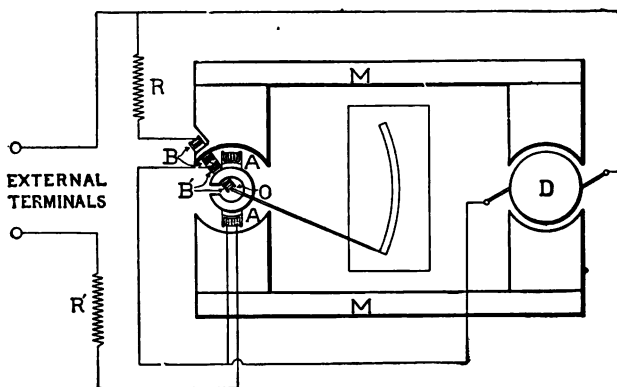


FIG. 70.

The Bridge-Megger. This is a special form of Megger used with a resistance box to perform the functions of a Wheatstone bridge. Its range of measurements is from 0 to 40 meg-ohms, the generator having two windings in series and delivering an e.m.f. of 500 volts. Included in the apparatus is a Change-Over switch for altering the ohmmeter connections so that it will act as a galvanometer, and a Ratio switch for changing the proportions of the bridge arms.

The resistance box is connected to the terminals at *R* and has four or more multiple point switches by which the resistance is varied, the unknown resistance to be measured being connected in at *X*.

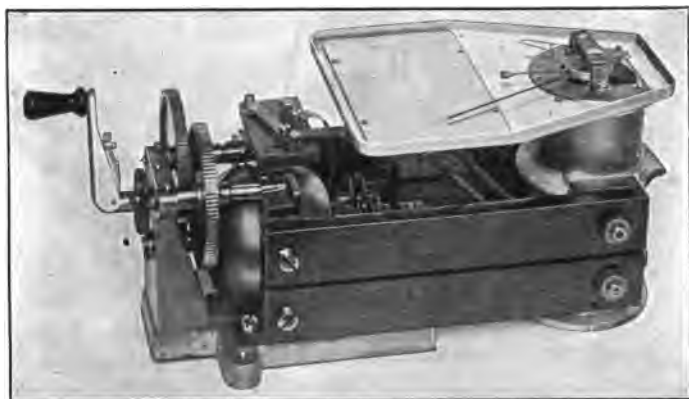


FIG. 70A.

After the preliminary adjustments which are few and simple, the handle of the generator is turned and the resistance varied by operating the switches on the resistance box. When a balance is attained the value of the unknown resistance is read off by the figures on the dials.

There are several types and ranges of Meggers, descriptions of which may be procured from the makers.

CHAPTER VII.

TESTING WITH GALVANOMETER.

Current flow and e.m.f. in a Circuit. The current flow in all parts of a circuit is equal, but the e.m.f. varies according to where it is measured.

In Fig. 71 a fine German silver wire is stretched between terminals $A E$ and connected in series with a battery and an ammeter G ; F is a second ammeter connected in the circuit or G may be shifted.

When the battery current flows through $A E$, the ammeter placed at G or F shows the same reading, the current at G , F , or in fact in any part of the circuit is the same.

Connect a voltmeter V at A and E and note reading, then connect V between $A B$, $B C$, $C D$ and $D E$ and the latter four readings added together will equal the one between A and E .

If the wire $A E$ is of uniform resistance, the e.m.f. between points at equal distances along it will also be equal.

If V be connected as at first to A and the connection at E be drawn along the wire towards A , the e.m.f. will be observed to decrease. This il-

illustrates what is termed "the fall of potential," or if electrical work is being performed, the fall of e.m.f.

If the voltmeter be connected across any two points and a greater current flow be permitted by means of an adjustable resistance R , the e.m.f. will rise with the current.

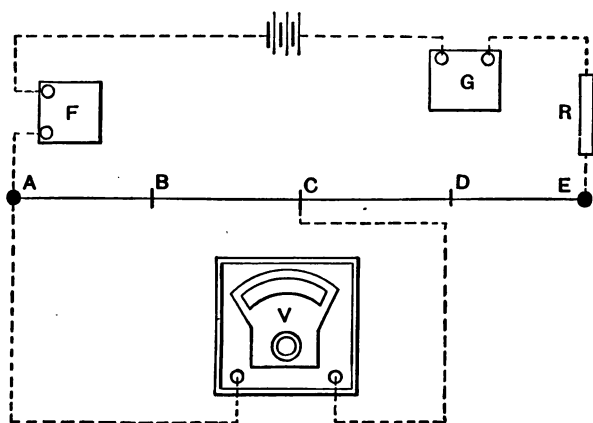


FIG. 71.

As long as the resistance of AE remains unchanged, a greater current flow requires a greater e.m.f.

Applications of Ohms Law. If the resistance between the connections of V be known, the cur-

rent flowing may be computed from the deflection of V by Ohms law. $I = \frac{E}{R}$ or the current equals the e.m.f. divided by the resistance.

For example, let $A E$ measure 2 ohms and V indicate 4 volts, then the current will be $\frac{4}{2}$ or 2 amperes.

If any two readings be known, the third can be computed by Ohms law.

From the e.m.f. and current flow find the resistance. $R = \frac{E}{I}$ or the resistance equals the e.m.f. divided by the current. In this example $\frac{4}{2} = 2$ ohms.

And from the current and the resistance find the e.m.f. $E = R \times I$ or the e.m.f. equals the resistance times the current in this example $2 \times 2 = 4$ volts.

The applications of Ohms law and the fall of potential will be found in the potentiometer, the shunt ammeter and in various tests to be described later.

Testing Resistance. Let an unknown resistance be placed in series with a battery of constant e.m.f. and a galvanometer. Note the deflection and replace the unknown resistance by an adjustable resistance. Adjust the latter until the second deflection is equal to the first and the two resistances are equal.

If the adjustable resistance be of known value this method can be pursued, but a more practical method is to first ascertain the constant of the galvanometer.

The "Direct Deflection" Method. This is the simplest method of testing resistances or insulations and is capable of extended application.

It is based upon the fact that the greater the current flow through the galvanometer the wider the angle of deflection.

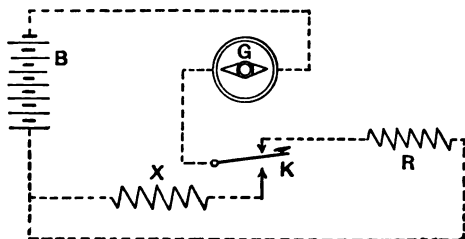


FIG. 72.

A known resistance R , Fig. 72, is put in circuit with the galvanometer G , battery B and double contact key K .

After noting the deflection, the key is depressed and the unknown resistance x thrown in circuit. The second deflection is then noted and compared with the first.

If the galvanometer deflections are proportional to the current, x will be as many times the resist-

ance of R as the deflection through R is greater than that through x .

For example, let R equal one hundred ohms and the deflection through it be ten degrees. The second deflection through x is twenty degrees. This shows that the current flowing through x is twice that which flowed through R and therefore the resistance of x is only one-half that of R , or fifty ohms.

Formula. As a formula, let x equal the unknown resistance, R the known resistance, D the deflection through R and d the deflection through x .

$$\text{Then } x = \frac{D \times R}{d}.$$

In case the galvanometer resistance is to be allowed for it is to be added to R but deducted from x . Calling it r , the formula stands

$$x = \frac{D \times (R + r)}{d} - r$$

Galvanometer Figure of Merit or Constant. It is customary in many uses of the galvanometer to determine the figure of merit or constant of the galvanometer.

This is the resistance through which the galvanometer will give a deflection of one scale degree for one volt of e.m.f.

If 100,000 ohms could be inserted in series with the galvanometer and one volt e.m.f. and still one degree of deflection be obtained, the figure of

merit or constant of the galvanometer would be 100,000 ohms.

And if 50,000 ohms with one and one-half volts gave a deflection of three degrees, the constant would be as follows:

The e.m.f. is one-half as much again as required, therefore the deflection will be one-half as much again or one-third more than it should be. Reducing it by one-third gives two degrees of deflection.

And it is evident that if two degrees are obtained through a certain resistance, one degree would be obtained through twice the resistance.

One degree of deflection only requires one-half the current to produce it that two degrees does, therefore the constant for the galvanometer is $50,000 \times 2$ or 100,000 ohms.

Formula. To make a formula out of this let D be the deflection through the resistance, R be the resistance, V the e.m.f., and x the constant required. Then $\frac{D \times R}{V} = x$ or $\frac{3 \times 50,000}{1\frac{1}{2}} = 100,000$.

The terms constant, figure of merit and sensibility are used to mean the same. Sensibility or constant are the most generally adopted, however.

With Shunts. If a shunt is used in obtaining the constant the deflection and resistance must be multiplied by the value of the shunt which will be known as n , as before.

In the above example let the shunt be $\frac{1}{10}$ with a multiplying power of 10.

The formula will be then

$$\frac{D \times R \times n}{V} = x \text{ and the last example } \frac{3 \times 50000 \times 10}{1\frac{1}{2}}$$

or 1,000,000 ohms.

Using the $\frac{1}{10}$ shunt only one-tenth of the current went through the galvanometer but as it is for example assumed to give the same deflection as at first, its constant is ten times higher.

In some cases the full battery perhaps 100 cells is used to get a constant by shunting the galvanometer.

The constant is then not for one cell but for the whole battery.

Its value will be the product of the deflection, the known resistance and the shunt.

Formula. Using the above letters, (battery of one hundred cells),

$$D \times R \times n = \text{constant.}$$

EXAMPLE. Let the deflection be 16° , resistance 100,000 and shunt $\frac{1}{100}$. Then constant for the battery in use will be $16 \times 100,000 \times 100$ or 160 megohms.

And it is evident that the constant for one cell will be one-hundredth of the above, or 1,600 000 ohms.

Where the figures of resistance become high it is preferable in calculating to substitute the value in megohms or fractions of a megohm.

Second Example. Let the figures used in determining the constant be, $R = .1$ (one-tenth) of a megohm, D be 100° and the shunt $\frac{1}{1000}$.

In testing an unknown resistance with the same shunt and battery the deflection is 10° .

Then the constant is $100 \times .1 \times 1000$ or 10,000 megohms.

And the unknown resistance $\frac{10,000}{10}$ or 1000 megohms.

Summary of Rule. When using the same shunt battery and galvanometer the value of an unknown resistance will be determined by dividing the constant by the deflection obtained through this unknown resistance. If a different shunt s is used the formula will be

$$\frac{D \times R \times n}{d \times s}$$

Deflection Constant. It is often convenient to use the deflection obtained through a known resistance as the constant. The degrees of deflection obtained in a test will then be calculated in terms of the resistance used when getting the constant.

For example, let the deflection through one megohm be 85° . Then the resistance-con-

stant would be one degree through 85 megohms, and the deflection-constant would be 85 degrees through one megohm.

In testing an unknown resistance under the same conditions of battery, shunt, etc., the deflection is 170 degrees. Then $\frac{85}{170} = \frac{1}{2}$ and the unknown resistance is one-half megohm.

Rule: Divide the deflection-constant by the deflection through the unknown resistance; answer is in terms of standard used to determine constant.

Formula. Let R be the standard resistance, D the deflection through it, x be the unknown resistance, d the deflection through x . Then

$$x = \frac{D \times R}{d}$$

If R be one megohm the calculation is simplified for quick working, and becomes $\frac{D}{d}$ in megohms.

Direct Deflection with Queen Set. The Queen Acme set, *T* 460, may be used for tests of this nature; a special set is, however, constructed having a resistance in series with the galvanometer. The latter can also be shunted by manipulating the Bridge.

The *T* 460 set is used as follows:

To obtain the constant, connect a resistance of about 100,000 ohms such as one of the glass slabs,

between the top left hand post and the + battery post.

Remove all plugs from the commutator between the Bridge arms and plug in all coils. Connect in one cell of battery by means of the flexible cords.

The setting should be as in Fig. 73. It will be seen from this diagram that the circuit starting downwards from the slab runs through the + battery post, battery cell, battery key *B a*, rheostat

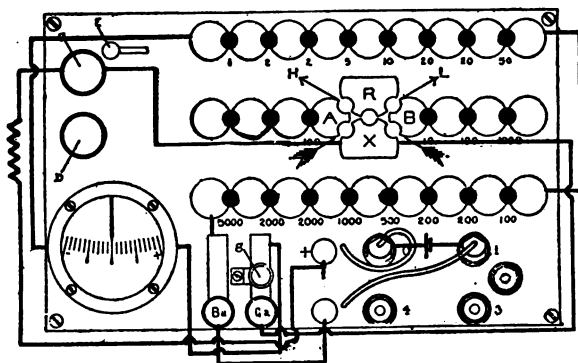


FIG. 73.

blocks, galvanometer, galvanometer key *G a*, to the main post and to the upper part of resistance slab.

The slab, galvanometer and battery will thus be in series when *B a* and *G a* are depressed.

If any plugs have been left out in the rheostat, the resistance of the coils they control will also be included in the circuit.

When *B a* and *G a* are depressed a deflection of the galvanometer will ensue. If this deflection is one degree, the constant is 100,000 ohms.

If more than 1 degree, remove plugs from rheostat and add resistance thereby interpolated to resistance of slab. The constant will be the figure so obtained.

And if less than one degree, a lower resistance than 100,000 ohms must be used, but it is not likely that this will be the case.

The constant having been obtained, detach slab and connect resistance or insulation to be measured in its place.

If the deflection is too small add more cells. A larger battery can be added by connecting it to the battery posts, detaching the flexible cords and cups from the battery in the case.

To determine the resistance now being measured, divide the constant by the deflections obtained and multiply the result by the number of cells used.

For example, let the deflection be 10° with 5 cells.

$$\frac{100,000 \times 5}{10} = 50,000 \text{ ohms.}$$

The deflection has been increased 10 times, which would show that the resistance was only one-tenth of the constant or 10,000 ohms.

But as five times the e.m.f. has been used, the result must be multiplied by five.

Another method is to use the deflection obtained through 100,000 ohms as the constant. Divide this constant by the deflection obtained in the test and

multiply by the number of cells used. The answer will be in terms of 100,000 ohms.

For example, let 8° be the deflection constant through 100,000 ohms with one cell, and 4° be that through the unknown resistance with five cells.

Then $\frac{8}{4} \times 5 \times 100,000 =$ the unknown resistance, or 1,000,000 ohms.

In such tests the resistance of the galvanometer may be neglected.

If it is desired to use only the galvanometer of the Acme set, remove the plugs from the commutator, insert all other plugs and connect to the main posts at left end of case. The galvanometer and key are now directly connected to these binding posts.

Measuring Resistance with Voltmeter. The voltmeter being considered merely as a galvanometer, it may be substituted for the latter.

The deflection obtained from a battery or other source of current is noted.

This deflection is actually through the resistance of the voltmeter.

A second deflection is then taken with the unknown resistance in series with the voltmeter and battery.

This deflection is dependent upon the current allowed to flow through both the voltmeter resistance and the unknown resistance.

The first deflection multiplied by the voltmeter

resistance will give the resistance through which one degree of deflection could be obtained, or the constant.

The value of the second deflection can then be calculated from this constant, the latter divided by the second deflection will give the total resistance. Subtract the voltmeter resistance and the result will be the value of the unknown resistance.

The connections are as in Fig. 74, more extended

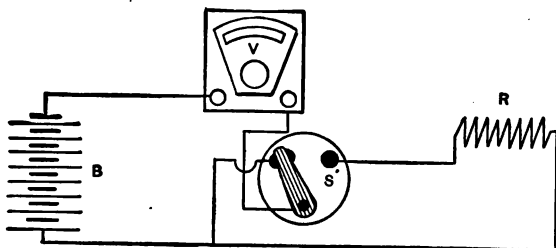


FIG. 74.

description of this test being given in the chapter treating on measurements with the voltmeter.

Testing Resistance of Galvanometer. In many tests it is necessary to know the resistance of the galvanometer being used. If it is not marked on the instrument it may be found by the following method:

Arrange the galvanometer G , Fig. 75, in series with an adjustable resistance R and battery C .

The resistance and battery should be so selected that the deflection of G is about one-half the scale. This is to enable readings to be made with accuracy.

Having noted first deflection, increase resistance R until deflection of G is just one-half of its first deflection.

The battery used must be constant or its variations will affect result. They may, however, be compensated for. Battery resistance is omitted.

The resistance of the galvanometer will equal the resistance of R as measured at the time of one-half deflection less twice the original resistance of R .

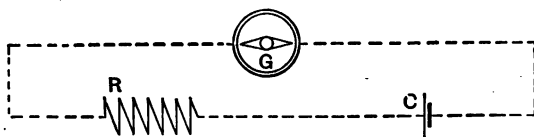


FIG. 75.

For example, let R equal 10,000 ohms and the first deflection be 40° , increase R until at one-half deflection it equals say 25,000 ohms. Then $10,000 \times 2$ or 20,000 from 25,000 leaves 5000 ohms, the resistance of the galvanometer coils.

Formula. As a formula, let R be first resistance, r be second resistance; then $G = r - 2R$.

Resistance of Galvanometer by Bridge. Connect the galvanometer G between the posts used in ordinary resistance tests, as in Fig. 76.

Adjust the rheostat R until the needle of G is at the same deflection when key is open or closed. Then the resistance of G will equal resistance in R multiplied by result obtained from dividing resistance in arm B by that in arm A .

Battery Tests. Tests of battery cells may be to determine e.m.f., internal resistance or life. For the latter no directions will be given.

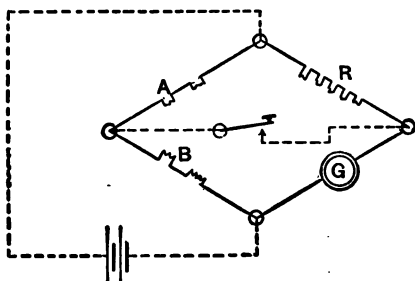


FIG. 76.

The simplest test for e.m.f. is to connect a voltmeter across the terminals of the cell. A high resistance voltmeter must be used, one with not less than 80 ohms per volt is desirable.

The cell being disconnected from any other circuit will give its e.m.f. on open circuit.

Tests of similar nature may be made by discharging cell through resistances of different values.

Tests for internal resistance are needful in many electrical operations and a number of methods will be described.

First Method of Measuring Resistance of a Cell.

Take open circuit e.m.f. of cell with voltmeter of known resistance R .

Add resistance r in series with cell and voltmeter and adjust until e.m.f. deflection is one-half.

Calculation: Add additional resistance to that of voltmeter so as to obtain total resistance in series with cell. Subtract twice the resistance of voltmeter and answer will be that of cell. This test is only suitable where internal resistance of cell is high.

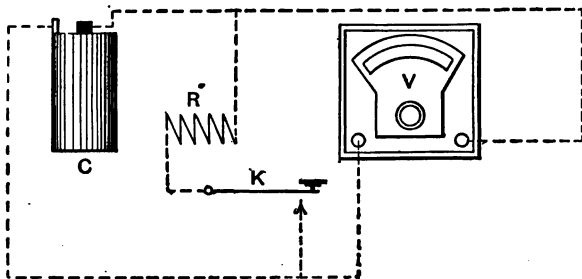


FIG. 77.

EXAMPLE. Voltmeter = 521 ohms, e.m.f. 1.5 volts; adding 530 ohms reduces e.m.f. to .75 volt. Total resistance $521 + 530 = 1051$. This less 521×2 or 1042 = 9 ohms.

Formula. $(R+r) - 2R = \text{cell resistance.}$

Second Method. In Fig. 77 C is the cell whose internal resistance it is desired to measure, V a

voltmeter, R a known resistance and K a key to connect R across the circuit when desired.

First read the e.m.f. of C with the key open; then depress key and read the e.m.f. across R .

Then the resistance of the cell will equal the first e.m.f. minus the second e.m.f. divided by the second e.m.f., or as a formula, $r = R \times \frac{v - v_1}{v_1}$.

The second e.m.f. is really the drop across the resistance.

EXAMPLE.—Let first e.m.f. be 2 volts, resistance be 2 ohms, second e.m.f. across resistance be 1.5 volts, then internal resistance is

$$2 \times \frac{2 - 1.5}{1.5} \text{ or } 2 \times \frac{.5}{1.5} \text{ or } 2 \times \frac{1}{3} = \frac{2}{3} \text{ ohm.}$$

Third Method. Another shunt method is by using a resistance which equals the total resistances in the circuit such as galvanometer, wires, etc. Connect this in shunt across the cell, also connect galvanometer and wires to which shunt was adjusted across cell.

The connection would be similar to the last figure, only the shunt should be directly connected across C .

Having made these connections, note deflection, and then remove shunt by opening key.

Note second deflection. Add resistance in galvanometer (or voltmeter) circuit until second deflection equals first.

The value of the added resistance will be the internal resistance of the cell.

At the first deflection, there was a two branch divided circuit of equal resistances, the current then through galvanometer was one-half the total flowing.

At the second reading the added resistance cut down the current by one-half, but by a series resistance instead of a shunt.

Fourth Method. The simplest test of internal resistance is with the voltmeter and ammeter. Connect ammeter so as to obtain reading of current using a resistance to control current.

Take e.m.f. reading directly across cell terminals.

Then by ohms law $R = \frac{E}{I}$.

In making this last test the current from the cell may be reduced which is an advantage by adding a resistance in series with it. The ammeter is to be included in the circuit.

The e.m.f. V of the cell on open circuit is first read. With the resistance in, the e.m.f. is again read v_1 . The current (I) is also read on the ammeter.

Then r or the resistance of the battery is equal to $\frac{V - v_1}{I}$.

Most of the above methods are only suitable for non-polarizing cells or cells of low internal resist-

ance. The Wheatstone Bridge methods are preferable.

Resistance of Battery by Bridge. In Fig. 78 C is the cell to be measured connected to line posts of bridge.

A key K and the galvanometer G is connected as shown.

Make the reading of G equal whether K is open or closed by adjusting rheostat R .

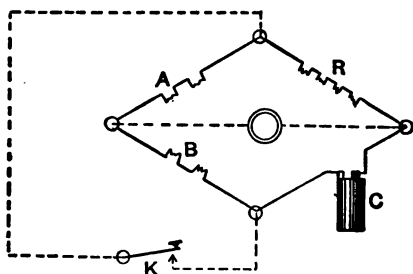


FIG. 78.

Formula. Then the resistance of C will equal the result obtained by multiplying R by $\frac{B}{A}$.

In the formula R is resistance of rheostat, B that of arm B of bridge, and A that of arm A of bridge.

Measuring Current Flow of Cell. The rate of current flow of a cell may be deduced from Ohms

law after reading the e.m.f. and the internal resistance.

But it may also be measured with an ammeter on short circuit.

The latter method is not always desirable. The test can also be made with a tangent galvanometer.

The tangent galvanometer can be used as an ammeter providing the directive force of the earth's magnetism be known for the place where the measurement is made.

This directive force is used in the form of a constant. It varies however, from time to time.

The constant of the place of measurement is multiplied by the radius of the galvanometer coil in inches and the result divided by the number of turns in the galvanometer coil. This gives a constant for the galvanometer.

When a current produces a deflection of the needle, the tangent of the angle of deflection multiplied by this galvanometer constant equals the current flow in amperes.

EXAMPLE.—A cell of battery applied to the terminals of a 10 turn coil which had a radius of 6 inches produced a deflection of 18° . The test was made in New York.

Constant of the galvanometer equals $\frac{.744 \times 6}{10}$
or .4464.

Tan (tangent) $18^\circ = .3249$.

Then $.4464 \times .3249 = 1.4518$ amperes.

Formula. Let I = current flow; H constant of location; r radius of coil; N number of turns of coil; G constant of galvanometer; T tangent of deflection. Then $G = \frac{H \times r}{N}$ and $I = G \times T$, or

$$I = \frac{H \times r}{N} T.$$

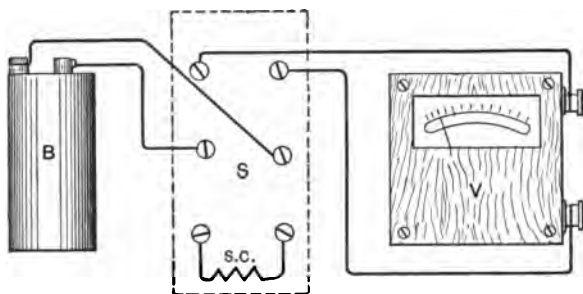


FIG. 78A

Testing the Polarization of a Battery Cell. It is often desired to determine the rate of polarization when comparing different battery cells.

The connections of a simple and easily operated apparatus are shown in Fig. 78A.

The cell of battery B to be tested is connected to the centre or hinge posts of a double pole double throw switch S . A voltmeter V is connected to two of the outside posts at one end of the switch, and the two remaining posts are short circuited by

a short piece of copper wire sufficiently large to carry the current.

The switch is first thrown to the right and the e.m.f. of the cell noted. The switch then being thrown to the left, short circuits the cell for the desired interval, say 30 seconds. The reading taken on the voltmeter with the switch thrown to the right will show the e.m.f. of the cell after the short circuit. Its fall of e.m.f. due to the polarization caused by the short circuit is calculated by subtracting the second reading from the first. According to the nature of the tests desired the cell may again be short circuited and a number of successive readings be taken showing the polarization of the cell on repeated short circuitings. The readings may also be made to show the polarization due to a short circuit of any length of time.

Again any desired resistance may be substituted for the copper wire. Tests may be desired to show the rate of recuperation of the cell after polarization, in which case readings will be taken after successive rests.

Different cells may be compared as to polarization and recuperation. It is also possible to compare two cells by connecting them in opposition and short circuiting them. Then the readings will only be the difference of e.m.f. between them, the one having the greatest sustaining powers giving the deflection. The particular cell may be located by the direction of the deflection.

With most voltmeters it will be necessary to include a reversing switch, otherwise the deflection may be in the reverse direction.

This test is not suitable for storage cells when the short circuiting wire is used, as the current flow would be too great.

CHAPTER VIII.

THE POTENTIOMETER.

The Potentiometer. Referring back to Fig. 71 let the negative terminals of two independent battery cells be attached to *A* and their positive terminals to *E*. No current will flow along *A E* if the e.m.fs. of the cells are equal. This can be proven by inserting a voltmeter in series with either cell or between *A E*. One e.m.f. is opposing the other e.m.f. If now two cells in series be applied to *A E* and a third cell be connected in multiple between *A E* as before, a voltmeter or galvanometer will show current due to the excess of e.m.f. from two cells against one. Or if all the cells are equal, the deflecting force will be equal to that of one cell.

Connect a voltmeter or galvanometer in series with the one cell and leaving the negative terminal connected to *A* move a wire from the voltmeter along *W*. A point will be reached where the voltmeter or galvanometer will settle to zero showing the e.m.f. of the one cell is equal to that opposing it from the two cells. As the wire *W* is of uniform resistance, the e.m.f. across one-half of

it will be one-half that across its entire length. The point of zero will then in the present case be in the middle at C , and the relation between the two lengths of wire will equal that between the two e.m.fs. Or the length AC will be to AE as the e.m.f. of the one cell will be to that of the two cells.

Although there is some similarity between a potentiometer and a Wheatstone bridge there exist two great differences. The bridge has one source

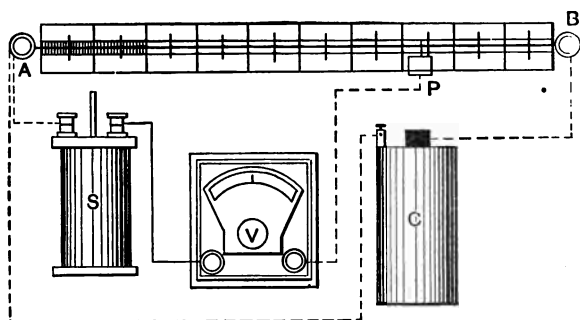


FIG. 79.

of e.m.f. and two circuits; the potentiometer two opposing e.m.fs. in one circuit.

A balance is obtained in the bridge when the points of equal e.m.f. in two circuits are connected through the galvanometer.

In the potentiometer the point of equal e.m.f. between two e.m.fs. is found in one circuit.

A simple form of potentiometer is shown in Fig. 79. Between the binding posts AB is

stretched a fine resistance wire. A sliding contact P moving on a guide rod carries an index which indicates on a scale graduated into 1000 equal divisions. This scale may conveniently be one metre long and the wire No. 24 B. & S. gauge. A standard cell S is connected, one terminal to A and the other terminal to the voltmeter V . The second terminal of the voltmeter goes to the sliding contact P . If V is not provided with a key one may be inserted between A and P . The cell the e.m.f. of which is to be compared is connected to A and B . Similar terminals of each cell must be connected to A .

Contact P being moved, S and C being both in circuit, a point on the wire will be found where V gives a zero reading.

When the balance is obtained, the scale divisions from A to P will bear the same relation to those from A to B that the e.m.f. of cell S does to the e.m.f. of C . The e.m.f. of S being known the e.m.f. of C is easily calculated.

Let S be 1.019 volts, AP be 755 and AB 1000. Then $755:1000 :: 1.019$ is to C or $\frac{1000 \times 1.019}{755} = 1.349 \dots \dots$ volts.

If AP were 10 and AB 1000, C would equal 101.9 volts. It is evident that S should be less than C or no point of zero will be found.

As this is a zero method, that is the object being to reduce V to zero, V need not be calibrated in volts but may be any form of galvanometer. If

it were desired to check a voltmeter, the latter would be inserted across $A B$, the reading on this voltmeter should equal the e.m.f. obtained by the calculation.

As the sliding of the contact P only cuts in or out a resistance formed by the wire, the resistance of the latter may be determined. Each scale division will then have a value in ohms or fractions of an ohm.

To make this clear, let the wire be for example 10 ohms, then each of the 1000 divisions on the scale will equal $\frac{1}{100}$ of an ohm.

And in the foregoing example 755 divisions would equal 7.55 ohms.

The whole example would read 7.55 ohms : 10 ohms : : 1.019 is to 1.349 volts. the same result as before.

To check an ammeter it would be placed in series with a steady source of current. A shunt or known resistance is also in *series* with the ammeter.

Leads of no appreciable resistance are run from each end of the shunt to A and B , The e.m.f. across this shunt is first determined. The current flowing is then calculated by dividing the e.m.f. across the shunt by the resistance of the shunt, or

$I = \frac{E}{R}$. This should equal the reading on the ammeter. The shunt must be of sufficiently high resistance to give a drop of e.m.f. greater than

the e.m.f. of S . For example, let the shunt be 5 ohms, and the e.m.f. across it be 10 volts. Then $\frac{10}{5} = 2$ amperes. Some of the foregoing rules may be modified in actual work with the more complicated forms of potentiometers, but the principles remain the same.

If it is desired to measure a high e.m.f. and yet not apply it in full to the potentiometer, any desired e.m.f. may be taken off by means of resistances. A number of resistance coils being placed in series with the high e.m.f., connection may be made between two or more for the desired e.m.f. In the case of an e.m.f. of 100 volts, let 10 volts be needed for purposes of measurement. Place in series two coils one of 100 ohms and one of 900 ohms. Connecting across the 100 ohm coil will give an e.m.f. of 10 volts, across the 900 ohm coil, 90 volts, and across the two coils, 100 volts. The reason for using such high resistances is to keep down the current. The actual resistances used may vary according to circumstances, their relation to one another is only of importance.

In some tests it is desirable to connect the known source of e.m.f. across AB and the unknown across AP .

By using a charged storage battery and suitable high resistance a steady e.m.f. may be obtained for some hours. Dry cells are not suitable as their curve of e.m.f. drops too abruptly.

In using the potentiometer with a large battery

of constant e.m.f. the latter is connected to $A B$, its positive pole to A .

A standard cell S is connected, positive pole to A and negative pole through voltmeter or galvanometer to P .

P is moved until there is no deflection. The divisions between A and P are noted, or if the scale is graduated in terms of resistance of W , the ohms are noted. Call this R .

The standard cell being replaced by the cell under test, a point of balance is again sought. Call the resistance of divisions from A to P in this case r . Then the e.m.f. of the cell being tested is to the e.m.f. of S as r is to R .

It may be necessary in order to obtain greater accuracy to use two or more standard cells in series between A and P .

For example, let e.m.f. of the standard cells S be 2.038 $A B$ equal 1000 and $A P$ equal 100. Then the e.m.f. across $A B$ will be $\frac{1000}{100} \times 2.038$ or 20.38.

Substituting the cell C to be tested for S it is evident that zero balance will be again obtained when its e.m.f. balances the e.m.f. between $A P$. It may be then read in terms of the e.m.f. across $A B$ or in terms of S .

Suppose balance now is obtained at 125. This will be $\frac{125}{1000}$ of $A B$ or 2.547 likewise $\frac{125}{100}$ of S .

In figures $\frac{125}{1000} \times 20.38 = \frac{125}{100} \times 2.038$.

The modern testing sets are generally suitable for use as potentiometers as the following examples will show.

As plug switches are used instead of sliding contacts, the relations corresponding to divisions on a scale will be those of resistances in ohms. In the testing sets, the resistance equivalent to that between AP , Fig. 79, is made 100 ohms and is not changed. The resistance equal to PB is adjustable and is first made high.

As the sliding of P merely changes the *relation* between AP and AB the same result is obtained by changing only PB . This which is not feasible in the slide wire instrument becomes an advantage in the plug switch pattern.

In the following two tests consider the B arm to represent the wire AP and the rheostat the wire PB . The B arm plus the rheostat will then equal the total resistance of wire W or from A to B .

Although the examples given are measurements of e.m.f. or current, the potentiometer is equally well adapted for comparing resistances and insulations.

Checking Voltmeters. Voltmeters may be checked by comparison with a standard instrument or by the potentiometer method.

In the former case, the standard instrument and the one being checked are connected in multiple across the e.m.f. at the same point and their readings compared.

Potentiometer Method with Queen Set. Unplug 10,000 ohms from rheostat, unplug *A* arm of Bridge and unplug 100 ohms only from *B* arm. Remove plugs from commutator except one in upper right hand hole, that is, connecting *R* and *B* blocks.

Disconnect flexible battery cords from battery tips. Connect negative terminal of a standard cell or cell of known e.m.f. to line post *C* and its positive terminal to + battery post.

Connect line e.m.f. to battery posts, + to + and - to -.

Depress *B a* and *G a* keys and change resistance in rheostat until no deflection occurs. If needle goes towards + on scale, reduce rheostat and vice versa. Proceed carefully as the balance will occur at a point of slight change in the rheostat.

When balance is obtained, add 100 ohms to the reading of the rheostat and divide by 100 or point off two places.

Multiply this result by e.m.f. of standard cell and answer is e.m.f. of outside circuit.

For example, suppose balance was obtained with 8855 ohms in rheostat and e.m.f. of standard cell is 1.44.

$$\text{Then } \frac{8855 + 100}{100} = 89.55 \text{ and this multiplied by}$$

1.44 = 128.95 volts as e.m.f. of outside circuit.

Formula. Let *R* = rheostat at balance; *B* =

resistance of B arm, $e = \text{e.m.f. of standard cell}$, and E , e.m.f. to be found.

Then $\frac{R+B}{B} e = E$, or what is the same thing,

$$\frac{R+B}{B} \times e = E.$$

Checking Ammeters and Voltmeters. When checking ammeters and voltmeters there are several methods to pursue. Perhaps the simplest is by the use of standard instruments which are connected in with the instruments being checked. In the case of an ammeter test, the standard ammeter is to be included in series in the same circuit with the ammeter under test. A voltmeter is connected in shunt across the same part of the circuit as the voltmeter under test.

An ammeter may be checked with a tangent galvanometer. As will be seen in chapter I the tangent of the angle of deflection is proportional to the current flowing. In order to give the deflections of the tangent galvanometer definite values in amperes, either a standard ammeter may be connected with it in series or the values may be arrived at by calculation. The method is explained in Chapter VII. but is of use only in laboratory or experimental work, as it involves too much chance of error and too many operations for the inexperienced. The simpler methods are those which use the potentio-

meter or are based upon a simple calculation of Ohms law.

In a shunt ammeter where the resistance of the shunt is known, a measurement of the e.m.f. across the shunt, and the resistance of that shunt give the

known quantities for the equation $I = \frac{E}{R}$.

EXAMPLE. Let the resistance of the shunt be

.005 and the e.m.f. across it be .05, then $\frac{.05}{.005}$ equals

10 amperes, which is what the ammeter should show at the time.

Where the resistance of the shunt is not known and it is inconvenient to measure it, a known resistance of low value may be inserted in series with the ammeter and the readings taken as before.

The electrolytic method of checking an ammeter is based upon the electro deposition of copper.

A solution of copper is placed in a suitable vessel and two copper plates of known weight suspended in the solution. The two plates are then included in series with the ammeter to be checked, and a source of e.m.f. Current at the rate of one ampere to each 8 inches of copper surface exposed on the cathode plate is allowed to flow for a stated time. The cathode plate is then removed, washed, dried and weighed again. The gain in weight is the de-

posited copper and the latter should equal .000325 gram per ampere per second of deposition.

As the current reading on the ammeter is noted and has been kept constant by means of a rheostat, the correction is obtained by calculation. This method is of service in the laboratory and is not used in commercial calibration.

Checking an Ammeter with the Whitney Portable Testing Set. The method adopted is to balance the drop across a shunt of low resistance (across which the ammeter is connected) against a fraction of the e.m.f. of a Standard cell.

Reference to the schematic diagram in Fig. 80 will show that the shunt S forms a common path for the current from e.m.f. A and e.m.f. B , but these two e.m.f.s. are in opposition as to their polarities. If both e.m.f.s. are equal, no current would flow through S and no deflection would occur on the instrument placed at V .

In the method used with the Whitney set, the e.m.f. A is a Standard cell and the e.m.f. B whatever source of current is being used on the ammeter V . By suitable use of the rheostat in the set only a portion of the e.m.f. from the Standard cell A is being used as in the potentiometer method, although the entire scheme is not the potentiometer method of testing ammeters.

The actual operations with the set are as follows: The flexible battery connector being thrown off re-

moves the enclosed battery from the circuit. The plug is removed from the multiply by or *B* side

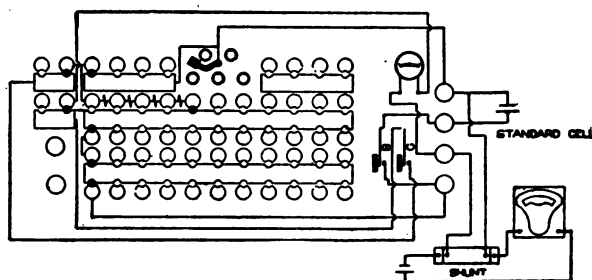
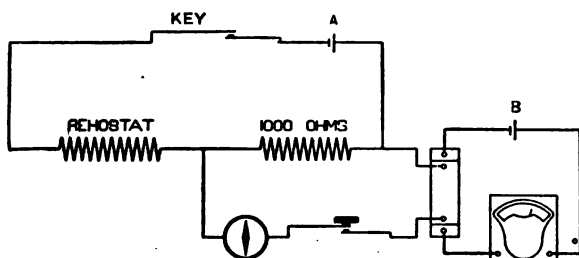


FIG. 80.

of the bridge and 100 ohms is plugged in the *A* or divide by arm. Then 3000 or 4000 ohms is plugged

in the rheostat and a low resistance of known value is connected in series with the ammeter under test. As the ammeter is already connected in its own circuit, the positive side of the resistance (which is connected to the positive battery post and the negative side of the resistance) is connected to the upper *X* post. The + side of the standard cell is connected to the + battery post and the — side to the — post. The combination key of the set is now closed and the rheostat changed until a balance is secured.

The potential difference is found as follows: Divide the voltage of the standard cell multiplied by the value of the bridge arm, by the value of the resistance in the rheostat plus that of the bridge arm.

The potential difference thus obtained divided by the value of the standard low resistance, gives in amperes, the current flowing through the ammeter.

EXAMPLE. The following are the figures of a test with the apparatus: Calculate the amperes that should be indicated on the ammeter. The e.m.f. of the Standard cell is 1.019 ohms. The standard low resistance measures .001 ohm. Balance is found with the rheostat at 1938 ohms.

$$\text{Then } \frac{1.019 \times 100}{1938 + 100} \text{ equals } \frac{101.9}{2038} \text{ or } .05 \text{ volt.}$$

.05 times .001 equals 50 amperes.

Checking a Voltmeter. The Whitney set may be used to check a voltmeter (Fig. 80A) in a manner very similar to that adopted in checking an ammeter.

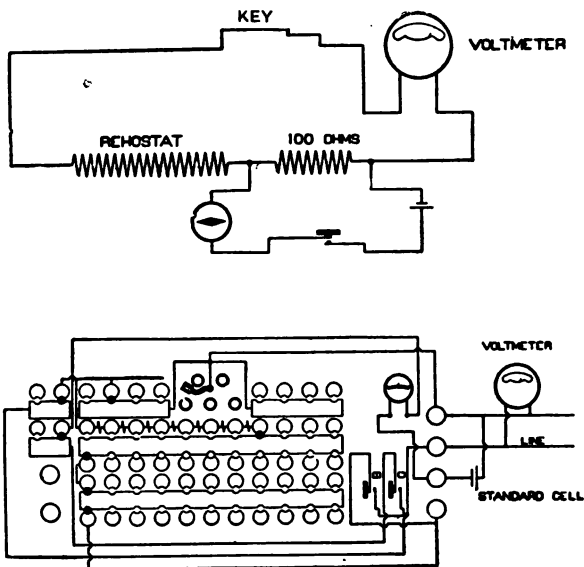


FIG. 80A

The set is prepared by removing the flexible battery terminal and connecting the circuit to which the voltmeter is connected to the battery posts. Care

is to be taken that the polarities are correct, positive to positive and negative to negative. The plug is then removed from the *B* or multiply by side of the bridge, and 100 ohms plugged in the *A* or divide by side of the bridge. 8,000 to 10,000 ohms is then plugged in to the rheostat. A Standard cell is connected to the set, its positive terminal to the positive battery post and its negative terminal to the upper *X* post.

Press the combination key and if the galvanometer deflects towards the plus sign, the rheostat is too high. If the deflection is to the minus side the rheostat is too low. When the balance is found add 100 to the rheostat reading, point off two places and multiply by the voltage of the standard cell.

EXAMPLE. Consider that a voltmeter on a 110 volt circuit is under test and the rheostat shows 10.000 ohms.

Then $\frac{10,000 + 100}{100}$ equals 101. and 101. times 1.019

(the e.m.f. of the standard cell) is 102.9 volts, which should be the reading on the voltmeter.

Calibration Curve. In checking voltmeters or ammeters it is customary to make a chart on which curves are drawn to indicate the error of the instrument. This calibration curve or correction chart

is then used with the instrument, if it is not desired to correct the instrument itself.

Alternating Current Tests. It should be clear that the tests for checking voltmeters or ammeters are equally applicable to the measurement of the electro motive force or current of direct current circuits. The measurement of alternating current presents peculiar conditions which come more within the province of a work on electric light and power.

CHAPTER IX.

CONDENSERS.

Charge and Discharge of Condenser. Connect a condenser C , battery B , galvanometer G and key K as in Fig. 81.

Depress K and a deflection will be noted at G . This is due to the charge of the condenser.

Release K and upon the back contacts meeting,

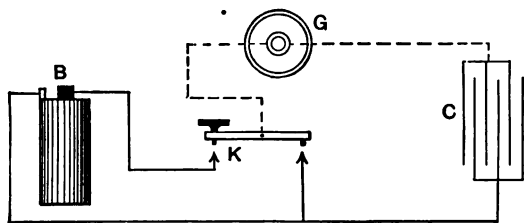


FIG. 81.

a second deflection takes place at G but in the opposite direction.

Again depress K so as to open the back contact but not close the battery circuit. Upon again releasing K a deflection will ensue but very much less than the previous ones. And this last operation being repeated, decreasing deflections will be noted.

If K be depressed for varying intervals of time and released the deflections will also vary.

It takes a certain length of time to charge a condenser. Increasing the battery will reduce the time of charge.

If a Kempe discharge key is used the connections will be practically the same. The back contact of the key in the figure will be the top contact in

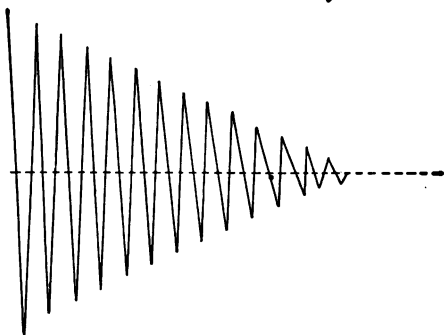


FIG. 82.

the Kempe key, and the front contact the lower contact of the Kempe key.

The capacity of the condenser is measured by the extent of the deflection upon first short circuiting it through the galvanometer. This is true when battery power and other conditions are equal.

If the discharge of C be observed with an appropriate apparatus, it will be seen to be oscillatory as in Fig. 82, not steady as that of a battery cell.

A condenser does not store up electricity, the operation called charging merely leaves it under an electric strain which is relieved when it is discharged.

It may be likened to a spiral spring which is suspended from one end and has a weight at the other end. If this spring be pulled still farther down, an ability to do work or a potential is given it. Now upon releasing the spring, it contracts, pulls up the weight, and then slightly expands again, causing the weight to jump up and down until the force is exhausted.

The spring and weight have an oscillating motion even as does the condenser discharge show an oscillatory character.

A condenser offers the resistance of its insulation to a direct current, in other words will not allow a current of this character to pass through it.

But the effect of an alternating current is not arrested by a condenser. An ordinary vibrating bell will not be rung by a battery through a condenser except one tap due to charge. The alternating current of a telephone magneto will ring the bell with which it is equipped without trouble through a condenser.

This does not show, however, that the alternating current actually passed through the dielectric. It is the inductive effect between the plates, a condenser acts therefore by the induction taking place between its parts.

To avoid complications of words the condenser

is said to retain a charge and to permit itself to be discharged.

Measurement of Capacity. The capacity of a condenser is measured by comparing the discharge deflections with those obtained from the discharge of a condenser of known capacity. The latter is termed a standard condenser.

The operation is as follows:

Charge standard (S) for a certain length of time, say one minute, then discharge it and note deflection (D).

Then replace standard by the condenser under test (C). Charge and discharge under precisely similar conditions and note deflection (d).

The connections may be similar to Fig. 81 or the battery may be connected directly to the condenser through the key. The galvanometer would not be then in circuit during charge. Or the galvanometer may be shunted or short circuited.

The condenser under test will bear the same relation in capacity to the standard that the deflections do to those obtained by discharging the latter.

Formula. Or as a formula,

$$S : C :: D : d. \quad \text{Then } C = S \frac{d}{D}.$$

Another method by the bridge is as follows: A standard condenser and the condenser to be tested are connected in a similar manner to the

two arms of the bridge. Each condenser takes the place of the coils in one arm.

An adjustable resistance is connected in the binding posts provided for the x or unknown resistance.

A discharge key and battery are connected taking the place of the regular battery in the bridge. The condensers are charged and discharged, and the resistances adjusted in the other sides of the bridge until no deflection obtains.

At balance the two condensers are to each other in respect to capacities inversely as the two resistances are to each other.

EXAMPLE. Capacity of standard condenser S is .3 (three-tenths) microfarad. Resistance R on same side of the bridge is 1800 ohms. Resistance r on side of bridge where condenser T to be tested is connected, is 1200 ohms. Then $T : .3 :: 1800 :$

1200 or $\frac{.3 \times 1800}{1200} = .45$ microfarad.

Formula. Using above letters, $T : S :: R : r$.

Insulation of Condenser. A condenser can be tested for insulation by charging it through a shunted galvanometer. When fully charged the deflection should return to zero.

If a deflection is shown after charging is completed it indicates that current is passing through condenser. The value of this current and the resistance of the fault in the condenser are computed

from the galvanometer constant as in any direct deflection method.

But this method does not give as close a test of very high insulations as the loss of charge method.

Insulation by Loss of Charge. A condenser retains its charge (or electric strain) for a length of time dependent upon the perfection of its insulation. This property gives a method for determining its insulation as well as by the usual resistance tests.

A condenser is attached to a battery for a given length of time and then discharged through a galvanometer. The deflection having been noted, the condenser is again charged under precisely similar conditions and left with its terminals insulated for a given period of time and again discharged through the galvanometer.

The two deflections being compared the loss of deflection in the second reading is a measure of the loss of charge. This loss being due to the state of insulation, the latter is therefore determined.

In a good standard condenser this loss for a short period of time is inconsiderable.

As cables and wires act in a similar manner to condensers, their insulation can also often be determined by the loss of charge method.

The following figures are approximate as the actual substance or fluid varies somewhat. For purposes of comparison, however, the figures are sufficiently accurate.

TABLE IV.

Specific inductive capacity taking air as 1.

Flint Glass.....	10
Plate Glass.....	7.06
Gutta-percha.....	2.62
Mica.....	5
Paraffin wax.....	2.31
Pitch.....	1.80
Shellac.....	2.59
Sulphur.....	3.82
Olive oil.....	3.16
Castor oil.....	4.78
Turpentine.....	2.23

CHAPTER X.

CABLE TESTING.

Cable Testing. If a resistance be shunted across a condenser it will be found that no steady deflection can be obtained until the condenser is charged.

The effect of shunting the resistance around the condenser has had the practical result of adding capacity to the resistance.

A few experiments with a key, battery and galvanometer will show what the effect would be of adding capacity to a signalling circuit.

As the insulation in a cable acts as a condenser dielectric, the choice of an insulation of little capacity is necessary (see Table IV.)

One of the tests of a cable is to determine its capacity, the other two main tests are as to its insulation and the conductivity of the conductor or core. The insulation of a cable may be that between its core and the earth. Or in the case of a multiple cable having several separate cores, the insulation between these cores.

Where cables are armored with iron or steel wire, or lead covered, the armor or lead covering forms one of the two connections in insulation or capacity tests.

As in most cases, the cable armor or lead sheath is in direct contact with the earth, a connection may be had with the latter. But a good contact on the armor or sheath is preferable.

In testing multiple cables, all the cores not being tested at the time should be grounded. This is done by twisting a bare copper wire around a bared end of each conductor and then to the armor or sheath.

If no very delicate instruments are at hand, all the cores may be so connected or "bunched" omitting the armor connection. An insulation test made between the bunch and the ground will give the multiple insulation of all the cores. This must be less than that of one core as the copper surface to the insulation is so much increased. If a large deflection results, the cores may be thrown off one at a time, watching the deflection. Should it suddenly decrease upon detaching a core from the bunch, trouble is evident in that core.

In preparing core ends for attachment to the testing circuit, much care must be exercised. The insulation should be removed in a tapering form as one would point a lead pencil.

Absolutely no dampness must exist on the insulation or current will be conducted from the copper to the sheath and destroy the value of the test. A good plan is to coat the cable from the copper under the binding posts with hot paraffin.

In testing cables one point will be noticed, the deflections on an insulation test will gradually de-

crease. This is due to electrification of the cable. The insulation apparently improves.

The deflections are generally read after one minute's electrification.

In a multiple cable the actual resistance of each conductor must be worked out and then all the resistances added together. Dividing this by the number of conductors will give the average insulation of the cable.

In speaking of insulation, the cable length must be specified. For instance, if a piece measuring 1000 feet had an insulation of 1000 megohms, a piece only half as long would be 2000 megohms. Its insulation surface and therefore its leakage surface is only one-half.

It is customary to reduce the result to the average per mile. In the foregoing example, the insulation would be 189 megohms per mile.

The rule is to multiply the insulation obtained by the length of the cable reduced to miles or parts of a mile.

In the direct deflection method of testing cable to be described, the same reading should be given whichever terminal of the battery is grounded. The negative terminal of the battery to the cable conductor shows up a fault in a cable quicker than the positive pole for reasons to be given later.

The temperature at which a test of insulation is made is of prime importance. The insulating properties of gutta-percha, rubber and compounds of similar nature decrease very rapidly with in-

crease of temperature. Taking the case of gutta-percha at 50 degrees Fahrenheit as of unit resistance, at 60 degrees it would be less than one-half, at 75 degrees only one-fifth, and at 90 degrees only one-twelfth. With other insulations the decrease of resistance may be even greater than this. It is customary to test cables at 75 degrees Fahrenheit or to correct the readings to that temperature.

Tables of corrections will be found in most large works on testing, such as Kempe's "Hand Book of Electrical Testing," or Hoskaier's "Electrical Testing of Telegraph Cables."

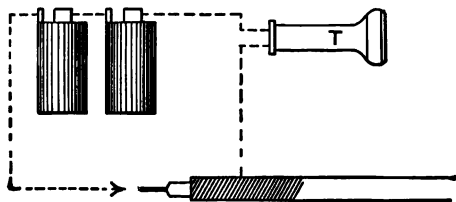


FIG. 83.

Telephone Test of Insulation. A simple test of cable insulation may be made with a telephone receiver and battery as in Fig. 83.

One wire makes good connection with the cable armor or lead covering the other wire being touched to the cable core.

Upon making the first contact with the core a click will be heard in the telephone owing to the charge of the cable.

But if the connection be kept closed for a minute or so the cable will be charged.

Subsequent tapping will give no click if the insulation be good. The clicks are of course due to current leaking through the insulation.

Cable Testing for Insulation Resistance. In Fig. 84 *B* is a battery, *R* a reversing key, *G* a galvanometer, *S* a shunt box, *C* the cable, *E* an earth con-

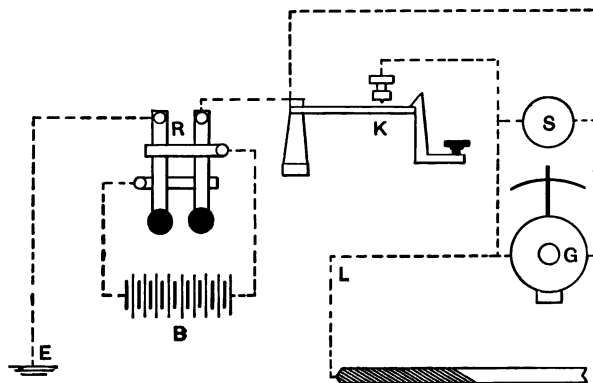


FIG. 84.

nection, *L* a lead from the cable core to the galvanometer connection. *K* is a short circuit key, that is, one which can be left closed to short circuit and cut out the galvanometer and also be opened to cut in the galvanometer.

Before connecting *L* to the cable, test the insulation of *L*. Connect one end as in the figure to

the contact of K , depress one button of R and open K , using a shunt according to the sensibility of G and strength of B .

No deflection should appear.

If any deflection deduct it later from readings in testing cable.

Close short circuit key K ; connect other end of L to cable and open K . Any deflection of G will be due to earth currents.

If the deflections so obtained are in the directions of deflection when battery is used, deduct same from cable readings, if in the reverse direction, add them.

Shunt G with say $\frac{1}{100}$ shunt, closing K and press one button of R . As long as K is closed the current all goes into the cable in a similar manner to charging a condenser. After ten seconds, open K , keeping R still down and a deflection of G will ensue.

If off the scale use a higher shunt or reduce battery and repeat charge of cable.

If too small, use lower shunts or increase battery.

Repeat the latter test several times, allowing cable to charge for one minute, noting readings. Then repeat as many more times, but depress other button of R , reversing battery and deflection.

The values of these readings may be worked out from the galvanometer constant and an average taken to represent the insulation of the cable.

The foregoing cannot necessarily cover every minor point, but the main operations are given.

To sum up: First, obtain the constant of the galvanometer; this operation was described elsewhere. Second, test insulation of the leads. Third, test for earth currents. Fourth, charge cable. Fifth, cut in galvanometer and shunt. Sixth, reverse battery and repeat fifth. Seventh, strike average of readings.

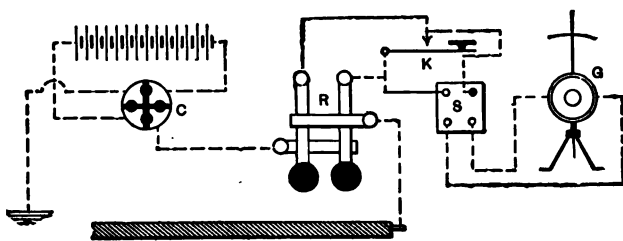


FIG. 85.

Principle allowances to be made are for: shunt, battery, insulation of leads and earth currents.

A slightly different connection for the above test is shown in Fig. 85.

The cable is connected to the reversing key *R*, and a commutator *C* permits of reversing the battery connections, putting either terminal to ground. Or the battery can be cut out entirely and the cable tested for earth currents by directly grounding the connection between *R* and *C*.

Testing Ground Connection. In most cases the sheath or armor of the cable forms the ground. Where this is not so, or in other cases where a ground is used, it is often necessary to test the ground connection.

There are several methods available, depending on whether the result is to be merely approximated or to be obtained with accuracy.

Connect positive terminal of battery to ground and negative terminal through shunted galvanometer to ground.

Shunt should be so adjusted that deflection is on the scale. If necessary, put resistance in series with galvanometer. Next short circuit ground connection by connecting positive wire to ground wire on galvanometer.

The deflections in each case will show the state of ground connection. The better it is the less difference there will be between the two deflections.

Measuring Resistance of Ground Connections.

When it is desired to ascertain the actual resistance of a ground connection as used in telephony or telegraphy an efficient method is that described by D. B. Grandy in the *Telegraph and Telephone Age*. It contemplates the use of a voltmeter or galvanometer with a considerable scale of known value and a Wheatstone bridge.

The diagram Fig. 86 shows the bridge arms AB , the voltmeter V , the rheostat R , battery C , key K ,

unknown ground x and a ground of negligible resistance G . The resistance of the lead L from the rheostat R to x is to be measured first by any of the resistance measurement methods. With all resistance R cut out and the battery K open, deflections on voltmeter V will be due to earth currents at the

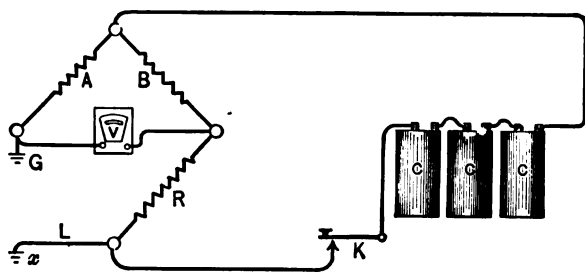


FIG. 86.

ground connections. A mean average of several deflections is adopted and termed Zero, or Z . Then with the positive (+) side of the battery to the terminal between L and R , close the battery key and alter R until deflection is obtained equivalent to Z . The resistance in R will now be termed X .

Reverse the battery and again make a balance deflection equal to Z . Resistance now in R will be termed Y .

To sum up.

Z = Deflection due to earth currents

Y = Resistance with positive pole to x ground

Y = Resistance with negative pole to x ground

The joint resistance of the two readings will be

$\frac{XY}{X+Y}$ which multiplied by 2 = $\frac{2XY}{X+Y}$ or the

resistance of the circuit including L , G and x . As G

is negligible and L of known value $x = \frac{2XY}{X+Y} - L$.

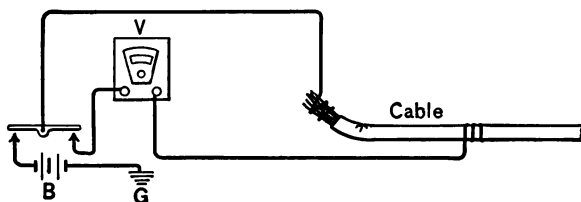


FIG. 87.

Cable Testing by Loss of Charge. In testing the insulation of cable by the loss of charge method, the cable is charged and discharged, its capacity being compared with that of a standard condenser.

The cable is then charged for one minute and disconnected with its terminal free in the air. This is done by using a key similar to the Kempe discharge key before described.

At the end of one minute, the cable is discharged and the reading noted. Comparison of several such readings shows the ability of a cable to retain a charge, *i.e.*, its insulation.

The connections using a voltmeter or a galvanometer are shown in Fig. 87, where V is the voltmeter, B the battery and G the ground connection.

Formula. Let D be the deflection upon instantly discharging the cable, d that after one minute. Then the percentage of loss of charge will be

$$\frac{D - d}{D} \times 100$$

In many cases where delicate apparatus is not at hand the loss of charge method can be used for approximate results. The throw of the needle due to the charging or discharging of a cable is usually sufficient to be easily readable on a voltmeter or galvanometer of low sensibility. The principle involved may be likened to that of testing a water tank for leakage.

The holes or pores might not be detectable in showing the water leaking out. If the tank were filled to a given mark and left closed to guard against evaporation, a glance after the lapse of several hours would show whether the level had lowered or had remained stationary.

The writer has tested multiple conductor submarine mine cables in New York Harbor with a few cells of dry battery and the galvanometer in a portable testing set.

When the direct deflection method failed owing to want of sensibility in the galvanometer, the difference of deflection made after charge, insulate and discharge gave results.

Capacity Tests. The capacity test of a cable is similar to that described in testing condensers.

A standard condenser is charged and discharged and the galvanometer deflections noted. The cable is then substituted for the standard condenser and the deflections noted. Comparison of the two results gives a comparison of the capacity of each.

The calculations are made as in condenser tests.

Conductivity Tests. The conductivity of a cable conductor may be measured with a Wheatstone Bridge as in any similar test for conductivity or resistance.

When both ends are available they are connected to the bridge in the usual manner.

If only one end is at hand it is often possible to use a second conductor of the same cable, looping the two together at the distant end. The resistance of one will be then one-half the reading if both are similar in diameter, material and length.

If a ground connection becomes necessary it

must be made with the greatest care that it imposes no undue resistance.

Locating Faults in Cable. If the grounded fault were of practically zero resistance and the ground connection between the fault and the battery also of zero resistance, the location of a fault would be easy.

It would then only be necessary to measure the resistance of the cable core up to the fault from both ends of the cable. As the core is of uniform resistance, the two results would be in the same ratio as the lengths of the cable.

For example, let the total length of the cable be 10,000 feet and its total resistance 21 ohms. Measuring from one end *A* the resistance to the fault is 14 ohms. From the other end *B*, the resistance is 7 ohms. Then as 14 is $\frac{2}{3}$ of 21, the fault would be $\frac{2}{3}$ of the total length of the cable from *A*. Or 6666 feet from *A* and 3333 feet from *B*, neglecting fractions.

But ground connections or faults rarely fulfil the above condition, in fact they usually vary from time to time. It becomes necessary to use a test such as the Varley or Murray, which disregard the resistance of the fault.

A method of locating a fault as in Fig. 88 may be adopted in certain cases.

A good cable and the faulty one are looped at

the distant end D . The resistance of the loop $A B$ is then measured by the Bridge method. One x terminal of a Bridge being grounded, the other terminal is applied to A and the resistance $A F$ is measured, and similarly the resistance $B F$. It is evident that both $A F$ and $B F$ include the resistance of the fault to earth.

The resistances $A F$ and $B F$ being added together, subtracting the resistance of $A B$ will leave twice the resistance of F to earth. Twice the resistance because it has been added in both $A F$ and $B F$.

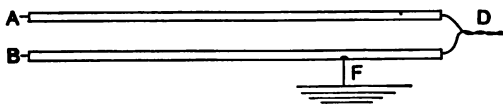


FIG. 88.

If now the resistance of F be subtracted from $B F$ it will give the resistance of B to F . Knowing the resistance of the cable per foot the distance can be determined.

EXAMPLE.—Let $A B$ measure 22 ohms; $A F$ equal 26 ohms, and $B F$ 16 ohms. Then $26 + 16 = 42$, and less 22 = 20 ohms. One-half this, or 10 ohms, is the resistance of the fault to ground. The resistance of $B F$ or 16 ohms less 10 = 6 ohms.

If the cable conductor resistance were 11 ohms per mile, the fault would be $\frac{6}{11}$ of a mile from B , or 2880 feet.

To properly understand the Varley and Murray tests consider a Wheatstone Bridge the arms of which are equal. The rheostat is replaced by a length of cable and the unknown resistance is also replaced by a length of cable, both being similar in resistance per foot.

If both lengths are the same, their resistances are the same, the Bridge balances and no deflection takes place.

Now shorten one length and add resistance R in series with it until the Bridge again balances,

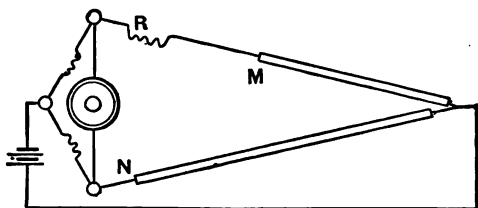


FIG. 89.

Fig. 89. The added resistance equals that of the piece cut off.

If the resistance per foot is known, the length of the shorter piece can be easily calculated.

Instead of connecting in the battery by a wire as shown, use earth connections at each point.

The only result will be to interpose a resistance in the battery circuit. It will not affect the balance but only reduce the effective power of the battery. This may be overcome by enlarging the battery.

Before making the test the copper core resistance of the loop between $M N$ must be known either by calculation or measurement.

The latter is known as the Varley test, a more complete description is as follows:

The Varley Test. In this test for a fault or earth connection the Bridge is connected as in Fig. 90. The battery is grounded and its circuit is completed by the ground at the fault F . The faulty

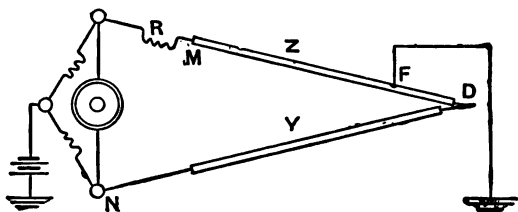


FIG. 90.

cable is looped or connected at its distant end D with a good cable.

To obtain a balance it is clear that the rheostat r must be adjusted until its resistance added to that of the cable core between M and F together equal the resistance of the core from F to N .

It is also clear that the resistance of the ground at F only reduces the energy of the battery current.

The test is therefore independent of the amount of core exposed to the ground at F .

Formula. Using the letters in Fig. 90, let z be the resistance of cable from F to M , Y that of the cable from F to N , R the resistance of the entire cable loop, F the resistance of the fault, r the resistance of the rheostat when galvanometer is balanced.

Then $Y + z = R$, and $Y + F = r + z + F$
or $Y = r + z$.

Z then equals $\frac{R-r}{2}$.

The distance x of the fault from M is

$$\frac{R}{z} = \frac{L}{x} \text{ or } x = \frac{z L}{R}$$

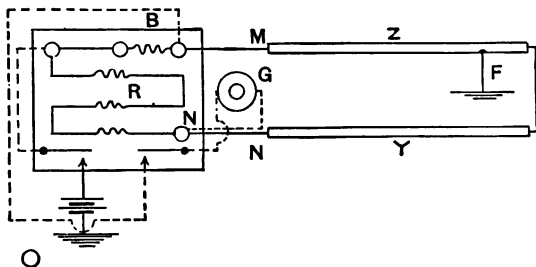


FIG. 91.

The Murray Test. This is somewhat similar to the Varley test but one arm of the Bridge is made adjustable. One regular Bridge arm is used, the other being replaced by the rheostat giving an arm of large adjustment.

The circuit is given in Fig. 91. B is one Bridge arm, R the rheostat now the second Bridge arm,

Z the faulty cable between M and the fault F , Y the remainder of the faulty cable and also the good cable in the loop.

The rheostat is adjusted until there is no deflection. From a previous discussion of the Bridge this will be seen to occur when the resistance R bears the same proportion to Y as B does to Z . Or $R : Y :: B : Z$.

Suppose for example only the total length of the loop from M to N is 400 feet. Its resistance is 40 ohms. A fault occurs at a distance from M to be determined.

Arm B is 100 ohms. R is adjusted and on balance reads 300 ohms. R is therefore three times B .

Y must bear the same proportion to R that B does to Z ; here Y must be three times Z .

As Y and Z added together equal the whole resistance of the cable, Y will be $\frac{3}{4}$ and Z $\frac{1}{4}$ of this resistance (or length).

Y therefore equals 30 ohms, and Z 10 ohms.

Or if the length is required, Y is $\frac{3}{4}$, that is, 300 feet, and Z $\frac{1}{4}$ or 100 feet.

Rule. To obtain the length of Z by means of a rule for calculation. The resistance B is to be divided by B added to R and the result multiplied by the length of the cable loop or L .

Same example, $\frac{100}{100 + 300} \times 400$ or $\frac{1}{4} \times 400 =$
100 feet.

Formula. $R:Y::B:Z$, or

$$\frac{R}{Y} = \frac{B}{Z}.$$

$Z + Y = L$, then $Z = L - Y$.

x or distance of F from $M =$

$$\frac{B}{B + R} \times L.$$

CHAPTER XI.

TESTING WITH THE VOLTMETER.

In telephone exchanges, electric light plants and similar places the majority of the commercial tests required in the daily routine may be made with a direct current voltmeter and a source of e.m.f. That this is being recognized more and more as the best practical method is shown by the large number of excellent direct current voltmeters which are being manufactured having scales and resistances adapted to the requirements of resistance and insulation tests.

A voltmeter is merely a galvanometer with its deflection values marked in volts, and may be used in place of a galvanometer of the same class provided its sensibility is sufficiently high. The most suitable instrument to use is one constructed on the Deprez-D'Arsonval type previously described. It should have a high sensibility. The better class of voltmeters have a resistance of 100 ohms per volt, but special voltmeters may be procured with a resistance of over 1000 ohms per volt. The sensibility of the latter instruments is therefore proportionately greater, but they are much more delicate to handle than the less sensitive ones.

The two factors that determine the range of measurement are the sensibility of the instrument and the e.m.f. available.

EXAMPLE. Consider a voltmeter of the above type having a scale reading from 0 to 150 volts in divisions of one volt. Its resistance at 100 ohms per volt would be 100 times 150, or 15,000 ohms. Applying an e.m.f. of 150 volts, the maximum resistance that could be measured when included in series with the voltmeter would depend upon the smallest fraction of a division which could be read by the operator.

A deflection of one division would indicate an added resistance of $150 \times 15,000$ minus the resist-

$$\frac{2,235,000}{}$$

ance of the voltmeter, giving a value of 2,235,000 ohms.

A deflection of one half a division would indicate twice the resistance, or $\frac{150 \times 15,000}{.5}$, minus 15,000 or 4,485,000 ohms.

As the sensibility is fixed, an increase in the e.m.f. would increase the value of the readings and a decrease in the e.m.f. would decrease them. For instance a deflection of one division on the same instrument but with an e.m.f. of 300 volts would equal $300 \times 15,000$ minus 15,000 or 4,485,000 ohms. And with an e.m.f. of 15 volts, the deflection of one divi-

sion would indicate $15 \times 15,000$ minus 15,000 or 210,000 ohms.

This should all be perfectly clear if it is remembered that it takes a certain amount of electrical power to move the armature of the voltmeter a given distance. An added resistance reduces the power applied to the armature and reduces its movement, as shown by the lessened swing of the pointer on the scale.

Although of service in many tests, a double scale instrument does not increase the range of measurement but aids in making close readings.

EXAMPLE. Suppose the voltmeter have two scales, one reading from 0 to 150 volts and the other reading from 0 to 1.5 volts. The resistance of the two windings will be probably 15,000 ohms and 150 ohms respectively.

Employing a low e.m.f. on the low scale, resistances of low value would be the easier measured, while on the high scale, a high e.m.f. would be preferable to read high resistances.

Using an e.m.f. of 1.5 volts on the low scale, a deflection of 140 degrees, or 1.4 volts would indicate the value of the added resistance to be,

$$\frac{1.5 \times 150}{1.4} = 150 \text{ or } 10.7 \text{ ohms, which would be}$$

difficult to read on the high scale using an e.m.f. of 150 volts.

The Source of e.m.f. When testing resistances or apparatus which can be kept free from grounds it is immaterial whether the source of e.m.f. be a battery, or an electric light circuit, provided it is direct current and steady.

An electric light plant, even a small isolated one that is free from grounds on the system is a rarity, although they may be of such high resistance as to constitute no danger to the apparatus.

It is often possible to remove the ground from the testing e.m.f. by various devices. Sometimes where the ground is on the wiring system, the latter can be disconnected for a period. It is rare that there is a ground on the generator so bad that it cannot be allowed for in the tests. Moreover a large proportion of electric plant tests are for the very purpose of determining grounds on the generating equipment.

A fruitful source of error in testing the insulation of generators, motors or electric circuits carrying a voltage of one hundred volts or over, is the ground caused by the tester or an assistant holding poorly insulated testing leads in his hands, or touching current carrying parts and causing a ground connection through his body. This should always be borne in mind when making tests.

The e.m.f. must be steady or errors will be introduced into the calculations. Not less than three readings should be taken for each test, one of the e.m.f., one of the work being tested, and another

immediately after, of the e.m.f. to determine whether it has changed since the first reading. Where the e.m.f. fluctuates it will be necessary to take a succession of readings and average them.

The available source of the testing current may be

- 1 A circuit of direct current for lighting or power purposes.
- 2 A number of storage cells.
- 3 A number of dry batteries, either those used in telephone work or for ringing bells.
- 4 A thermopile, which latter giving a steady but small current is not often available.
- 5 A magneto generator taken from a telephone set and provided with a commutator to rectify the current.

The first four need little description, the requirements have been stated, and with the exception of No. 4 they are generally available. With regard to No. 5, a telephone generator (Fig. 92) having five magnets and with a winding suitable to deliver a high voltage is readily procurable, or rewound. Those in use on the regular telephones for ringing magneto bells give an average of only 75 to 80 volts. A commutator is easily made, particularly as the armature of the generator has but two poles. The principal difficulty in using such a device is to keep the e.m.f. steady and thus not affect the calculations. A method which works well in practice is to

use two voltmeters, one of which may be the switch-board voltmeter. An assistant watches this voltmeter, which is connected in circuit all the time and turns the handle of the generator, keeping the



FIG. 92.

e.m.f. steady while the test is being made. In an engine room it will be easy to run the generator with a pulley and a belt from one of the engines or it may be belted to a motor, especially to an alternating current motor which is maintaining a steady speed.

The liability of grounds and the absence of portability of the source of e.m.f., together with the fact

that a calculation is required are the principal arguments used against the voltmeter for any other service than testing the e.m.f. of a circuit.

The question of grounds and unsteadiness is one that can best be met for each individual case. Portability is a requisite only in special cases, and then it may often be secured by using a dry battery or the magneto before referred to. The calculation requires merely a few seconds at most and is well within the ability of a person fit to be entrusted to make tests of any value at all. With regard to the quick testing devices on the market, most of them are expensive, whereas a voltmeter is to be found almost everywhere a test is desired.

It is noteworthy that within the last few years many voltmeters and wattammeters designed for the testing of resistance and insulation have been placed on the market and have come into very general use.

Even when the latter are not available it may again be noted that the direct current voltmeter generally found in the central station or isolated plant is well capable of use in many instances, as will be realized when the conditions governing its use are fully understood.

Testing Resistance with Voltmeter. In Fig. 93 is shown the direct deflection method of testing resistance R with a voltmeter V . S is a switch and B a battery.

The e.m.f. of B is first noted by moving the switch arm to the left. The arm is then moved to the right, and the e.m.f. through R noted. The first deflection showed the deflection of V caused by the battery B through the resistance of V . The second deflection was through the resistances of R and V . If V was 10,000 ohms and the first deflection 10 degrees, only one degree of deflection would be obtained through ten times 10,000 or

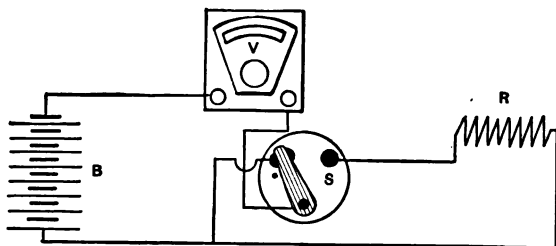


FIG. 93.

100,000 ohms, which latter is the constant for V . Let the second deflection be 5 degrees. The resistance of R is therefore $\frac{100,000}{5}$ or 20,000 ohms, but the resistance of V must be deducted. $20,000 - 10,000 = 10,000$ ohms as the resistance of R .

The constant for the voltmeter may be calculated as above, or a formula may be used for the entire calculation.

Formula. In the formula let V be the first deflection, V_1 the second deflection, and r be the resistance of the voltmeter,

Then $\frac{V \times r}{V_1} - r = x$ or the resistance of R .

For measuring low resistances a low reading voltmeter or a millivoltmeter is preferable if a battery or low e.m.f. be used.

In many cases the current in a power circuit may be used instead of an extra battery.

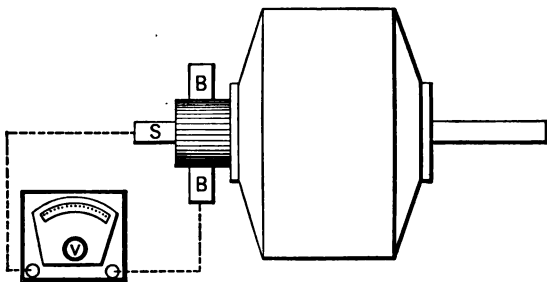


FIG. 94.

Insulation of Generator with Voltmeter. In testing the insulation of a generator in operation, connection is made as in Fig. 94 where V is the voltmeter, S the shaft or part of the framework, and $B B$ the brushes.

The generator furnishes its own testing current. The e.m.f. across $B B$ is first noted, then the e.m.f. from B to S and the last formula applied.

Care must be taken that the voltmeter is not affected by the generator and that all connections and contacts are good.

Insulation Resistance of Armature. A test of the insulation resistance of the entire armature may be made with the armature removed or disconnected from the fields.

A bare copper wire is wound tightly around the commutator so as to touch all segments. The testing circuit is between the shaft and this copper wire. This will give the combined leakage paths between all coils and segments to the frame.

The copper wire ensures a better contact with all segments. Where the armature coils are in series with each other and tapped to the commutator, no lower insulation will be noted than between one segment and the shaft. That is, providing all armature coil connections are good. The voltmeter should be in circuit with a tap from the mains as in the test for insulation when generator is running.

If the armature is in its place and all field and brush connections complete, the test will show the entire insulation of the machine. Brushes must in the latter case rest on the commutator.

Testing Electric Light Wiring. A short circuit on an electric light system generally indicates itself. On the other hand, it may be desired to test the mains for such before cutting in the dynamos.

Testing an extensive system of wiring for short circuits from a main switchboard is a difficult task without having an idea of the lamps, etc., connected to it.

If there were 1000 lamps in multiple between the leads, the resistance might be as low as one-fifth of an ohm.

Such a low resistance between two leads would be a veritable short circuit if it was due to defective insulation. If it is practicable to turn off all lamps first, the matter does not then present much difficulty.

A simple test on a circuit of say 110 volts would be to use an incandescent lamp of the same voltage.

This lamp would be connected between one side of the main being tested and a source of e.m.f., perhaps a dynamo. A wire is run from the other side of the circuit to the other terminal of the dynamo. A short circuit on the main would cause the lamp to light up. Of course the dynamo is to be disconnected from the main by its switch.

If a voltmeter be used instead of the lamp, the actual resistance is determined by simple calculation.

The e.m.f. of the dynamo is to be multiplied by the resistance of the voltmeter and the result divided by the e.m.f. observed when in series with the circuit being tested. Deducting the resistance of the voltmeter the answer is the actual resistance between the two sides of the main.

In either method, the actual location of the trouble is assisted by opening the branch circuit switches on the board one at a time. When the faulty circuit is reached upon opening its switch, the lamp goes out or the voltmeter reading decreases.

The practice of connecting pilot lamps across the circuit at motors and other apparatus is liable to cause perplexity.

A magneto bell would ring easily through one lamp. It would be troublesome to measure low resistances such as one lamp on the voltmeter, as the change of deflection would be slight. The lamp test would therefore be preferable for a short circuit.

A short circuit in a branch leading from a cut out may be hunted for after replacing one of the fuses by a lamp. The lamp will burn until the short circuit is removed. A lamp twisted loose from its base, or a faulty socket is best found by this method.

If the circuit be cut between the fault and the cut-out, the test lamp in the cut-out will cease to burn.

Measuring Switchboard or Line Insulation. This method has been before described in the direct deflection tests, except that here one terminal of the voltmeter is grounded.

A brief summary of the test between a switchboard bus and the ground will be given.

Take reading of total e.m.f. of circuit.

Take reading between bus and a ground connection.

Multiply resistance of voltmeter by e.m.f. of circuit, divide result by e.m.f. of bus to ground, and deduct resistance of voltmeter.

Answer will be ohms to ground of opposite bus to one used in calculation.

Should both sides of the circuit be grounded the calculations become vastly more complicated.

When testing one side, the voltmeter will be in multiple with the fault from the side to which it is connected, and vice versa.

And the two ground connections will form a connection between the mains of a resistance equal to the resistance of each ground added together.

If it were possible to disconnect one main from all connection with the other main, the test would be simple. The insulation to ground of each main would be ascertained while the other main was disconnected. But this is not often practicable.

Another method would be to disconnect both main wires from the bus-bars or from the current source, strap them together with a piece of bare copper wire, and measure the insulation to ground as if they were one wire.

This would give the joint insulation-resistance of the main to ground which after all is what is needed most in practice.

Formula. Let V equal e.m.f. of circuit, V_1 e.m.f.

between main to ground, R resistance of insulation to be ascertained, r resistance of voltmeter.

$$\text{Then } R = \frac{V \times r}{V_1} - r.$$

In using the voltmeter for measuring resistances there is another method of working which gives the same result.

First deduct the second reading on voltmeter from the first reading. Then divide this result by the second reading and multiply by resistance of the voltmeter.

EXAMPLE. Let e.m.f. of testing circuit or that between mains be 110 volts, resistance of voltmeter 15,000 ohms, reading through resistance ground, etc., be 10 volts. Then $110 - 10 = 100$, $\frac{100}{10} \times 15,000 = 150,000$ ohms.

Working this by the first method, $\frac{110 \times 15000}{10} = 165,000$, deduct 15,000 = 150,000 ohms as before.

The chief advantage of the first method is that a constant can be figured for the particular instrument at a given voltage as described below.

The formula will be for the second method. Let V be e.m.f. of circuit, V_1 e.m.f. through resistance, r resistance of voltmeter then

$$\frac{V - V_1}{V_1} \times r = x$$

TABLE V.
Insulation Resistance.

V_1 or Voltage through Unknown Resistance.	V or Voltage of Circuit = 110 r or Resistance of Voltmeter = 16,500	V or Voltage of Circuit = 110 r or Resistance of Voltmeter = 17,000.
1	1,798,500	1,853,000
2	891,000	918,000
3	588,500	606,333
4	432,750	450,500
5	346,500	357,000
10	165,000	170,000
20	74,250	76,500
30	44,000	45,333
40	28,875	29,750
50	19,800	20,400
100	1,650	1,700
110	0	0

Resistance Table for Voltmeter Tests. Table V. is computed for ready reference in insulation or resistance tests with a Weston voltmeter or other instrument of high resistance.

The first column gives the reading through the resistance, the second and third columns the actual resistance of the insulation.

Both the latter columns are figured for a line e.m.f. of 110 volts and for an instrument resistance of 16,500 and 17,000 ohms respectively.

Where a number of readings have to be made it is handy to figure a constant, that is, multiply the resistance of the voltmeter by the e.m.f. of the circuit.

The result or constant then can be used by simply dividing it by the e.m.f. read with insulation in series and subtracting the resistance of the voltmeter.

Resistance of Voltmeter. This may be determined by one of the tests given to find resistance of a galvanometer. The one-half deflection is perhaps the simplest.

Connect voltmeter to a steady e.m.f. and read deflection.

Then add resistance in series with voltmeter until deflection is reduced to one-half. The added resistance will equal resistance of voltmeter.

This is clearly so because the one-half deflection is due to one-half current, and one-half current is due to doubling the resistance of voltmeter. In order to double the voltmeter resistance a resistance has been added which is equal to it.

Formula. Let V represent first deflection, V_1 the one-half deflection, R resistance of voltmeter, r resistance added.

Then $V_1 = \frac{V}{2}$ when $r = R$.

Testing e.m.f. around Armature. There are two methods of testing the distribution of e.m.f. around the armature and determining the equality of the magnetic field.

One is as shown in Fig. 95. H is an insulating handle carrying two carbon brushes $A B$ from which run wires $C D$ for the voltmeter connection.

$A B$ are set just far enough to span the insulation between two commutator segments.

By making contact on the commutator and

gradually moving H in the direction of rotation, the e.m.f. is tested at different points in the armature under the influence of various parts of the field.

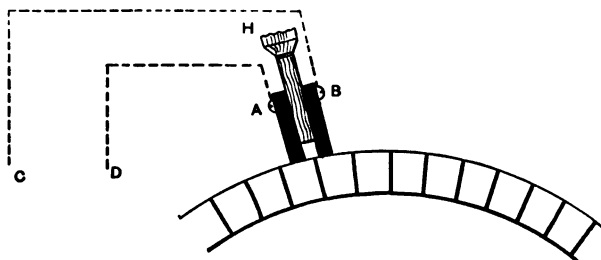


FIG. 95.

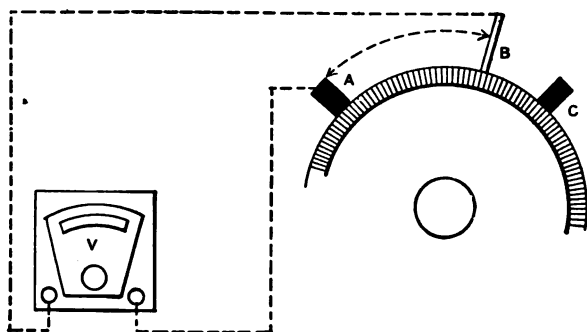


FIG. 96.

These e.m.fs. being plotted on a chart give a graphic record of the distribution of field flux.

Another method is Fig. 96, where one movable brush is used, the other voltmeter connection being on one of the regular working brushes.

This shows a gradual rise or fall of e.m.f. until the total e.m.f. of the generator is reached, but not so readily the actual e.m.f. of any one coil at a time.

Measuring Drop across a Lamp. To measure the drop or voltage expended in a lamp while burning the connections should be as in Fig. 97. L is the lamp, V the voltmeter and an ammeter A is shown

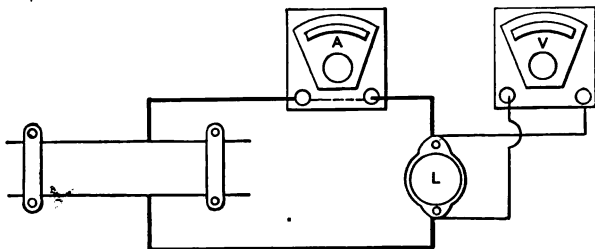


FIG. 97.

so that the current consumption can also be figured. The ammeter also is necessary if the resistance of L is to be determined.

The drop across L will be directly indicated by V . The current shown by A multiplied by the e.m.f. at L will give the watts expended in L .

Lamp Current Efficiency Test with Voltmeter. The amount of current taken by an incandescent

lamp on a given voltage depends upon the resistance of its filament.

Comparison of the hot resistance of one lamp with that of a standard lamp will therefore show the current efficiency of the first lamp.

If a low reading ammeter be available the current measurement may be made direct. But such an ammeter is not always at hand.

The drop across each lamp is to be measured with the voltmeter as in Fig. 97.

The e.m.f. indicated in each case will be proportional to the resistance of the lamp. And the current consumption will be in proportion to the resistance of each. The higher the e.m.f. across the lamp terminals, the higher will be its resistance and the lower its current consumption.

The e.m.f. of the testing circuit must be the same in both operations.

This test does not show the true efficiency of the lamp as no data is obtained of the relative candle powers.

And the resistance of L can be computed by dividing the current reading of A by the e.m.f. reading of V or $R = \left(\frac{E}{I}\right)$.

Suppose V reads 100 volts and A .5 ampere. Then $100 \times .5 = 50$ watts, and $100 \div .5 = 200$ ohms.

Testing a High Voltage with a Low Reading Voltmeter. This is accomplished by the fall of poten-

tial method. Suppose it is desired to measure the e.m.f. across a circuit of about 550 volts with a voltmeter reading to 150 volts.

Connect five 110 volt lamps L in series across the circuit, Fig. 98.

Then measure the e.m.f. between $A B$, $B C$, $C D$, $D E$ and $E F$ and the sum of these e.m.fs. will equal that between $A F$.

Care must be taken that the connections of the voltmeter are made so as to include all parts of the circuit in turn. For instance, when shifting from $A B$ to $B C$, the terminal from A must be

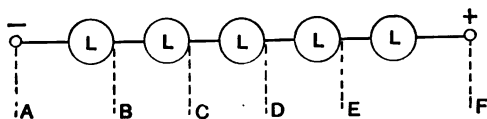


FIG. 98.

placed exactly at B where the other terminal of the voltmeter was. Or part of the e.m.f. in the wire or connection may be unmeasured, if this be neglected.

The lamps must all be kept burning during this test, or the voltmeter may be injured.

Temperature and Resistance. In testing coils of wire with their normal current flowing it will be found that the resistance will gradually increase, due to the current flowing in them. This increase is about .004 ohm per degree Centigrade for each

ohm of the coil (see Table VI). This is of very great importance in generator and motor tests. It would reduce the e.m.f. of a shunt generator owing to decreased field current. And it would speed up a shunt motor by reason of decreased field and decreased counter e.m.f.

TABLE VI.

Table Showing Increase in Resistance of Pure Copper Wire for Rise of Temperature above a Given Point.

Centigrade	Increase in ohms per ohm	Centigrade	Increase in ohms per ohm
1	.00389	21	.08169
2	.00778	22	.08558
3	.01167	23	.08947
4	.01556	24	.09338
5	.01945	25	.09725
6	.02334	26	.10102
7	.02723	27	.10503
8	.03112	28	.10894
9	.03501	29	.11281
10	.03890	30	.11680
11	.04279	31	.12059
12	.04669	32	.12450
13	.05051	33	.12837
14	.05447	34	.13226
15	.05835	35	.13615
16	.06225	36	.14061
17	.06613	37	.14593
18	.07030	38	.14782
19	.07391	39	.15171
20	.07780	40	.15575

This table is computed from an average between five authorities. The increase in ohms per ohm shows the fraction of an ohm which a wire one ohm in resistance would increase if its temperature were raised as per column of temperature. The coefficient .00389 is used but for ready calculation; .004 is used in the formulas, there is no absolute standard. To reduce Centigrade to Fahrenheit, multiply by 9, divide by 5, and add 32 to result.

All tests of generator and motor should record the temperature both of the air and machine.

The thermometer used for the air test should lie a few feet from the end of the shaft so as not to be influenced by air currents from a revolving armature.

And of course it should not be near any hot pipes or engine cylinders.

Temperature tests of field coils and armatures may be made with a thermometer laid on the part under test and covered with cotton waste. But such tests do not show heat inside the coils and are unreliable. A more reliable method is by taking advantage of the rise of resistance in the coil due to temperature rise.

The resistance of the coil cold may be measured and recorded. Its resistance is then measured from time to time and the temperature calculated.

For example, let the cold resistance of a shunt field coil be 120 ohms. One-half hour later it is 135 ohms.

This is a rise of 15 ohms, as it is the rise on 120 ohms, the temperature will be $\frac{15 \text{ ohms}}{120 \text{ ohms}} \div .004$ or 32° C.

A simple way to find the temperature rise is to divide the difference between the cold resistance and the hot resistance by the total resistance cold. Then find the nearest figure in the resistance column of Table VI. Against this will be the temperature. In this example

$135 - 120 = 15, \frac{15}{120} = .125.$ In Table nearest is .124 or 32° C.

A third method is divide the difference between the two resistances by the cold resistance and multiply by a constant 252 for the Centigrade rise.

Same example. $.125 \times 252 = 32.$

The temperature test by resistance shows the average temperature of the coils.

For test of the temperature at specific points, a special device may be prepared.

A strip of mica about two inches square is wound with 200 turns of No. 36 silk covered copper wire.

The strip is then fixed to a convenient handle if desired. This coil may be inserted in crevices between the coils of the armature, etc., its resistance being measured from time to time will give the temperature by computation as before.

Formula. Let R be the cold resistance, H be the hot resistance, $t C$ the temperature in Centigrade degrees, $t F$ the temperature in Fahrenheit degrees.

$$\text{Then } t C = \frac{H - R}{R} \times 252.$$

$$\text{And } t F = \frac{H - R}{R} \times 423.$$

Testing for a Break in a Coil. Testing for a break in a field coil of a dynamo or motor may be done by simply connecting a bell and battery in

series with the coil, Fig. 99, and trying if the bell rings. If it rings the circuit of course is complete.

But some field coils are of high resistance in which case a voltmeter, Fig. 100, is substituted for

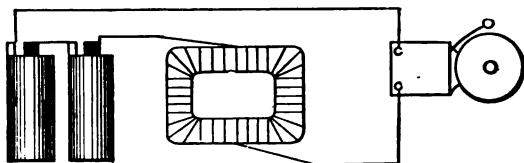


FIG. 99.

the bell. A lead may be taken from an electric light circuit to replace the battery. And if no definite readings are desired an incandescent lamp may replace the voltmeter.

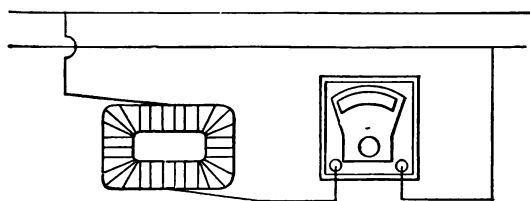


FIG. 100.

But the voltmeter method will show the extent of any fault in the circuit or the resistance of the latter. The resistance is calculated from the deflections by the method given elsewhere.

Testing the Resistance of Coils. If a number of coils are to be compared, as in a motor or dynamo, they may be tested without disconnecting them other than from the brushes, Fig. 101.

Pass current through the coils as they are in series, using a rheostat R if necessary, a bank of lamps will answer.

Measure the e.m.f. across the terminals of each coil, $A B$, $C D$, $E F$ and $G H$, and compare the readings; they should all be equal. If not, the

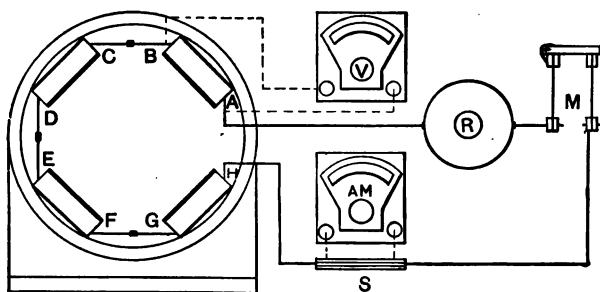


FIG. 101.

low reading coils are short circuited unless the high reading coils are faulty or open.

The actual resistance of each coil can be computed if an ammeter $A M$ be put in series to give the current value.

By ohms law the resistance will equal the e.m.f. divided by the current or $R = \frac{E}{I}$. For example,

if the readings show one ampere and 15 volts, the resistance will be $\frac{15}{1}$ or 15 ohms.

Testing Armature Coils. Armature coils being of low resistance can best be tested by the fall of potential method.

Current from an electric light circuit is passed through a number of coils by connecting the leads to segments situated some distance apart. The exact distance will depend upon the instruments available whether low or high reading. The e.m.f. is then read between segments or from one segment to the others in succession.

A rheostat or lamp bank should be inserted in series with the current leads.

If the e.m.f. between segments is uniform the coils and connections are good.

If a jump occurs at a segment, the voltmeter needle moving a greater distance than usual, a poor connection is probably the cause.

Testing Armature Resistance. The connections for this test are in Fig. 102. *S* is a switch connected to a source of current. *A* an ammeter, *V* a millivoltmeter, *L* one or more incandescent lamps for resistance purposes to control testing current. Field wires of a shunt machine are disconnected. Current from *S* passes through the armature coils, ammeter, and resistance lamps.

The connections may be made with copper wire making good contact on commutator segments by inserting them under brushes *B B* of opposite polarity.

The armature is stationary during this test.

The contact must be good, a strip of copper sheet laid flat on the segments is better than a mere wire. The copper strips should be held firmly down on the segments.

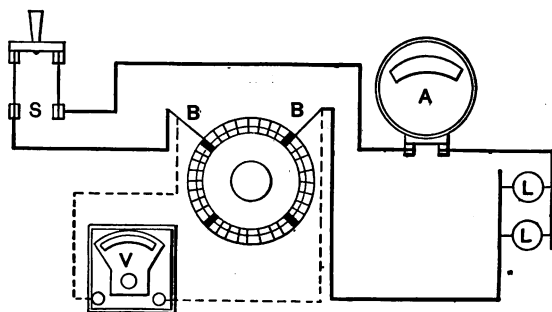


FIG. 102.

Readings are taken of the e.m.f. between the segments to which the connections of the current circuit are made.

Readings are also taken of the current indicated by the ammeter.

Then by ohms law, the resistance equals the e.m.f. divided by the current.

For example, let the current I be one ampere, the e.m.f. one-tenth of one volt, then one-tenth

divided by one equals one-tenth, or one-tenth of an ohm.

If the armature resistance is so low that a very small reading is given on the millivoltmeter, increase the current flow by adding more lamps in multiple.

The importance of good voltmeter contact at the commutator segments will be seen if a number of tests are made. Improve the contact in each case until no increase of deflection is noted on

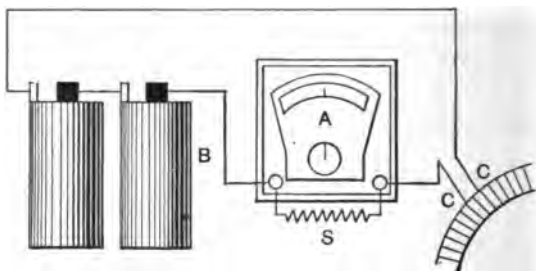


FIG. 103.

voltmeter. The larger the current through the coils the easier the test.

A form of test where an ammeter and battery are used is given in Fig. 103.

The milliammeter *A* is adjusted by means of a shunt *S* to give a large deflection when the terminal leads are short circuited. Its leads are then touched on adjacent segments of the commutator.

The deflections as compared with one another give the comparisons of coil resistance.

If a coil is open or a connection broken, the ammeter will indicate low reading when spanning its segments.

If one coil is partly open, the high resistance at the fault will give a low reading on the ammeter.

If a coil is short circuited, the deflection will be greater than that across a good coil. It will be equal of course to the deflection obtained by touching the ammeter leads together.

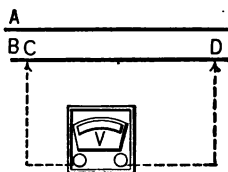


FIG. 104.

Voltmeter Test of Current in Circuit. It is often desired to know the current flowing in a lead when it is impracticable to cut the lead and insert an ammeter.

A very close approximation may be made with a low reading voltmeter by measuring the drop along a part of the wire.

The voltmeter is connected between two bared points on the cable as in Fig. 104. The resistance of the cable between these points is read from a table of copper wire resistance. The connections must be made with care and the leads to the in-

strument be large enough to interpose no appreciable resistance.

Take the case of a cable No. 0 B. & S. The resistance per thousand feet is .096 ohm.

The low reading or millivoltmeter is connected across ten feet and a reading observed of 14.4 hundredths of a volt.

As the resistance of ten feet is .00096 ohm, by ohms law = $\frac{.144}{.00096} = 150$ amperes.

This low reading is perfectly possible on the double scale instruments mentioned at the beginning of this chapter.

Another method would be to measure the e.m.f. across the line at two points. The first would be nearest to the generator or switchboard, the second at the point where the cable branched off to smaller wires.

The difference between the two readings would be due to the loss in the line and the loss beyond the point of second reading.

The loss in the line divided by the computed resistance of the line would give the current flowing.

The line resistance is of course its length multiplied by its resistance per thousand feet. The actual length of wire is to be used, that is twice the distance between the two points.

EXAMPLE. Cable 250 feet between points, in all 500 feet No. 0 B. & S. Resistance .048 ohm. E.m.f. at first point 115 volts at second point 110 volts; loss in line thus 5 volts.

$\frac{E}{R} = \frac{5}{.048}$ or neglecting fraction $\frac{5}{.05} = 100$ amperes.

The readings must be made as near simultaneously as possible, two adjusted voltmeters will be preferable although not absolutely necessary.

Corrections in Resistance by "Drop" Methods.

If great accuracy is desired the drop through the voltmeter may be allowed for. The current in the voltmeter will be found by dividing the e.m.f. indicated by the voltmeter resistance. This current is to be deducted from the ammeter readings.

In armatures where the coils are in series but tapped off at commutator segments, the path through the armature coils will be double between brushes. One path is through coils on top section and another path through bottom section coils.

If a coil were open in its windings, current would still flow to the testing instruments by the second path. But a jump of the needle would be shown on the voltmeter when spanning the break.

If a connection between coil and segment were broken, the ammeter would not show any less current except at the instant when the faulty segment was under the brush or connection.

Plotting Curves. A graphical method of recording tests is by plotting curves on sectional paper.

In Fig. 105 are shown the readings of e.m.f.

taken half way around the commutator of a bipolar generator.

The figures at the left hand side of the chart are the voltage readings. The figures along the bottom of the chart represent the segment number.

The readings along the horizontal lines are the abscissæ, those up the perpendicular lines, the ordinates.

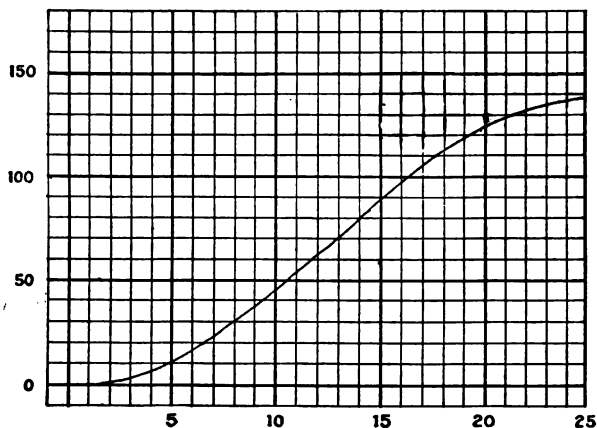


FIG. 105.

In making such a chart, the figures representing the voltage and segment numbers are first marked.

Then as a reading is taken a dot is made at the intersection of the lines representing the e.m.f. and the segment. When the total number of readings is made, the dots are connected by a line running through them.

For example, the first reading between brush and segment 4 is 5 volts, a dot is made on the vertical line corresponding to No. 4 segment, and at the point where a horizontal line from 5 volts would cut it.

A second reading of 10 volts on segment 5 is similarly marked, and another 15 volts at segment 6, and so on.

If the divisions represented by the abscissæ and ordinates are not close enough, computation may be made with a pair of dividers. But cross section paper for such charts is made with 8 and 10 divisions per inch

CHAPTER XII.

TESTING TELEPHONE LINES WITH THE VOLTMETER.

The use of the voltmeter in the testing work of the telephone exchange is rapidly becoming universal. This testing may be concerned with the actual telephonic apparatus or with the lines that radiate to the subscribers. The testing of telephone apparatus is not of particular interest here, but the testing of the lines occupies a very important place in the electrical field.

In general the tests may be divided as follows:

1. Talking test to ascertain condition of a subscriber line.
2. Ringing test to ascertain whether the bell of the subscriber's set rings satisfactorily.
3. To ascertain whether the line is unbroken.
4. Test for short circuits.
5. Tests for grounds.
6. Test for crosses between the telephone line and lines carrying electric light or power current.
7. Tests for crosses with other telephone lines.
8. Wheatstone bridge tests for grounds and crosses.

In order to carry out the foregoing tests, parts of the apparatus will be purely telephonic but the actual testing instruments may be either a galvanometer, resistance coils and a Wheatstone bridge, or one of the simpler yet fully adequate sets employing a direct reading voltmeter.

In the latter case much calculation is dispensed with, the liability of error reduced and the operations of testing the telephone lines of a large exchange brought within the reach of a person of ordinary intelligence after a short preliminary training.

Although the work can be successfully carried on with a voltmeter, battery, and some wire and switches, the operations involved will perhaps be better understood in a description of a standard equipment. The apparatus shown in Fig. 105a is made by the Western Electric Company and is known as the 1407 Testing Cabinet.

This testing cabinet is intended mainly for the more limited telephone exchanges where accurate tests are required but where an elaborate wire chief's desk is not desirable.

It may be mounted in any convenient location but its simplicity of operation and small size make it convenient for attachment at the end of any regular switchboard section.

The apparatus included comprises the following: a testing circuit with a round pattern voltmeter, an operator's telephone circuit, a ringing circuit and a grounding cord.

The voltmeter used is a Weston round pattern switchboard instrument having a regular resistance of 10,000 ohms and a shunt coil of 200 ohms. These



FIG. 105A.

even figures of resistance are of advantage by reducing the work of calculation, which, however, is already reduced to a minimum in this testing set.

The tests may be made at the switchboard or at the distributing frame as convenient.

The test at the switchboard is provided for by inserting the test plug in the jack of the line to be tested, while a pair of binding posts on the back of the cabinet permit of direct connection to the distributing board or terminal board.

In the latter case wires are run from the binding posts to the distributing frame where they terminate in a flexible cord and a test plug suitable to fit the protectors.

In Fig. 105b is a diagram of the connections in this test cabinet, the letters on the diagram referring as follows:

- V. G. Key to ground the voltmeter.
- V. S. Key to cut in the 200 ohm voltmeter shunt.
- B. Key to cut in the battery of 30 volts.
- V. Key which cuts in the voltmeter circuit.
- V. R. Key which reverses the voltmeter connections.
- G. Key to cut in a ground.
- R. Reversing Key.
- S. Key to test the switchboard wiring.
- L. Key to test the line through the distributing frame.

Tests 1 and 2. These are tests which are made with the telephone apparatus itself, and therefore need not be considered here.

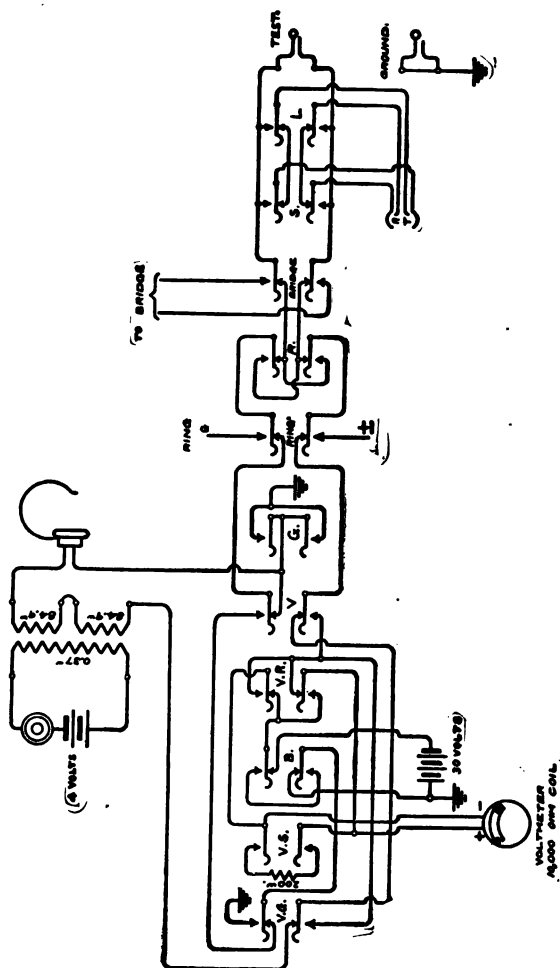


FIG. 105B.

Test 3. For continuity of a metallic line, the key V is operated, the appearance of the circuit being then as in Fig. 105c, which is drawn in a simplified manner.

The battery and the voltmeter being in series no deflection will of course appear if the line is open, but if the line is continuous, the needle of the voltmeter will be deflected. In testing a line with

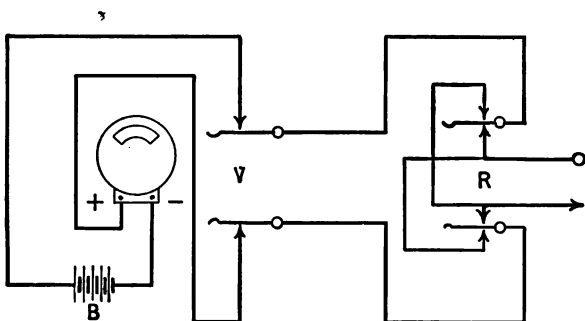


FIG. 105c.

ground return, the key VG being operated grounds the testing circuit. If the tip of the jack is connected to the line wire the reversing key R must be operated. In the case of condensers being used in series with the ringers of the subscribers' sets, there will be no permanent deflection. Key B is then operated in conjunction with key V or VG and will charge and discharge the condenser, causing momentary deflections of the voltmeter needle.

Test 4. A short circuit or path of low resistance will be indicated by the voltmeter deflection when key V is operated. The value of this resistance is computed as in the tests with the voltmeter previously described.

Key VS is provided to cut in or out the 200 ohm shunt. When the shunt is cut in multiple with the 10,000 ohm coil, the combined voltmeter resistance will be 196 ohms.

The deflection for a given line resistance will consequently be less when the shunt is used but is of advantage when testing low resistances.

Test 5. In testing for grounds on the lines, the necessary second ground is put on the testing circuit by the key VG. It is thus merely necessary to depress this key alone when the ground fault is being sought for on the ring side of the line. But the reversing key R must be operated when locating a ground on the tip side of the line.

If the voltmeter shows no deflection the line is entirely free from grounds. The value in ohms of a deflection indicating a ground may be determined by means of the voltmeter and a simple calculation as in test 4.

In the case of party lines and ringers operated with ground return, the resistance should be measured and compared with a similar standard line in good condition.

Test 6. In testing for the foreign currents introduced into a line by its contact with a telegraph line or a common battery telephone line, the testing battery is first cut out by the key B. Keys VG and R are then used as usual, but where the needle of the voltmeter is deflected backward, key VR must also be used to reverse the voltmeter connections.

Test 7. Test for crosses with other telephone lines. When it is suspected that a cross exists between two lines, one of them should be grounded by means of the grounding plug and the other line tested for ground as in test 5.

The revelation of a ground will now indicate a cross between the two lines. Naturally the line not grounded by the plug should itself be free from all grounds.

Where this is not possible measure the resistance of the ground on the line selected. The presence of the cross will then be indicated by the greater deflection of the needle.

Test 8. When it is desired to use a Wheatstone bridge in connection with this cabinet the X terminals of the bridge should be connected to the binding posts on the face of the cabinet. The key BRDG being then operated will connect the line to the bridge.

Ballistic tests may be made to determine the

capacity of the line or of attached condensers by first charging the line and then discharging it through the voltmeter.

EXAMPLE. To test the metallic circuit of a subscriber where there is a condenser in series with the ringer, the voltmeter key and the grounding key should be thrown, then the reversing key moved quickly back and forth. The capacity of the line will be shown by the deflection of the voltmeter as explained when treating of condenser testing elsewhere in this book. Grounded lines will show a deflection on charge only.

The ballistic tests may also be used to locate a break in a well insulated line.

TABLE VII

DATA ON LEAD COVERED TELEPHONE CABLE
(Western Electric Company)

As a measure of transmission efficiency, a No. 19 B & S gauge cable having a mile loop resistance of 88 ohms and a mutual electrostatic capacity of .060 microfarad per mile is used as a basis. With the present improved apparatus 30 miles of this cable is considered the maximum distance over which commercial transmission can be secured.

One mile of this cable equals approximately:

No. 12 BWG BB Iron Wire	3.4 mile circuit
No. 10 " " " "	4.3 " "
No. 14 NBS hard drawn bare cop- per wire	8.0 " "
No. 12 B & S hard drawn bare cop- per wire	8.0 " "
No. 12 NBS hard drawn bare cop- per wire	13.4 " "

Electrostatic Capacity

Grounded electrostatic capacity is obtained by measuring one wire against all the other wires grounded to the sheath.

Mutual electrostatic capacity is obtained by measuring one wire against its mate, constituting a pair, all other wires being grounded to the sheath.

TABLE VII—(Continued)
WESTERN ELECTRIC LEAD COVERED CABLES

Type	Conductor B. & S.	Per Mile		Insulation Resistance Meg. Ohms	Insulation	Pairs
		Average Mutual Electro- static Capacity	Average Grounded Electro- static Capacity			
AA	22	.070	.105	500	single paper	5 to 500
BB	19	.074	.111	500	"	5 to 300
H	16	.072	.108	500	"	10 to 150
J	13	.072	.108	500	"	10 to 75
P	22	.08	.12	500	double paper	5 to 400
R	22	.08	.12	500	single paper	5 to 400
S	22	.083	.125	500	"	600
W	19	.08	.12	500	single or double paper	5 to 300
Y	22	.07	.105	500	"	5 to 400
Z	22	.08	.12	500	"	5 to 400

TABLE VIII
(Bureau of Standards. S. W. Stratton, Director)
STANDARD ANNEALED COPPER WIRE
(British) Standard Wire Gage

Gage No.	Diameter in Mils	Circular Mils	Ohms per 1000 Feet	
			15.6° C (=60° F)	65° C (=149° F)
0000	400.	160 000.	.063 69	.076 33
0	324.	105 000.	.097 08	.1163
1	300.	90 000.	.1132	.1357
2	276.	76 180.	.1338	.1603
3	252.	63 500.	.1605	.1923
4	232.	53 820.	.1893	.2269
6	192.	36 860.	.2764	.3313
8	160.	25 600.	.3981	.4771
9	144.	20 740.	.4914	.5890
10	128.	16 380.	.6220	.7454
12	104.	10 820.	.9422	1.129
13	92.	8464.	1.204	1.443
14	80.	6400.	1.592	1.908
15	72.	5184.	1.966	2.356
16	64.	4096.	2.488	2.982
17	56.	3136.	3.250	3.894
18	48.	2304.	4.423	5.301
19	40.	1600.	6.369	7.633
20	36.	1296.	7.864	9.424
21	32.	1024.	9.952	11.93
22	28.	784.0	13.00	15.58
23	24.	576.0	17.69	21.20
24	22.	484.0	21.06	25.23
25	20.	400.0	25.48	30.53
26	18.	324.0	31.45	37.69
27	16.4	269.0	37.89	45.41
28	14.8	219.0	46.52	55.76
29	13.6	185.0	55.10	66.03
30	12.4	153.8	66.28	79.43
31	11.6	134.6	75.74	90.77
32	10.8	116.6	87.37	104.7
33	10.0	100.0	101.9	122.1
34	9.2	84.64	120.4	144.3
35	8.4	70.56	144.4	173.1
36	7.6	57.76	176.4	211.4
37	6.8	46.24	220.4	264.1
38	6.0	36.00	283.1	339.3
39	5.2	27.04	376.9	451.7
40	4.8	23.04	442.3	530.1

The *fundamental resistivity* used in calculating the table is the Annealed Copper Standard, viz., 0.153 28 ohm (meter, gram) at 20° C. The *temperature coefficient* for this particular resistivity is $\alpha_{20}=0.003\ 93$, or $\alpha_0=0.004\ 27$. However, the temperature coefficient is proportional to the conductivity, and hence the change of resistivity per degree C is a constant, 0.000 597 ohm (meter, gram). Hard-drawn copper may be taken as about 2.7 per cent. higher resistivity than annealed copper. Ohms per mile, or pounds per mile, may be obtained by multiplying the respective values above by 5.28.

TABLE IX
PROPERTIES OF COPPER CLAD WIRE

S. & S. GAUGE NUMBER	AV. RESISTANCE INT. OHMS PER 1000 FEET				AV. RESISTANCE INT. OHMS PER MILE AT 75° F.			
	Copper Clad				Copper Clad			
	GRADE				GRADE			
	Average 80%	Average 40%	Average 47%		Average 80%	Average 40%	Average 47%	
0000	0.170	0.127	0.108		.90	.67	.57	
000	0.213	0.160	0.136		1.13	.85	.72	
00	0.270	0.202	0.172		1.43	1.07	.91	
0	0.340	0.255	0.218		1.80	1.35	1.15	
1	0.433	0.325	0.277		2.30	1.72	1.46	
2	0.543	0.407	0.347		2.87	2.15	1.83	
3	0.687	0.515	0.438		3.63	2.72	2.32	
4	0.867	0.650	0.553		4.57	3.43	2.92	
5	1.09	0.817	0.695		5.77	4.33	3.67	
6	1.37	1.03	0.876		7.26	5.45	4.63	
7	1.74	1.31	1.11		9.20	6.90	5.86	
8	2.20	1.65	1.40		11.63	8.73	7.42	
9	2.77	2.08	1.76		14.63	10.97	9.35	
10	3.47	2.60	2.21		18.30	13.72	11.69	
11	4.37	3.27	2.79		23.	17.25	14.68	
12	5.50	4.12	3.51		29.	21.75	18.53	
13	6.97	5.22	4.45		36.70	27.57	23.45	
14	8.80	6.60	5.62		46.47	34.85	29.63	

TABLE X

PROPERTIES OF IALA RESISTANCE METAL MANUFACTURED BY
HERMANN BOKER & COMPANY

Temperature coefficient for 1° C. (soft annealed) +.000005.

Temperature coefficient for 1° C. (hard drawn) —.000011.

Specific resistance per centimeter cube at 15° C. (soft annealed)
47.1 micro ohms.

Specific resistance per centimeter cube at 15° C. (hard drawn)
50.2 micro ohms.

This material has a uniform quality and does not deteriorate, is absolutely non-corrosive and has 29 times the resistance of copper. Used in instruments and electrical devices where extreme accuracy, reliability and permanency are required.

Owing to its extreme low temperature coefficient (soft annealed Iala+.000005), it is especially adapted for use in measuring instruments and shunts.

SIZE B. & S. Gauge	DIAMETER Inches	AREA Circular Mils	RESISTANCE	
			Ohms per Foot	Ohms per Pound
12	.0808	6,530.	.0461	2.32
14	.0641	4,107.	.0735	5.90
16	.0508	2,583.	.1169	15.0
18	.0403	1,624.	.1859	37.8
20	.0320	1,024.	.2956	95.5
21	.0285	812.8	.3744	151.
22	.0253	640.1	.4701	241.
23	.0225	506.2	.5966	386.
24	.0201	404.	.7476	608.
25	.0179	320.4	.9456	975.
26	.0159	252.8	1.193	1,550.
27	.0142	201.6	1.498	2,460.
28	.0126	158.8	1.890	3,900.
29	.0113	127.7	2.408	6,255.
30	.0100	100.	3.005	9,850.
31	.0089	79.2	3.789	15,650.
32	.0080	64.	4.779	25,000.
33	.0071	50.4	6.025	39,750.
34	.0063	39.7	7.600	63,100.
35	.0056	31.6	9.583	99,980.
36	.0050	25.	12.081	159,600.
37	.0044	19.8	15.229	253,000.
38	.0040	16.	19.213	404,000.
39	.0035	12.3	24.218	642,000.
40	.0031	9.6	30.570	1,018,000.

(Several gauges omitted from Table.)

PROPERTIES OF VARIOUS RESISTANCE WIRES

Manufactured by Driver Harris Wire Co.

"Nichrome" Resistance Wire.—Resistance per Mil-Foot 600 Ohms at 75° F. (24° C.). Temperature Co-efficient .00024 per degree F. Specific Gravity 8.15. Weight per cubic inch .29 lb.

"Nichrome" II Resistance Wire.—Resistance per Mil-Foot 660 Ohms at 75° F. (24° C.). Temperature Co-efficient + 0.00009 per degree F. Specific Gravity 8.02. Weight per cubic inch 0.29 lb.

"Climax" Resistance Wire.—Resistance per Mil-Foot 525 Ohms at 75° F. (24° C.). Temperature Co-efficient .0004 per degree F. Specific Gravity 8.137. Weight per cubic inch .29 lb.

Pure "Nickel."—Resistance per Mil-Foot 64.3 Ohms at 75° F. (24° C.). Specific Gravity 8.9. Weight per cubic inch .32 lb.

"Advance" Resistance Wire.—Resistance per Mil-Foot 294 Ohms at 75° F. (24° C.). Temperature Co-efficient NIL. Specific Gravity 8.9. Weight per cubic inch .32 lb.

"Therlo" Resistance Wire.—Resistance per Mil-Foot 280 Ohms at 75° F. (24° C.). Temperature Co-efficient .000031 per degree F. Specific Gravity 8.15. Weight per cubic inch .29 lb.

"Yankee Silver."—Resistance per Mil-Foot 200 Ohms at 75° F. (24° C.). Temperature Co-efficient .000086 per degree F. Specific Gravity 8.6. Weight per cubic inch .31 lb.

18% "German Silver."—Resistance per Mil-Foot 218 Ohms at 75° F. (24° C.). Temperature Co-efficient .00017 per degree F. Specific Gravity 8.5. Weight per cubic inch .307 lb.

"Ferro-Nickel."—Resistance per Mil-Foot 170 Ohms at 75° F. (24° C.). Temperature Co-efficient .00115 per degree F. Specific Gravity 8.2. Weight per cubic inch .29 lb.

TABLE XI

DIFFERENCE BETWEEN THE MOST COMMONLY USED WIRE GAUGES, IN DECIMAL PARTS OF AN INCH

No. of Wire Gauge	B. & S.*	B. W. G.† Also called Stub's Wire Gauge	Washburn & Moen Also known as Steel Wire Gauge Am. Steel & Wire Co.'s Gauge	New British Standard N. B. S.‡
0000	.46	.454	.393	.4
000	.40964	.425	.362	.372
00	.3648	.38	.331	.348
0	.32495	.34	.307	.324
1	.2893	.3	.283	.3
2	.25763	.284	.263	.276
3	.22942	.259	.244	.252
4	.20431	.238	.225	.232
6	.16202	.203	.192	.192
8	.12849	.165	.162	.16
10	.10189	.134	.135	.128
12	.080808	.109	.105	.104
14	.064084	.083	.08	.08
16	.05082	.065	.063	.064
18	.040303	.049	.047	.048
19	.03589	.042	.041	.04
20	.031961	.035	.035	.036
21	.028462	.032	.032	.032
22	.025347	.028	.028	.028
23	.022574	.025	.025	.024
24	.0201	.022	.023	.022
25	.0179	.02	.02	.02
26	.01594	.018	.018	.018
27	.014195	.016	.017	.0164
28	.012641	.014	.016	.0148
29	.011257	.013	.015	.0136
30	.010025	.012	.014	.0124
31	.008928	.01	.0135	.0116
32	.00795	.009	.013	.0108
33	.00708	.008	.011	.01
34	.006304	.007	.01	.0092
35	.005614	.005	.0095	.0084
36	.005	.004	.009	.0076
38	.003965008	.006
40	.003144007	.0048

*Known as American Wire Gauge. Generally used for copper wires.

†Generally used for iron and steel wires.

‡Also called "Standard Wire Gauge (SWG) and English Legal Standard or Imperial Wire Gauge. Used abroad.

APPENDIX.

Testing Telegraph Wires and Cables and Locating Faults in Telegraph and Telephone Wires and Cables.

BY

JESSE HARGRAVE

Asst. Electrical Engineer Postal-Telegraph-Cable Company.

With 28 new Diagrams.

CHAPTER XIII.

TESTING TELEGRAPH WIRES AND CABLES.

Early Morning Tests. All wires leading from each terminal station are tested out regularly every morning not later than seven o'clock for continuity, earth contacts, or "grounds," and crosses. The wire chief simply calls up the distant chief on some wire known to be intact and tries out first one wire and then another until all have been tested. Failing to get the distant station on any particular wire, if it is found that the testing relay fails to open while the distant end stands open, then the wire is in contact with the earth or else mixed with some other wire. If the testing relay remains open while the distant end is to opposite battery pole or to the ground, then the wire is open. If the wire be mixed with some other wire it is usually quickly found out by noticing a disturbance between the two or the development of some other symptom with which the testing chief very soon becomes familiar. Fig. 106 shows the testing relay standing closed to an intermediate ground while the distant end is open.

Fig. 107 shows the testing relay standing open to a break while the distant end is grounded.

When a wire has been found to be grounded or open, each intermediate testing station, beginning at the distant end and proceeding in regular order,

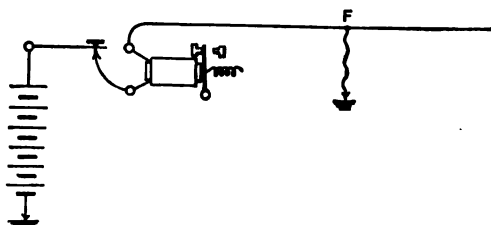


FIG. 106.

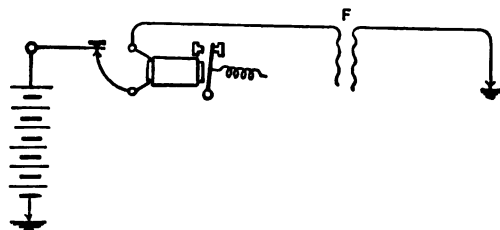


FIG. 107.

is told to open or ground the wire until the location of the fault is definitely determined between stations.

In the case of crossed wires, the distant station is told to open both wires, and a battery is then applied to one, and the other is grounded through the testing relay. If the battery on one wire closes

the relay on the other wire, it is quite obvious that it does so through a cross between the testing station and the distant station's opening. Successive stations are then called up in regular order and told to open the wires until the fault is localized. Fig. 108 shows the testing relay closed through the cross.

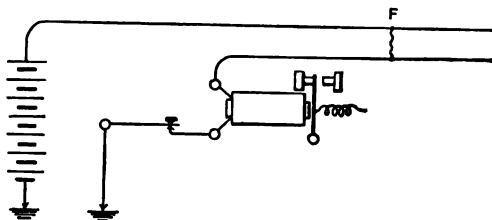


FIG. 108.

Wrecks. It frequently happens that all wires become mixed by reason of a tree falling across the line or other cause. Such conditions are known as "wrecks." It usually takes some little time for the wire chief to test the wires out and determine definitely just what conditions exist at the wreck—that is, what wires are crossed with each other, which are open or grounded, etc. The writer has found that the test outlined in Fig. 109 to be the quickest and surest way of testing out such trouble. Each wire in turn is connected to the test relay whose other terminal is connected to ground. All other wires are then opened and a moderately strong battery is placed to one at a time. Those that are

crossed with the wire under test will close the relay the moment the battery is applied to them, provided the wires are not "dead grounded." The spring of

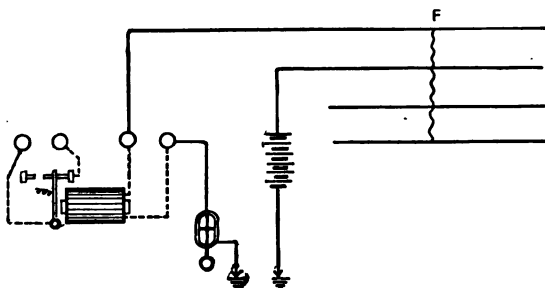


FIG. 109.

AUTOMATIC SIGNAL WHEN WIRES COME O. K.

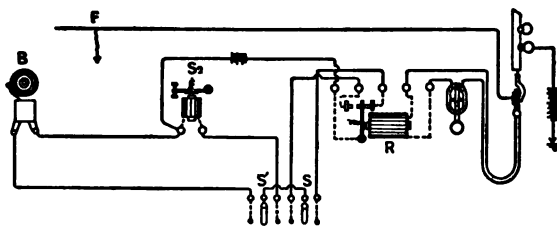


FIG. 110.

relay should be left reasonably slack while this test is being made.

When an "opening" or "ground" is located on a wire by any of the methods described above, it is

customary for the wire chief to order the lineman upon whose section the fault is located to hunt for it, giving all the information possible which will assist him in finding and removing it with the least possible delay. It is then very important that the wire chief know the instant that the wire is cleared of the trouble, and in order to accomplish this result the arrangement outlined in Fig. 110 is much used by one of the large telegraph companies. If the wire be "grounded" the first station beyond F is instructed to leave it open. Relay R is then cut into the wire as shown and switches S and S^1 thrown to the left, which throws bell B into circuit on back contact of relay R . So long as the wire remains grounded at F relay R remains closed and its armature rests on its front contact, thus leaving the bell circuit open. The moment the lineman removes the ground, however, relay R opens and rings bell through its back armature contact. The wire chief then reverses both switches, throwing sounder S_2 into circuit on front contact of R , and proceeds to test out the wire in the usual manner and assign it to a circuit.

Should the wire be open instead of grounded, the distant station is instructed to keep it to his ground, switch S is thrown to the right and switch S^1 to the left. It will then be seen that the moment the lineman closes the break R will close to the distant station's ground, thus ringing bell B through its front contact. Both switches are then placed

to the right, and the chief proceeds to test out the wire in the usual manner.

Locating Grounds by Wheatstone Bridge Measurements. When a ground has been located on a wire by the method described above it is often desirable to make a closer and more exact location of it by electrical measurement in order that every facility may be afforded the lineman in finding and clearing the trouble in the shortest possible time. In making this test the Varley loop measurement described in Chapter X is usually employed. A good wire is looped with the faulty one at the next station beyond the fault, as shown in Fig. 90. The mode of procedure is then the same as described under head of "The Varley Test" on pages 157-8. The values of the ratio arms A and B are usually made equal and should be that value which more nearly approaches the resistance of the loop under measurement.

For example, if the looped wires measure 400 ohms, then 100 ohms should go in arms A and B , whereas if loop measures 600 ohms, then 1000 ohms should be used in the arms of bridge. Great care should be taken in this, as in all other bridge tests, to insure a proper battery strength. If dynamo current be used, a reducing rheostat should invariably be placed in circuit and sufficient resistance should be cut in as to obviate any danger of overheating the coils of the bridge. It is sometimes

advisable to cut a milli-ammeter in circuit with the battery so that a check can be kept on the current entering bridge.

Measurement for Crosses using Varley Test. If a good, clear wire is available it should be looped

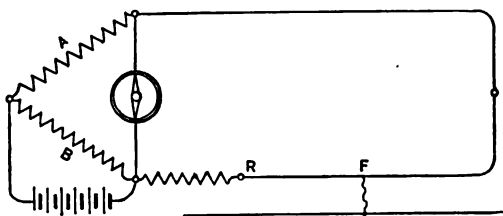


FIG. 111.

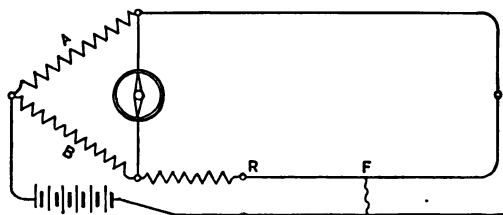


FIG. 112.

with one of the crossed wires at the first convenient station beyond the fault and a measurement made of the loop, as in Fig. 111.

Another measurement is then made as in Fig. 112.

It will be seen that the battery instead of being led to the loop through the distant fault via ground, as is shown in Fig. 90, is carried via the crossed wire to point F . It sometimes happens that the end of the second crossed wire is not available for joining on to battery, and in this case the distant station at which the wires have been looped is told to ground the second crossed wire, and the battery is then grounded as in Fig. 113. It should be borne in mind that in this case the earth serves the same purpose of conducting the battery to the loop via the fault as did the second crossed wire in Fig. 112, the path of current being shown by the small arrows.

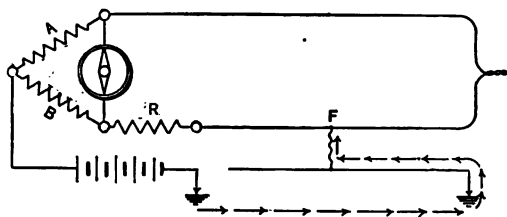


FIG. 113.

The formula for this test is the same as given on page 158, or, boiled down, it is simply $F = \frac{L - R}{2}$ where F is resistance to the fault, L is resistance of loop as per Fig. 111, and R is result of measurement as per Fig. 112 or Fig. 113. The above is, of course, assuming that equal values are used in

arms *A* and *B* of bridge, which is almost always the case in measurements of overhead wires. The reason for this is that, as one ohm represents the fractional part of a mile of overhead wire, it is not necessary, except in unusual cases, to obtain closer locations by means of multiplied ratios in arms *A* and *B*. In No. 9 B. & S. gauge copper wire one ohm represents a little less than a quarter of a mile, whereas in No. 8 gauge iron wire it represents a little less than one-twelfth of a mile.

When the value of *F* has been obtained by the above formula it is easy to arrive at the distance by dividing this result by what the wire measures in ohms per mile.

Although the Varley loop test only is spoken of in connection with the foregoing tests for crosses and grounds, yet it will readily be seen that the Murray test described in Chapter X is applicable. The Murray test is preferred by some by reason of its simplicity, especially where the two looped wires are of the same kind and gauge, in which case the initial, or loop, resistance measurement is dispensed with and the result is figured out directly in miles. It should be borne in mind in connection with this test, however, that any chance inequality in resistance of the two wires or imperfect connection or contact at distant stations or elsewhere would disturb its correctness. For this reason it is perhaps safer to figure the result in ohms and then divide by the ohms per mile to get at the distance.

Measurement for Crosses Using the Two Crossed Wires Only. A study of Figs. 112 and 113 will soon convince the reader that the resistance of the cross between the two wires can be neglected. So long as it furnishes a path for sufficient current to enter the loop in order to give a readable galvanometer deflection on an unbalanced condition of one or two ohms, then it does not matter whether its resistance be 1 or 1000 ohms so far as affecting the result is concerned. It is this fact which makes this particular measurement so reliable and to be preferred to a simple resistance measurement through the cross.

FIRST MEASUREMENT.

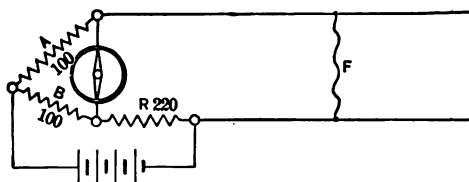


FIG. 114.

When a third good wire is not available for this measurement, however, a reasonably reliable location of the trouble can be made by tests outlined in Figs. 114 and 115. In these tests the resistance of the fault is taken into account, and so long as it does not change between the two tests it does not alter the result.

SECOND MEASUREMENT.

The simple formula for this test is $F = \frac{L - r}{2}$, where F is resistance to fault, L is result of measurement as per Fig. 114, and r is result of measurement as per Fig. 115. The explanation of this test is that the resistance necessary in rheostat in Fig. 115 to make a balance simply places point X at the theoretical center of the loop formed by this resistance and the two wires through F . It is obvious, then, that to deduct this resistance from the loop measurement of Fig. 114 and then halve the result gives the resistance to point X .

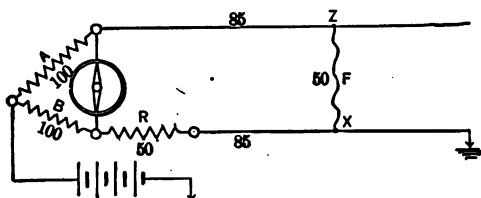


FIG. 115.

If in Fig. 114 we obtain a balance with 220 in R and in Fig. 115 a balance is obtained with 50, then $220 - 50$ gives us 170, and this divided by 2 gives us 85 as resistance to X . Let us further assume that the wires are of the same kind and gauge; then it follows that 85 ohms is also the resistance between A and Z , which leaves us 50 ohms for the fault. Then $50 + 85$ on one side of X equals $50 + 85$

on the other side, which balances the system and proves the result to be correct.

Fig. 116 outlines another method of locating a cross by employing the two crossed wires only. In this test it will be seen that the resistance of F is neglected, but the resistance between wire a and ground at distant station figures in the result and cannot be accounted for. "Earth currents," or difference in potentials between the earth at points Gr and G , would cause considerable error in this measurement, and for this reason it is not recommended, although the writer has at times made use of it with more or less success.

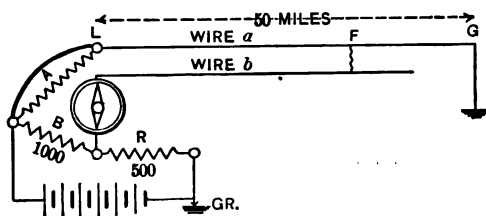


FIG. 116.

In arranging for this test the next distant station beyond fault F is asked to ground one of the wires, and this wire is connected to point L of bridge. Arm A of bridge is then shunted out with a piece of copper wire or else plugged up, and that terminal of galvanometer which normally connects to point L is detached and connected to the

second crossed wire *b*. It will then readily be seen that that section of wire *a* between *L* and *F* bears the same ratio to the section between *F* and *G* as arm *B* bears to the resistance unplugged in *R* to obtain a balance. It will also be seen that *a* can either be reckoned as having so much ohmic resistance or as being so many miles in length. For example, assuming that the distance to point *G* is 50 miles, then the distance to *F* is $\frac{50 \times 1000}{1500}$ or $33 \frac{1}{3}$ miles.

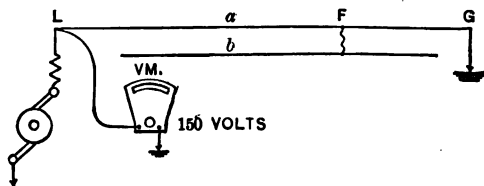


FIG. 117.

Locating a Cross by Voltmeter Test. A simple volt-meter measurement for a cross on two wires, but one which is not recommended for its reliability, is outlined in Figs. 117 and 118. Were it not for errors which are likely to arise by reason of earth currents this test would be convenient and useful. The same objections hold against it, however, as are cited against the test described in the foregoing paragraph. A volt-meter of very high resistance is necessary in this test.

If the potential at point L in Fig. 117 is found to be 150 volts, then it is obvious that there is a gradual drop of 150 volts along wire a until point G is reached, where it falls to zero provided G is a neutral ground and there is no difference of potential between it and the ground at the testing station. It then follows that if the potential has fallen from 150 at L to 100 volts at fault F , as indicated in Fig. 118, it has dropped through one-third of its value

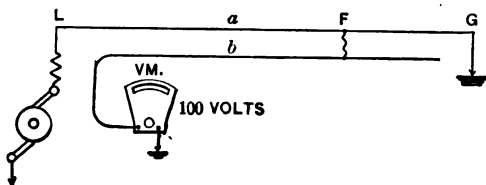


FIG. 118.

between L and G , and if the distance between those points be 100 miles it is further obvious that the distance between L and F is $33\frac{1}{3}$ miles, or

$$\frac{150 \text{ volts} - 100 \text{ volts} \times 100 \text{ miles}}{150 \text{ volts}} = 33\frac{1}{3} \text{ miles.}$$

It will be noted that the accuracy of this test is affected by the ratio that the resistance of the wire b between F and VM bears to the resistance of VM . If, however, the resistance of VM be very high, say sixty thousand or more ohms, then the error due to this cause would be slight. The test, however, is a rough one at best and is not recom-

mended except where the instruments for the other more reliable tests are lacking.

Insulation Tests—Milliammeter Method. It is customary to make periodical tests of the insulation resistance of telegraph and telephone wires. Up to a few years ago these tests were made with the tangent galvanometer almost exclusively, but with the advent of direct-reading voltmeters and milliammeters, quicker and in every way more satisfactory methods for making them were found.

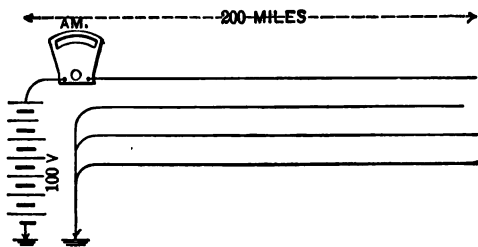


FIG. 119.

The milliammeter furnishes an accurate and ready means of making these measurements, the results of which are directly figured out by Ohm's law. The simple method of making the tests is illustrated in Fig. 119, where the wires are all shown open at the distant station. The wire chief then quickly inserts his meter in first one and then the other, applying an appropriate battery as shown,

until he has run through the entire number on that particular pole line. It will be noted that all wires are grounded except the one under test. This is to not only measure the amount of leakage between the wire under test and earth, but also any leakage which may exist between it and any of the other wires.

The test is ordinarily made with a double-scale milliammeter, employing the most sensitive coil, which usually records on the lower scale. This lower scale is usually calibrated in divisions of .2 milliampere (or .0002 ampere). Referring to Fig. 119, suppose that with a pressure of 100 volts a deflection of 2 divisions on "lower" scale is noted. This would denote that a current of four-tenths of a milliampere (.0004 ampere) was passing out to the "open" wire from 100 volts' pressure. Then,

by Ohm's law, $R = \frac{E}{I}$; hence $100 \text{ volts} \div .0004 \text{ ampere} = 250,000 \text{ ohms}$, which would represent the total insulation resistance of the wire. In considering this test the mileage insulation resistance is always reckoned, and this is arrived at by multiplying the length of the wire in miles by the total or absolute insulation resistance. Assuming that our wire in this instance is 200 miles in length, then the insulation per mile would be 50,000,000 ohms, or 50 megohms.

Insulation Test by Voltmeter Method. This test is made in the same manner as described in Fig. 119,

substituting the voltmeter for the milliammeter. The total insulation resistance is arrived at by methods described in Chapter XI, and the mileage insulation resistance is then computed as shown in last paragraph.

When the presence of an abnormal leakage is disclosed by these tests it is customary for the testing chief to localize the same by means of sectional tests; that is, by having successive intermediate stations open the wire for test until the leakage is located between two of them. The lineman who has charge of that section is then ordered to clear it.

Conductivity Tests. Conductive resistance tests are for the purpose of checking up joints and connections and, in the case of iron wire, the rate of deterioration from year to year. Some of the big wire companies have these tests made semi-annually, usually in the extremes of temperature, such as in July and January. The wires are looped at the distant station and simple resistance measurements made, such as shown in Fig. 111, for example.

In order to arrive at the exact resistance of each particular wire it is customary to measure three wires in loop combinations three times. For example, suppose we want to arrive at the exact resistance of each of three wires numbered 1, 2, and 3. We loop them in the following order:

1 and 2 = 400 ohms.

1 and 3 = 450 “

2 and 3 = 500 “

As each wire has been measured twice, we add the three results and halve their sum. Then if we deduct the result of any one of the three loop measurements from half of the sum of all three results it gives us the resistance of the wire which is not involved in the loop deducted. For example, the sum of the above three measurements is 1350, the half of which is 675. Now, if from this we deduct the loop measurement of 2 and 3, which was 500 ohms, it gives us the resistance of No. 1 = 175; and so, if we deduct the loop measurement of 1 and 3, which was 450, it gives us 225 ohms as the resistance of No. 2, etc.

Any one of the wires whose resistance is thus arrived at may then be used in a loop measurement with any other wire, whose resistance can then be definitely determined.

When the resistance of each wire has been definitely arrived at by the above method, the results are then divided by their length in miles and their resistance per mile recorded, due allowance having been made for all cable in circuit, etc.

Whenever the resistance of any particular wire is found to be abnormally high, sectional measurements are made to determine the location of the added resistance, which usually consists of a bad joint or loose contact. This sectional measurement is accomplished by having successive intermediate stations loop the wires for measurement.

CHAPTER XIV.

LOCATING FAULTS IN TELEGRAPH AND TELEPHONE CABLES.

This subject has been pretty thoroughly covered in the chapter on "Cable Testing" and there is little left to be said. There are a few useful points, however, which have not been covered and which it is thought best to call attention to.

Location of Grounds and Crosses by Varley Method Using Multiplied Arm Ratios. Telegraph and telephone cables are usually of such short lengths that it is necessary to make very close locations of faults in them in order that the measurements may be of much real value. This is especially true of aerial cables, and more especially true of aerial cables carrying important trunk-line telegraph wires. If trouble on a few of the conductors in such cables can be successfully located and cleared without interrupting the working circuits on the other conductors, then the test becomes of great value.

In order to do this, a multiplied ratio must be used in arms *A* and *B* of bridge in the Varley test. The arm values are usually made 10 in *A* and 1000 in *B*, which gives a ratio of 1 to 100.

In making this test the operator should be very careful that he is getting perfect contacts in all of his bridge connections. If his bridge is of the plug type he should see to it that all plugs and holes are perfectly clean and bright. If there is any sign of oxidization in either, then the plugs should be cleaned with very fine sandpaper, such as 00, and a wooden peg, such as the end of a wooden pen staff, should be forced into each plug receptacle and turned around several times. If a radial-type bridge is used, the stubs should be bright and clean, as should also the under side of the radial arm which makes contact with them. The operator should be

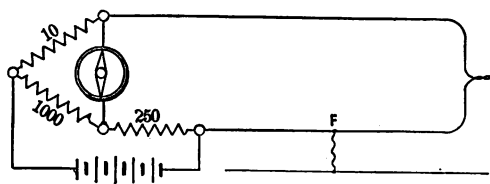


FIG. 120.

absolutely sure that everything is all right before proceeding, and if he does this he will frequently save much extra trouble and annoyance.

Suppose that we have two crossed conductors in a cable 1000 feet long. One of the conductors involved is looped to a good conductor at one end of cable, and a measurement is made from the other

end. A simple loop resistance measurement is first made, as in Fig. 120. Say that our result is 250.

According to the formula for this test, $r = \frac{A \times R}{B}$,

where r is the loop resistance, $A = 10$, $R = 250$, and $B = 1000$. We get 2.5 ohms for our loop.

The connections are then made as in Fig. 121.

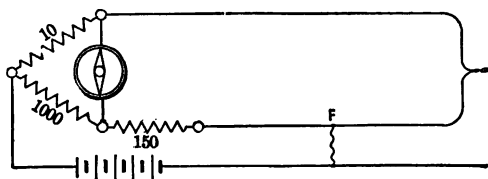


FIG. 121.

If it now takes 150 ohms in R to effect a balance, then $F = 2.5 - \frac{(2.5 + 150) \times 10}{1000 + 10} = .99$, where F is resistance to the fault. This divided by what the conductor measures per foot would give the distance to the fault.

Let us assume in this instance that our conductor is a No. 14 B. & S. gauge copper wire which measures approximately .0025 ohm per foot at normal temperature. Then our distance to fault would be $.99 \div .0025 = 396$ feet.

In making this test the operator should invariably verify his measurement by reversing the wires

leading to his bridge and make a third test, seeing whether or not this test checks correctly with the previous one. Fig. 122 shows the connections for this "check" measurement, the formula for which is

$$F = \frac{(L + R) \times A}{A + B}, \text{ or } \frac{(2.5 + 97) \times 10}{1000} = .99, \text{ where}$$

F represents the distance to the fault, L the loop measurement as in Fig. 120, R the resistance it takes to balance in this measurement, and A and B the arms of the bridge.

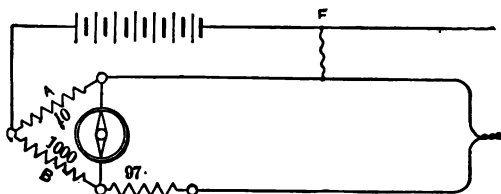


FIG. 122.

It will readily be seen that instead of connecting the battery to the second crossed conductor as shown in Figs. 121 and 122, it could just as well be grounded and then the second crossed conductor grounded, as is shown in the case of overhead-wire measurement, Fig. 113. In cases where the conductors are grounded instead of crossed the measurements are made in the same manner as described above, except that in Figs. 121 and 122 the battery

terminal is grounded instead of being connected to the second crossed wire.

How to Find Trouble after Located. It does not, perhaps, belong within the scope of this work to deal with cable faults beyond describing electrical methods necessary in their localization. It might, however, prove helpful to a great many to briefly describe one or two very useful and simple methods of finding the trouble after it has been measured for. This is not so important in the case of underground cables, where it is usually only necessary to know between which manholes the trouble lies. In aerial lead cable, however, it is a matter of considerable importance.

It rarely happens that a cross between two conductors of a lead cable is a solid or metallic one. There is usually a medium of charred paper or carbonized compound which forms the cross. This is equally true of conductors crossed with the lead sheath of the cable. If, now, a current of some considerable volume, say anywhere from 100 to 250 milliamperes, can be made to traverse the cross for any length of time it will be found that the lead sheath will quickly rise in temperature at the point where the trouble exists. It is, therefore, only necessary, after a measurement has been made, to ground one conductor of a crossed pair and apply a grounded source of e.m.f. of 100 or more volts to the other crossed wire, and then have the sheath

examined for a distance of ten or fifteen feet on either side of the point where the measurement placed the fault, until the warm spot is found.

This method has frequently been made use of to great advantage where the sheath of cable bore no sign to indicate the location of the fault. There are comparatively few cases of trouble in aerial lead cable where the method will not work out, and it makes the repairs possible without severing a conductor, the faulty conductors being simply separated, their insulation repaired, and a split sleeve wiped on over the bared section.

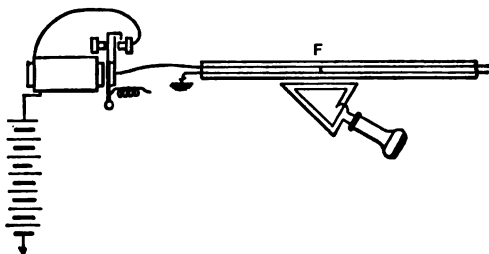


FIG. 123.

Another method which is referred to in some of the text-books (see Young's *Work on Electrical Testing*) is to send a moderately heavy intermittent current into the faulty conductor and then search for the fault by passing a coil of insulated wire in series with a telephone receiver along in close prox-

imity to the sheath. If the current be sufficiently strong, and if the interruptions are of sufficient frequency, then a coil of wire of a conveniently large number of convolutions should pick up enough of the current by induction to produce a noise in the telephone receiver, and this noise should begin to diminish when the point of trouble is passed. Fig. 123 outlines a convenient plan for applying this method.

This method also presents a convenient means of identifying a faulty cable among a number of others in a common subway, or in the case of submarine cables in midstream when they have been brought up by grappling irons.

Varley Test by Leeds & Northrup's Dial Testing Set. In this bridge the arms *A* and *B*, instead of consisting of the conventional 10, 100, and 1000 ohm units, are composed of a fixed set of coils so arranged that a radial arm, representing the point of divide and carrying one pole of the battery, in moving about among them creates the usual proportions. The bridge is also arranged so that a throw of certain switches places the organization in position for the Varley, Murray, or simple loop resistance measurement.

It will be noticed in Fig. 124, which shows the theoretical arrangement of the bridge, that there is no chance for imperfect contacts in arms to give rise to multiplied error. The only possible imper-

fect contact is between the radial arm and stubs, and this would throw the added resistance in the battery circuit where it would do no harm.

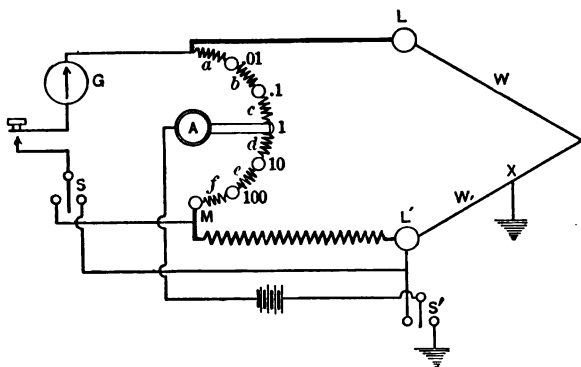


FIG. 124.

The resistances in the arms are as follows: $a = 9.901$ ohms, $b = 81.008$ ohms, $c = 409.091$ ohms, $d = 409.091$ ohms, $e = 81.008$ ohms, $f = 9.901$ ohms. Total, 1000 ohms. It will be found by adding the resistances on each side of any given ratio point that the marked ratio obtains when the battery divides at that point.

Let us now suppose that we are measuring for a ground in a cable or conductor W , which we have looped with a good conductor W' at the distant end of cable. Let us call the result of a loop resistance measurement with all switches to left L ; then with switch S still to the left, and switch S' to the

right let us call the Varley measurement R , then $F = \frac{L - DR}{D + 1}$, where F is resistance to fault and D is the ratio dial reading. It will be noticed that this formula is very simple and quickly handled. For example, take the measurement described in Figs. 120, 121, and 122, where for L we got 2.5 ohms, and for R we got 150 ohms, then with our ratio arm on the .01 point we would read

$$F = \frac{2.5 - (.01 \times 150)}{.01 + 1}, \text{ or } \frac{1}{1.01} = .99 \text{ ohm.}$$

Fig. 125 shows the actual arrangement of this bridge.

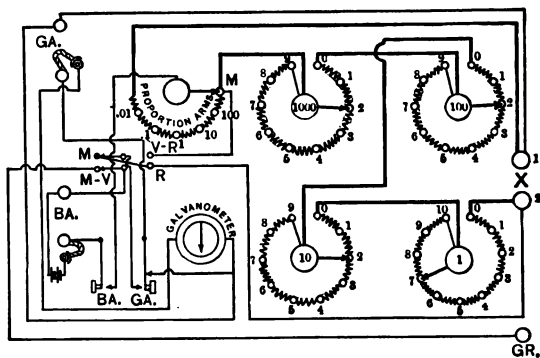


FIG. 125.

Murray Test with Leeds & Northrup Dial Bridge.

Now let us throw both switches to the right and place arm A on point M . If the student will trace

out the connections he will see that the arrangement is identical with that outlined in Fig. 91. It will also be found that we have 1000 ohms, or the entire ratio arms resistance as our fixed ratio factor. If now we obtain a balance on 656 ohms and call this Result R , then

$$F = \frac{R \times L}{1000 + R}, \text{ or } \frac{656 \times 2.5}{1000 + 656} = .99 \text{ ohm.}$$

Locating Openings in Cable Conductors by Bridge Method. Several methods for locating cable conductor openings have been described in different text-books, but the following will perhaps be found to be about the simplest and most reliable :

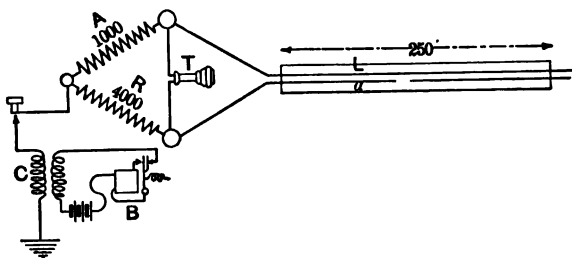


FIG. 126.

In Fig. 126 A and R are the arms of the bridge, R being made the rheostat resistance so as to get a wider range of adjustment than would be afforded by the B arm. C is a small induction coil, such as is used in telephones. B is an ordinary buzzer con-

nected in series with the primary of C and a few cells of dry battery. When a balance is obtained there will be no noise in telephone receiver T , and A will then bear the same inverse ratio to R as L does to a , hence the formula

$$F = \frac{A \times L}{R}, \text{ or } \frac{1000 \times 250}{4000} = 62.5 \text{ feet,}$$

where F equals the distance to opening on conductor a , L the length of conductor and R the balancing resistance.

Another arrangement for localization of this class of trouble is outlined in Fig. 127.

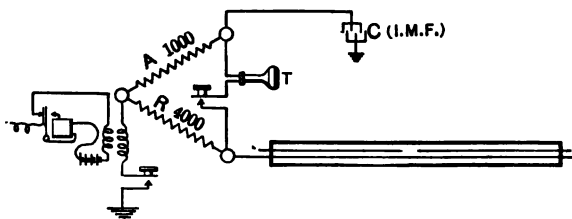


FIG. 127.

In this case C is a standard condenser, against whose capacity we are measuring the capacity of the good conductor in cable. Let us call the capacity of C 1 M.F. Then say it takes 4000 ohms in R to balance the system so that there will be minimum noise in T . Then

$$C' = 1 \times \frac{A}{R}, \text{ or } \frac{1 \times 1000}{4000} = .25 \text{ M.F.}$$

This gives the capacity of the good conductor, and it is then only necessary to find the capacity of the broken conductor up to the break to ascertain how far out the fault lies.

Let us say that a like measurement of the open conductor gives us .125 M.F. and that the cable is 7500 feet long, then it is obvious that our fault is $\frac{.125}{.250}$ of 7500 feet, or 3750 feet distant.

In order to obviate any error which might arise by reason of a difference in the static capacity of the broken and good conductor it would be well to make a third test involving the good conductor looped to the open one at the distant end. This would give us the capacity of the one that is open beyond the fault. Let us call the result of this measurement .380 M.F.; then it is obvious that the open conductor has a slightly greater capacity per foot than the good one and our formula would be

$$F = \frac{A L}{A + (C - B)} \text{ or } \frac{125 \times 7500}{125 + (380 - 250)} = 3676 \text{ ft.},$$

where F equals distance to fault, A equals capacity of open conductor, B equals capacity of good conductor, C equals capacity of good conductor looped to open conductor at distant end, and L equals length of cable.

It will doubtless occur to the reader that tests B and C can be dispensed with if the open conductor is measured from both ends.

In order that these tests for open conductors may be accurate it is necessary that the opening be a clean break and clear of escape.

The Leeds & Northrup Fault Finder. This instrument has been designed with a view of reducing calculations connected with fault location to a minimum, also so as to simplify the manipulation as much as possible.

It may be used to measure conductor resistance to locate faults by four distinct tests and to locate opens using a buzzer and telephone. (With it a rough measurement can also be made of the resistance of faults so that the operator can decide how closely he should be able to make the location.)

Resistance Measurement with the Leeds and Northrup Fault Finder.

The essential feature of the apparatus is the uniform resistance AB , which is wound in a circle and is about 100 ohms. By a special construction it is arranged so that contact can be made at any point along it, and it is, therefore, equivalent to a high resistance wire. It has a moving contact C and a scale of 1000 divisions. In series with this there are the two resistances E and R . E has exactly the same resistance as the wire AB . R has a resistance of 100 ohms. Either resistance may be short-circuited by a small switch. The resistances shown

between the ground post and the battery and between post *BA* and the battery key are simply to protect the battery and the apparatus from excessive current. Fig. 128 shows the proper connections for measuring conductor resistance. As in the ordinary slide wire bridge, the resistance *X* between the

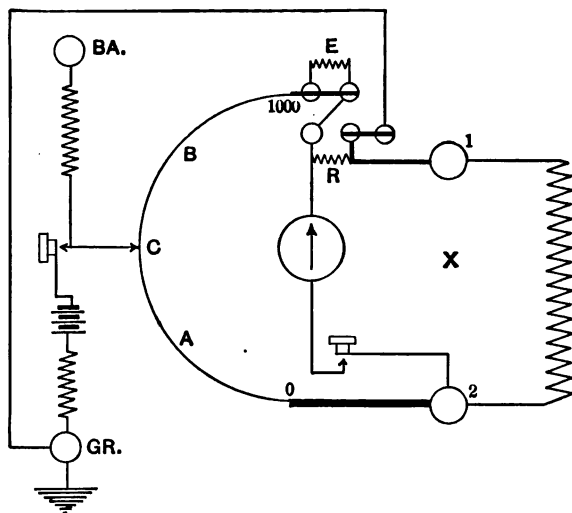


FIG. 128.

two posts 1 and 2 is gotten from the formula $X = \frac{A}{1000 - A} R$. To avoid the necessity of solving in each case the fraction $\frac{A}{1000 - A}$, a table is furnished with the set, giving the value of this fraction

for each value of A . The resistance is accordingly determined in each case by simply setting the contact C for a balance and reading from the table the value opposite the number on the scale and multiplying by 100.

Fault Location with the Leeds & Northrup Fault Finder.

FIRST METHOD.

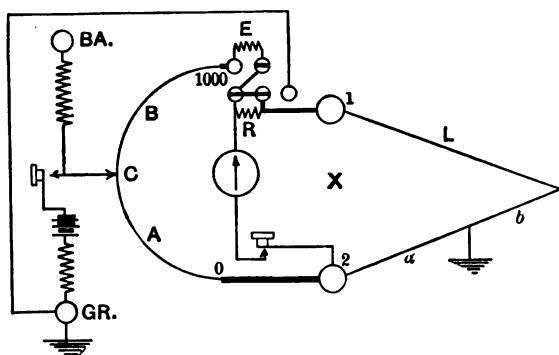


FIG. 129.

This method is to be used in locating faults where there are two wires having equal resistance, in one of which there is a fault. Connect and set switches as shown in Fig. 129. It will be remembered that E is equal to the resistance of the wire AB . From the symmetry of the arrangement it will be obvious that the contact point C would rest for a balance at 1000 on the scale if the fault were exactly at the junc-

tion between the good and the bad wires ; it would rest at 500 if the fault were half way along the bad wire ; and at whatever point it comes to rest, the reading, divided by 1000 and multiplied by the length of the bad wire, is the distance from the instrument to the fault.

SECOND METHOD.

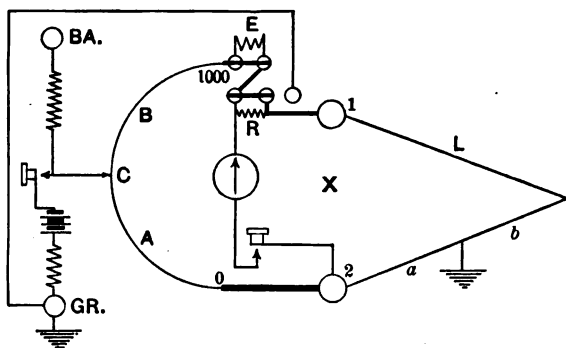


FIG. 130.

This method is to be used for locating faults where the good and the bad wires are not equal to each other. The connections are shown in Fig. 130. It is the ordinary Murray loop, and it will readily be seen that the resistance a to the fault will be gotten from the formula $a = \frac{A}{1000} L$, where L is the resistance of the loop and A is the reading of the contact C on its scale.

THIRD METHOD.

This method may be used as a check on either of the above. The connections are shown in Fig. 131.

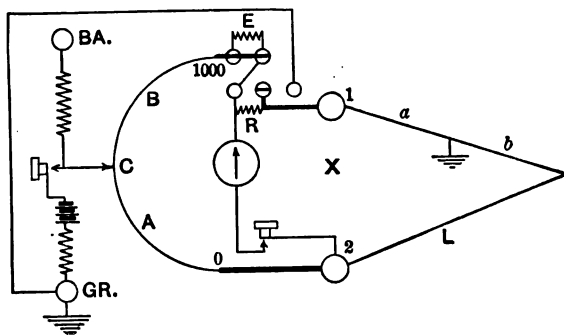


FIG. 131.

The resistance a to the fault from the binding-post 1 is $a = \frac{BL - 100A}{1000}$, where L equals the total resistance of the loop, A equals the reading of contact C , and $B = 1000 - A$.

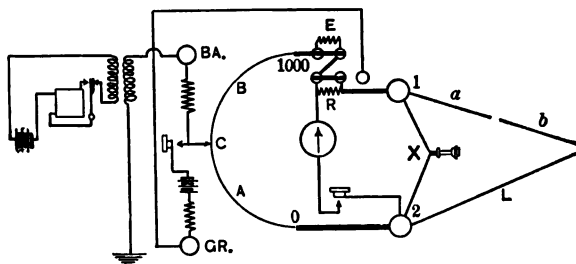


FIG. 132.

To Locate Opens, Using Buzzer and Telephone.

In this case the slide resistance circle AB is used without either of the resistances E or R . Connect to binding-post 1 the broken wire and to binding-post 2 a perfect wire having the same capacity per mile. Connect buzzer as shown and telephone X between the binding-posts 1 and 2. For a mini-

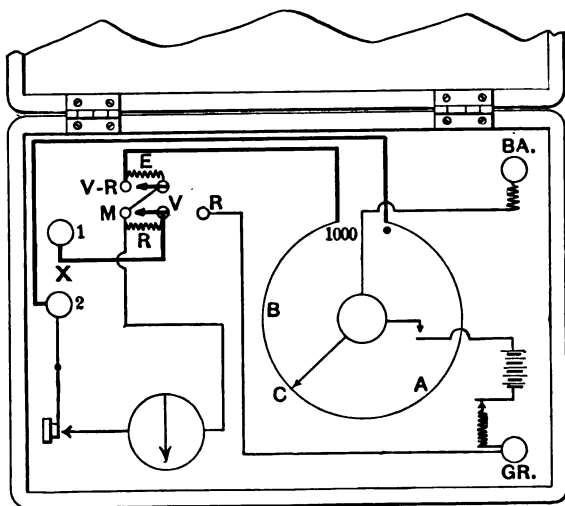


FIG. 133.

mum sound in the telephone we will have the relation $\frac{A}{B} = \frac{a}{L}$, where a is the length of the broken wire and L that of the whole cable. From this, $a = \frac{A}{1000 - A} L$. As in the case of resistances, the

value $\frac{A}{1000 - A}$ can be read from the table. The connections are shown in Fig. 132.

Fig. 133 shows the actual arrangement of the Leeds & Northrup Fault Finder.

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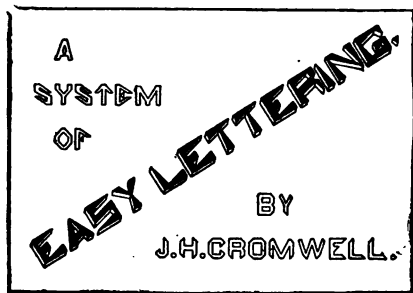
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